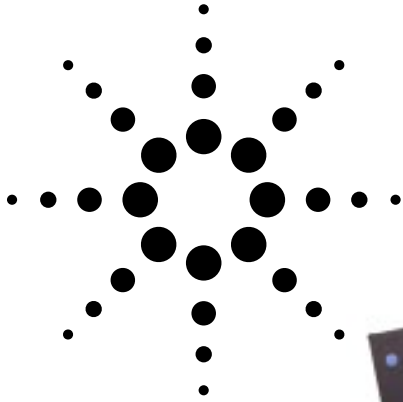# [Obsolete product]

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Agilent HDMP-1685A
1.25 Gbps Four Channel SerDes
with 5-pin DDR SSTL_2
Parallel Interface
Data Sheet

Features
- **5-bit wide Tx, Rx bus pairs**
- **208-ball, 23 mm TBGA package**
- **Parallel data I/O and clocks compatible with SSTL_2 (EIA/JESD8-9)**
- **125 MHz TC, RC clocks**
- **One TC clock for 4 channels**
- **Single or paired RC clocks**
- **LVTTL RefClk input**
- **Source synchronous clocking of transmit data**
- **Source centered clocking of receive data**
- **Double data rate (DDR) parallel transfers**
- **Parallel loopback**
- **Differential BLL serial I/O**
- **Single +3.3 V power supply**
- **Copper drive capability**

Applications
- **High density fast ports**
- **Fast serial backplanes**
- **Clusters of computers**
- **Clusters of network units**
- **Link aggregation, trunks**

Functional Description

This data sheet describes HDMP-1685A, a 1.25 Gbps, four-channel, 5-pin per channel parallel interface SERDES device. The HDMP-1685A 5-pin parallel interface device enables a single ASIC to drive twice as many channels using half as many parallel lines. This is accomplished without increasing the clock frequency by utilizing the bandwidth on the parallel interface more efficiently.

The HDMP-1685A SERDES is a single silicon bipolar integrated circuit packaged in a 208-pin BGA. This integrated circuit provides a low-cost, small-form-factor physical-layer solution for multi-link 1.25 Gbps cables or optical transceivers. Each IC contains transmit and receive channel circuitry for all four channels.

A 125 MHz LVTTL reference clock must be supplied to the reference clock input pin, RFCT.

The transmitter section accepts four, 5-bit-wide parallel SSTL_2 data (TX [0:3] [0:4]), a 125 MHz SSTL_2 byte clock (TC) and serializes them into four high-speed serial streams. The parallel data is expected to be "8B/10B" encoded data, or equivalent. TX and TC are source synchronous. New data are accepted on both edges of TC; this is called Double Data Rate (DDR). HDMP-1685A finds a sampling window in between the two edges of TC to latch TX [0:3] [0:4] data into the input register of the transmitter section.

This timing scheme assumes that the driving ASIC and HDMP-1685A operate in the same clock domain. 8B/10B encoded data comes in 10-bit characters. This data is latched onto the 5 TX pins of each channel in 5-bit groups. It is expected that the beginning half of each 10-bit character is latched on the rising edge of TC.

The transmitter section's PLL locks to the 125 MHz TC. This clock is then multiplied by 10 to generate the 1250 MHz serial clock for the high-speed serial outputs. The high-speed outputs are capable of interfacing directly to copper cables for electrical transmission or to a separate fiber optic module for optical transmission.

The receiver section accepts four serial electrical data streams at 1250 MBd and recovers the respective original 10-bit-wide data for each channel over a 5-pin parallel interface. The receiver PLL locks onto the incoming serial signal and recovers the high-speed serial clock and data. The serial data is converted back into 10-bit parallel data, optionally recognizing the 8B/10B comma character to establish byte alignment. If comma character detection is enabled by raising the SYNC signal high, the receiver section is able to detect comma characters and indicate their presence on each channel with the appropriate SYN [0:3] signal(s).

The recovered parallel data are presented at SSTL_2-compatible outputs RX [0:3] [0:4], and a pair of 125 MHz SSTL_2 clocks, RC [0:3] [1], and RC [0:3] [0], that are 180 degrees out of phase from one another and which represent the remote clock for that channel. Rising edges of RC [0:3] [1] and RC [0:3] [0] may be used to latch RX data at the destination. Alternatively, both edges of either RC [0:3] [1] or RC [0:3] [0] may be used to latch RX data (DDR). When SYNC is high, the beginning half of the comma character shows up at the rising edge of RC [0:3] [1].

The timing of transmit and receive parallel data with respect to TC and RC [0:3] [0:1] is arranged so that the upstream protocol device can generate and latch data very simply. Specifically, in the TX direction, the ASIC drives four sets of 5-pin TX lines and the TC line with the same timing. The TC line is similar to a 6th data line that is always toggling to provide timing information to the SERDES. On the RX side, the SERDES drives four sets of 5-pin RX data centered between the edges of RC [0:3] [1] or RC [0:3] [0].

For test purposes, the transceiver provides for on-chip parallel loopback functionality controlled through an input pin. Additionally, the byte-edge alignment feature via detection of the positive comma (K28.5+) character may be disabled. This may be useful in proprietary applications that use alternative methods to align the parallel data.

HDMP-1685A Block Diagram

The HDMP-1685A (Figure 2) is designed to transmit and receive 10-bit 8B/10B character data over 5-pin-wide parallel busses via high-speed serial communication lines. The parallel data applied to the transmitter is expected to be encoded per the 8B/10B encoding scheme with special reserve characters for link management purposes. Other encoding schemes will also work as long as they provide dc balance and sufficient number of transitions. In order to accomplish this task, the HDMP-1685A incorporates the following:

- SSTL_2 Parallel Data I/O
- High-Speed Phase Locked Loops
- Parallel-to-Serial Converters
- High-Speed Serial Clock and Data Recovery Circuitry
- Comma Character Recognition Circuitry (K28.5+)
- Byte Alignment Circuitry
- Serial-to-Parallel Converter

PARALLEL INPUT LATCH

For each channel, the transmitter accepts 10-bit characters as two groups of 5-pin single-ended SSTL_2 parallel data at inputs TX [0:3][0:4]. The SSTL_2 TC clock provided by the sender of transmit data is used for all channels as the transmit byte clock. The TX [0:3][0:4] and TC signals must be properly aligned, as shown in Figure 3.

TX PLL/CLOCK GENERATOR

The transmitter Phase Locked Loop and Clock Generator (TX PLL/CLOCK GENERATOR) block generates all internal clocks needed by the transmitter section to perform its functions. These clocks are based on the transmit byte clock (TC). TC is also used to determine the sampling window for the incoming data latches. Incoming data is synchronous with TC (see Figure 3).

FRAME MUX

The FRAME MUX accepts the 10-bit-wide parallel data from the INPUT LATCH. Using internally generated high-speed clocks, this parallel data is multiplexed into the 1250 MBd serial data streams. The data bits are transmitted sequentially, from TX [0] to TX [4]. The leftmost bit of K28.5 is on TX [0].

SERIAL OUTPUT SELECT

The OUTPUT SELECT block provides a parallel loopback mode for testing purposes. In normal operation, PLUP is set low and the serialized TX [0:3] [0:4] data are placed at SO [0:3] +/-.

When parallel wrap-mode is activated by setting PLUP high, the SO [0:3]+/- pins are held static at logic 1 and the serial output signal reflecting TX [0:3] [0:4] data is internally wrapped to the INPUT SELECT block of the receiver section.

SERIAL INPUT SELECT

The INPUT SELECT block determines whether the signal at SI [0:3]+/- or the internal loopback serial signal is used to drive RX [0:3] [0:4]. In normal operation, PLUP is set low and the serial data is accepted at SI [0:3]+/-.

When PLUP is set high, the outgoing high-speed serial signal is internally looped back from the transmitter section to the receiver section. This feature allows parallel loopback testing, exclusive of the transmission medium.

RX PLL/CLOCK RECOVERY

The RX PLL/CLOCK RECOVERY block is responsible for frequency and phase locking onto the incoming serial data stream and recovering the bit and byte clocks. It does this by continually frequency locking onto the 125 MHz reference clock, and then phase locking onto the selected input data stream. An internal signal detection circuit monitors the presence of the input, and invokes the phase detection once the minimum differential input signal level is supplied (AC Electrical Specifications).

Once bit locked, the receiver generates the high-speed sampling clock at 1250 MHz for the input sampler.

SERIAL INPUT SAMPLER

The INPUT SAMPLER converts the serial input signal into a retimed bit stream. In order to accomplish this, it uses the high-speed serial clock recovered from the RX PLL/CLOCK RECOVERY block. This serial bit stream is sent to the FRAME DEMUX AND BYTE SYNC block.

FRAME DEMUX, BYTE SYNC

The FRAME DEMUX AND BYTE SYNC block is responsible for restoring the 10-bit character from the high-speed serial bit stream. This block is also responsible for recognizing the comma character (K28.5+) of positive disparity (0011111xxx). When recognized, the FRAME DEMUX AND BYTE SYNC block works with the RX PLL/CLOCK RECOVERY block to properly select the parallel data edge out of the bit stream so that the comma character starts at RX[0:3][0]. When a comma character is detected and realignment of the receiver byte clock RC[0:3][0:1] is necessary, this clock is stretched, not slivered, to the next possible correct alignment position. This clock will be fully aligned by the start of the second 2-byte or 4-byte ordered set. The second comma character received will be aligned with the rising edge of RC[0:3][1]. Comma characters of positive disparity must not be transmitted in consecutive bytes to allow the receiver byte clocks to maintain their proper recovered frequencies.

PARALLEL OUTPUT DRIVERS

The OUTPUT DRIVERS present the recovered 10-bit character in two groups onto the 5-pin RX bus, properly aligned to the receive byte clock RC [0:3] [0:1] as shown in Figure 5. These output data buffers provide single-ended SSTL_2 compatible signals. Unlike the TX, where all four channels are driven with the same transmit byte clock (TC), each receive channel provides its own clock aligned with its own data, so the recovered clocks may not be phase aligned.

SSTL_2 COMPATIBILITY

HDMP-1685A works with protocol devices whose VDDQ voltage is nominally set at 2.5 Volts. RX [0:3][0:4], RC [0:3][0:1] pins generate output voltages that are compatible with the SSTL_2 standard (EIA/JESD8-9). In addition, these devices provide a VREFR output pin allowing the receiving device to differentially detect a high or a low. The devices receive inputs on their TX [0:3][0:4] and TC pins that are also SSTL_2 compatible. The VREFT input pin is driven by a voltage divider whose supply voltage is at the same level as the VDDQ supply of the protocol device. This allows differential detection of a high or a low at TX parallel inputs.
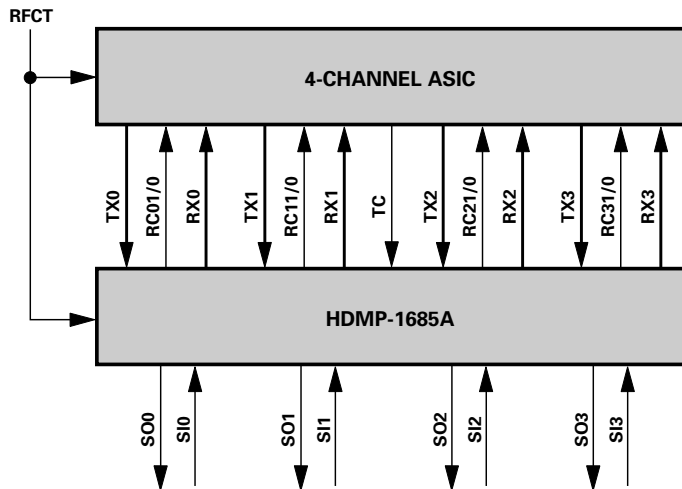
Figure 1. Typical application of the four channel SERDES.
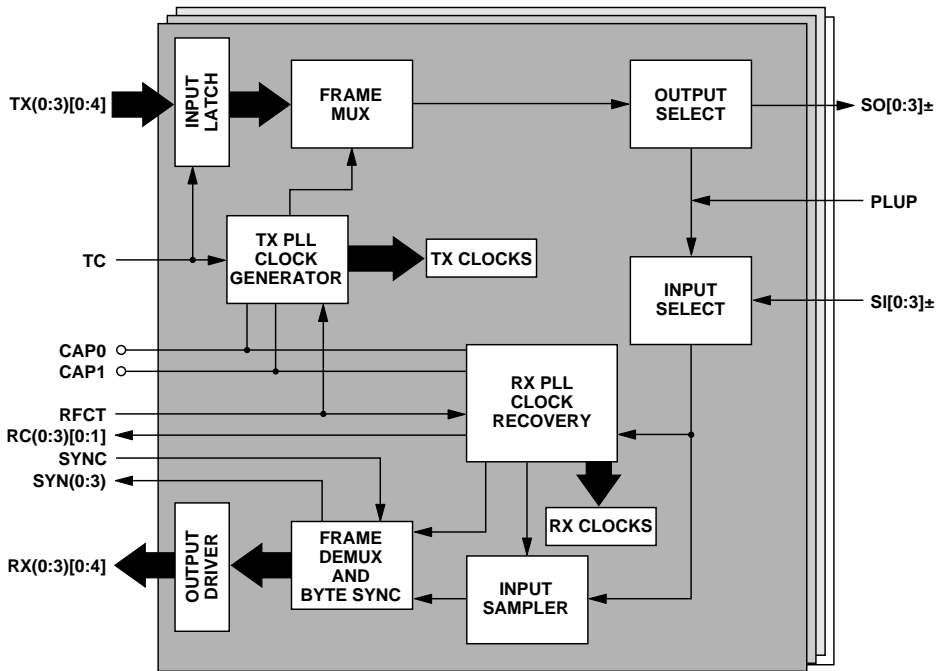
Figure 2. Block diagram of HDMP-1685A.

4

HDMP-1685A Timing Characteristics – Transmitter Sections

$T_A = 0°C$ to $T_C = 85°C$, $V_{CC} = 3.15$ V to 3.45 V

Symbol	Parameter	Units	Min.	Typ.	Max.
t_{TXCT}[2]	TX [0:3][0:4] Input Data and TC Clock Transition Range	ps			1600
t_{TXCV}[2]	TX [0:3][0:4] Input Data and TC Clock Valid Time	ps	2400		
t_txlat[1]	Transmitter Latency	ns		4	
		bits		5	

Note:

1. The transmitter latency, as shown in Figure 4, is defined as the time between the leading edge of the first half of a parallel 10-bit word and the leading edge of the first transmitted serial output bit of that 10-bit word.

2. Agilent's HDMP-1685A internally generates another clock which is 90 degrees out of phase with the TC clock supplied. This clock, which will have its edges at the center of the data valid eye, is used to clock in the TX[0:4] data. Setup and hold times are taken care of by the HDMP-1685A provided the specifications indicated are met.
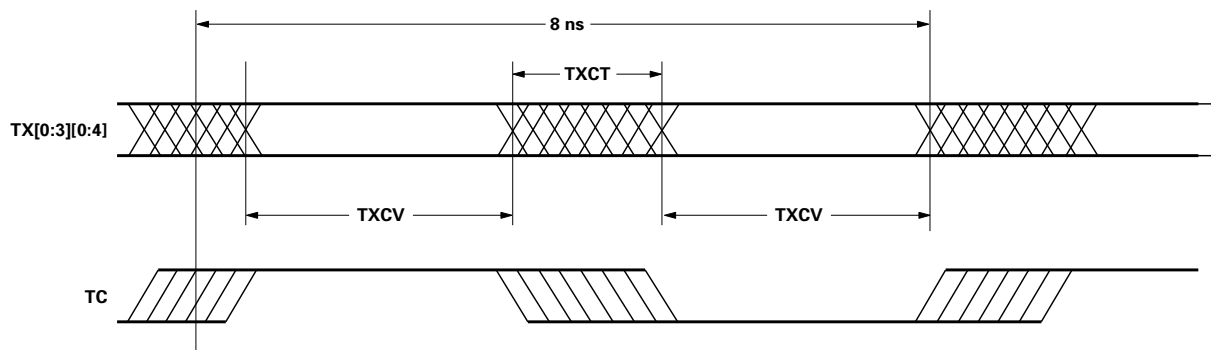
Figure 3. Transmitter section parallel input timing.
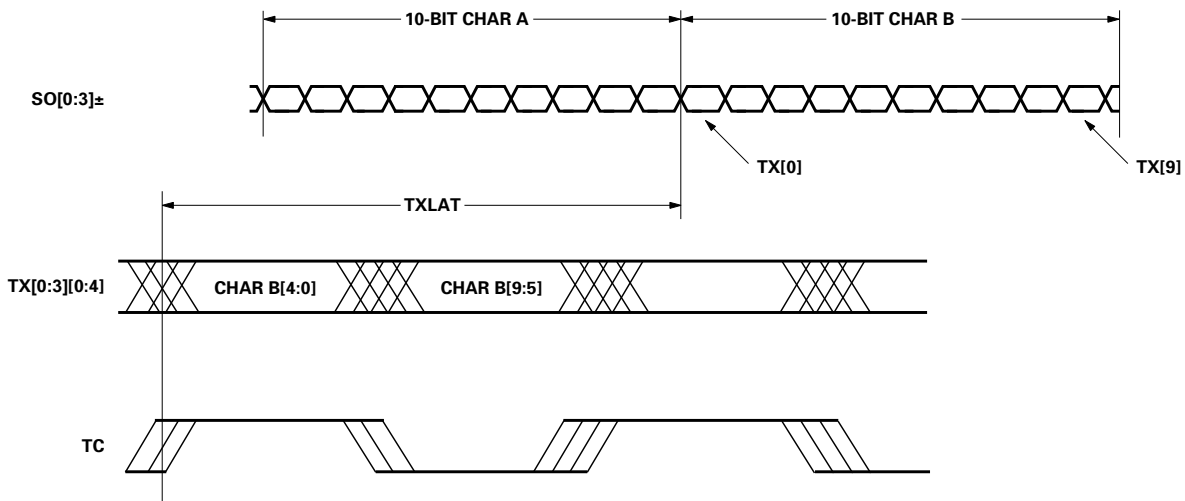
Figure 4. Transmitter section latency. TX[0] is first on serial wire.

5

HDMP-1685A Timing Characteristics – Receiver Sections – Rising Edge Clocking

$T_A = 0°C$ to $T_C = 85°C$, $V_{CC} = 3.15$ V to 3.45 V

Symbol	Parameter	Units	Min.	Typ.	Max.
f_lock	Frequency Lock at Powerup	µs			500
B_sync[1,2]	Bit Sync Time	bits			2500
t_{RXS}	RX [0:3][0:4] Setup Time (Data Valid Before Clock)	ps	1200		
t_{RXH}	RX [0:3][0:4] Hold Time (Data Valid After Clock)	ps	800		
	RC [0:3][1] to RC [0:3][0] Skew	ns	3.5		4.5
	RC [0:3][1] and RC [0:3][0] Duty Cycle	%	40		60
t_rxlat[3]	Receiver Latency	ns		16	
		bits		20	

Notes:

1. This is the recovery time for input phase jumps, per the Fibre Channel Specification X3.230-1994 FC-PH Standard, Sec 5.3.
2. Tested using $C_{PLL} = 0.1$ µF.
3. The receiver latency, as shown in Figure 6, is defined as the time between receiving the first serial bit of a parallel data word (defined as the edge of the first serial bit) and the clocking out of that parallel word (defined by the rising edge of the receive byte clock, RC [0:1]).

Figure 5a. Receiver section parallel output timing using rising edge of both RC[0:3][0] and RC[0:3][1].

6

HDMP-1685A Timing Characteristics – Receiver Sections – Rising and Falling Edge Clocking

$T_A = 0°C$ to $T_C = 85°C$, $V_{CC} = 3.15$ V to 3.45 V

Symbol	Parameter	Units	Min.	Typ.	Max.
f_lock	Frequency Lock at Powerup	µs			500
B_sync[1,2]	Bit Sync Time	bits			2500
t_{RXS}	RX [0:3][0:4] Setup Time (Data Valid Before Clock)	ps	1000		
t_{RXH}	RX [0:3][0:4] Hold Time (Data Valid After Clock)	ps	800		
	RC [0:3][1] and RC [0:3][0] Duty Cycle	%	40		60
t_rxlat[3]	Receiver Latency	ns		16	
		bits		20	

Notes:

1. This is the recovery time for input phase jumps, per the Fibre Channel Specification X3.230-1994 FC-PH Standard, Sec 5.3.
2. Tested using $C_{PLL} = 0.1$ µF.
3. The receiver latency, as shown in Figure 6, is defined as the time between receiving the first serial bit of a parallel data word (defined as the edge of the first serial bit) and the clocking out of that parallel word (defined by the rising edge of the receive byte clock, RC [0:1]).
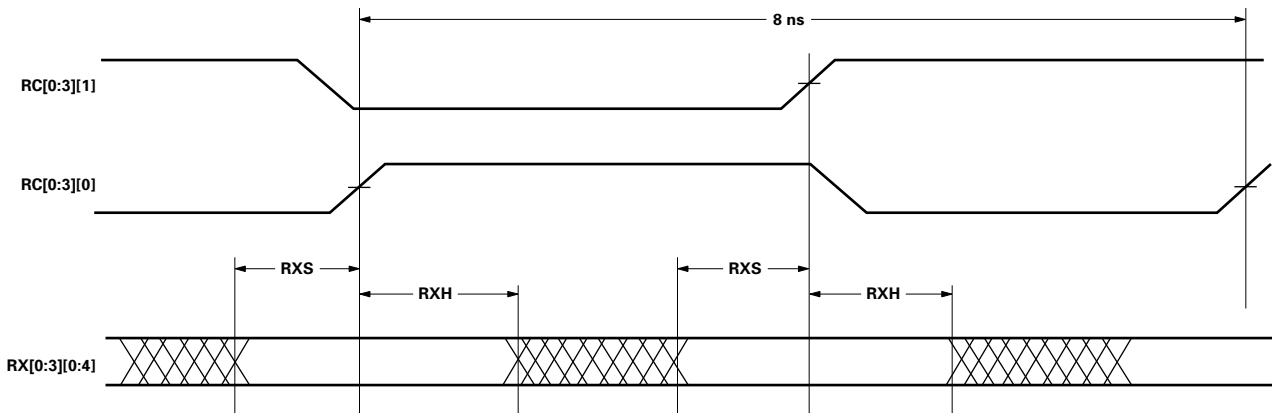

Figure 5b. Receiver section parallel output timing using rising and falling edge of either RC[0:3][0] or RC[0:3][1].
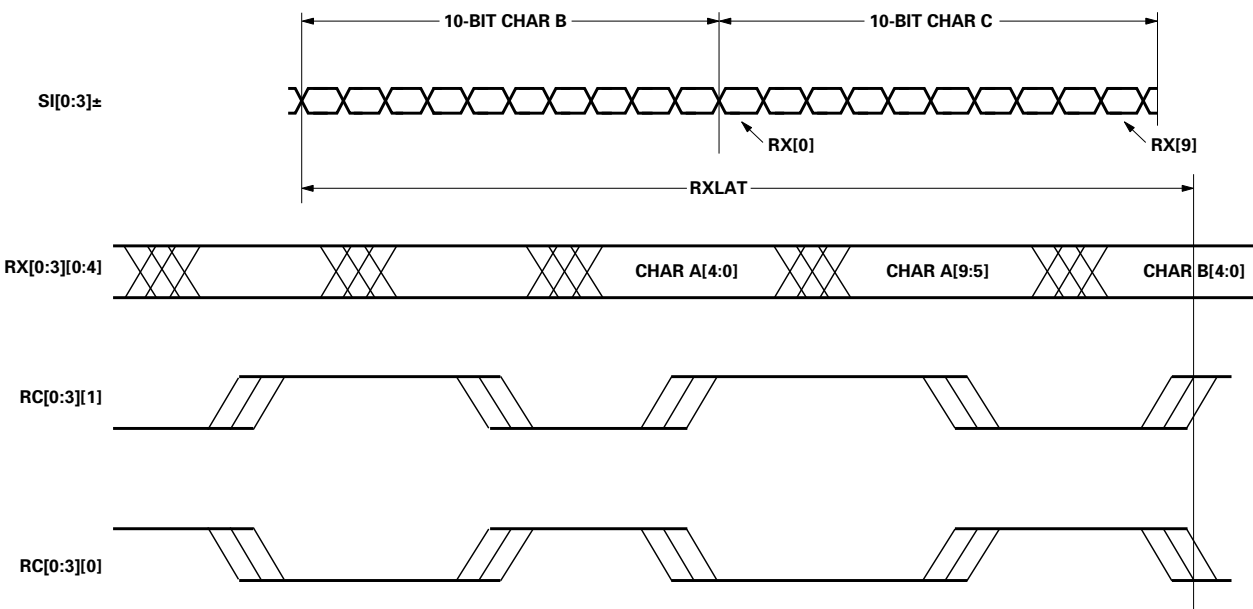
Figure 6. Receiver section latency. First bit on serial wire drives RX[0:3][0].

7

HDMP-1685A Absolute Maximum Ratings

$T_A = 25°C$, except as specified. Sustained operation at or beyond any of these conditions may result in long-term reliability degradation or permanent damage, and is not recommended.

Symbol	Parameter	Units	Min.	Max.
V_{CC}	Supply Voltage	V	−0.5	5.0
$V_{IN,LVTTL}$	RFCT LVTTL Input Voltage	V	−0.7	V_{CC} + 2.8
$V_{IN,SSTL}$	SSTL Input Voltage	V	−0.7	V_{CC} + 0.7
V_{IN,HS_IN}	HS_IN Input Voltage (Differential)	V		2.2
T_{stg}	Storage Temperature	°C	−65	+150
T_j	Junction Temperature	°C	0	+125
T_C	Case Temperature	°C	0	95

HDMP-1685A Guaranteed Operating Rates

$T_A = 0°C$ to $T_C = 85°C$, $V_{CC} = 3.15$ V to 3.45 V

Parallel Clock Rate (MHz)		Serial Baud Rate (MBaud)	
Min.	Max.	Min.	Max.
124.0	126.0	1240	1260

HDMP-1685A Reference Clock and Transmit Byte Clock Requirements

$T_A = 0°C$ to $T_C = 85°C$, $V_{CC} = 3.15$ V to 3.45 V

Symbol	Parameter	Units	Min.	Typ.	Max.
f	Nominal Frequency	MHz		125	
F_{tol}	Frequency Tolerance	ppm	−100		+100
$Symm_{RFC}$	Symmetry (Duty Cycle) Reference Clock	%	40		60
$Symm_{TC}$	Symmetry (Duty Cycle) Transmit Byte Clock	%	40		60

HDMP-1685A LVTTL I/O DC Electrical Specifications

$T_A = 0°C$ to $T_C = 85°C$, $V_{CC} = 3.15$ V to 3.45 V

Symbol	Parameter	Units	Min.	Typ.	Max.
$V_{IH,LVTTL}$	LVTTL Input High Voltage Level, Guaranteed High Signal for All Inputs	V	2		5.5
$V_{IL,LVTTL}$	LVTTL Input Low Voltage Level, Guaranteed Low Signal for All Inputs	V	0		0.8
$V_{OH,LVTTL}$	LVTTL Output High Voltage Level, $I_{OH} = −400$ µA	V	2.2		V_{CC}
$V_{OL,LVTTL}$	LVTTL Output Low Voltage Level, $I_{OL} = 1$ mA	V	0		0.5
$I_{IH,LVTTL}$	Input High Current, $V_{IN} = 2.4$ V, $V_{CC} = 3.45$ V	µA			40
$I_{IL,LVTTL}$	Input Low Current, $V_{IN} = 0.4$ V, $V_{CC} = 3.45$ V	µA			−600

HDMP-1685A SSTL_2 I/O DC Electrical Parameters

T_A = 0°C to T_C = 85°C, V_{CC} = 3.15 V to 3.45 V, VDDQ = 2.30 V to 2.70 V. VDDQ is the FC-1/MAC device I/O supply voltage. SSTL-2 inputs can receive LVTTL signals successfully. SSTL-2 outputs do not output LVTTL compliant levels.

Symbol	Parameter	Units	Min.	Typ.	Max.
VREFT	SSTL_2 Input Reference Voltage	V	1.15	1.25	1.35
V_{IH}	Input High Voltage	V	VREFT +0.18		VDDQ +0.30
V_{IL}	Input Low Voltage	V	−0.30		VREFT −0.18
VREFR	SSTL_2 Output Reference Voltage	V	1.15	1.25	1.35
V_{OH}	Output High Voltage	V	VREFR +0.38		VDDQ
V_{OL}	Output Low Voltage	V	GND		VREFR −0.38

HDMP-1685A AC Electrical Specifications (TRx)

T_A = 0°C to T_C = 85°C, V_{CC} = 3.15 V to 3.45 V

Symbol	Parameter	Units	Min.	Typ.	Max.
$t_{r,RFCT}$	RFCT LVTTL Input Rise Time, 0.8 to 2.0 Volts	ns	0		2.4
$t_{f,RFCT}$	RFCT LVTTL Input Fall Time, 2.0 to 0.8 Volts	ns	0		2.4
$t_{r,SSTLo}$	SSTL Output Rise Time, 1.0 V to 1.6 V	ns		0.4	1.5
$t_{f,SSTLo}$	SSTL Output Fall Time, 1.6 V to 1.0 V	ns		0.28	1.5
t_{rs, HS_OUT}	HS_OUT Single-Ended (SO[0:3]±) Rise Time (20% - 80%)	ps	85	205	300
t_{fs, HS_OUT}	HS_OUT Single-Ended (SO[0:3]±) Fall Time (20% - 80%)	ps	85	180	300
t_{rd, HS_OUT}	HS_OUT Differential Rise Time	ps	85		300
t_{fd, HS_OUT}	HS_OUT Differential Fall Time	ps	85		300
V_{IP,HS_IN}	HS_IN (SI[0:3]±) Input Peak-To-Peak Differential Voltage	mV	200	1200	2000
V_{OP,HS_OUT}[1]	HS_OUT Output Pk-Pk Diff. Voltage (Z0=50 Ohms, Fig.10)	mV	1000	1300	1800

Note:

1. Output Peak-to-Peak Differential Voltage specified as SO[0:3]+ minus SO[0:3]−. The amplitude will be 25% higher when terminating into 75 Ω loads.
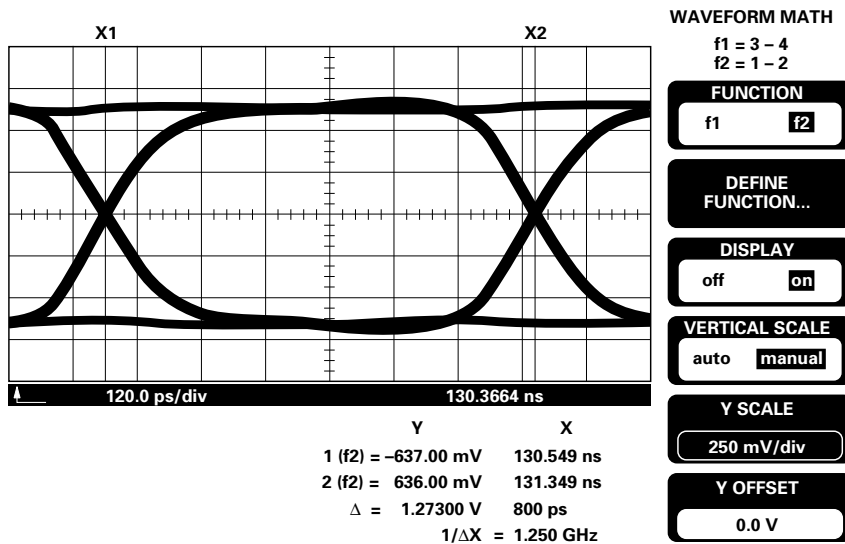
Figure 7. Eye diagram of a high speed differential output.

HDMP-1685A Output Jitter Characteristics – Transmitter Section

$T_A = 0°C$ to $T_C = 85°C$, $V_{CC} = 3.15$ V to 3.45 V

Symbol	Parameter	Units	Typ.
RJ[1]	Random Jitter at DOUT, the High Speed Electrical Data Port, specified as 1 sigma deviation of the 50% crossing point (RMS)	ps	11
DJ[1]	Deterministic Jitter at DOUT, the High Speed Electrical Data Port (pk-pk)	ps	26

Note:

1. Defined by Fibre Channel Specification X3.230-1994 FC-PH Standard, Annex A, Section A.4 and tested using measurement method shown in Figure 8.

A. BLOCK DIAGRAM OF RJ MEASUREMENT METHOD

B. BLOCK DIAGRAM OF DJ MEASUREMENT METHOD
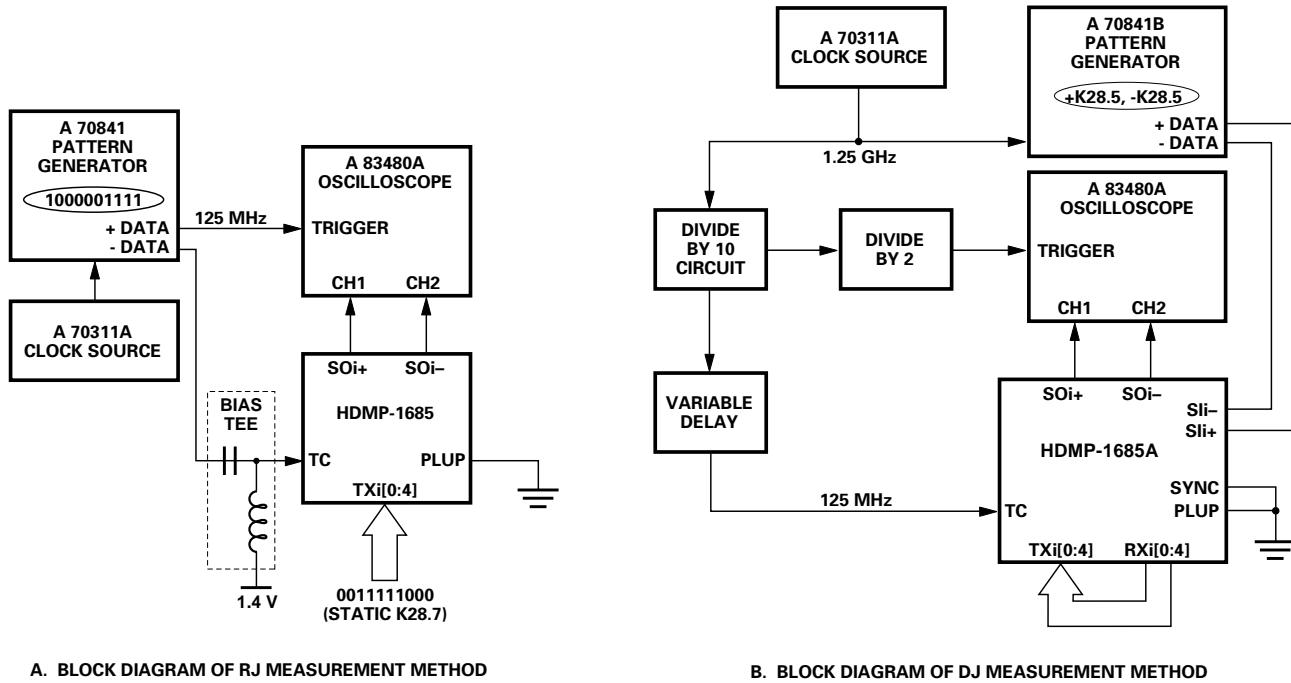
Figure 8. Transmitter jitter measurement method.

HDMP-1685A Thermal and Power Temperature Characteristics (TRx)

$T_A = 0°C$ to $T_C = 85°C$, $V_{CC} = 3.15$ V to 3.45 V

Symbol	Parameter	Units	Typ.	Max.
$I_{CC,TRx}$	Transceiver V_{CC} Supply Current, $T_A = 25°C$	mA	900	
$P_{D, TRx}$[1]	Transceiver Power Dissipation, Outputs Connected per Recommended Bias Terminations with 2^7-1 PRBS Pattern	W	2.97	3.5
Θ_{jc}[1]	Thermal Resistance, Junction to Case	°C/W	7.8	
T_j	Junction Temperature (absolute maximum)	°C	0	+125
T_C	Case Temperature (absolute maximum)	°C	0	+95

Note:

1. Based on independent package testing by Agilent. Θ_{ja} for these devices is 36°C/W for the HDMP-1685A. Θ_{ja} is measured on a standard 3x3" FR4 PCB in a still air environment. To determine the actual junction temperature in a given application, use the following: $T_j = T_C + (\Theta_{jc} \times P_D)$, where Tc is the case temperature measured on the top center of the package and P_D is the power being dissipated.

HDMP-1685A I/O Type Definitions

I/O Type	Definition
I-LVTTL	Input LVTTL, floats high when left open
I-SSTL2	Input SSTL_2, floats low when left open
O-SSTL2	Output SSTL_2
HS_OUT	50 Ω matched Output Driver. Will drive AC coupled 50 Ω loads. PECL level compatible (Figure 10).
HS_IN	PECL level compatible. Must be AC coupled (Figure 10).
C	External Circuit Node
S	Power Supply or Ground

HDMP-1685A Pin Input Capacitance (TRx)

Symbol	Parameter	Units	Typ.	Max.
C_{INPUT}	Input Capacitance on SSTL input pins	pF	1.6	
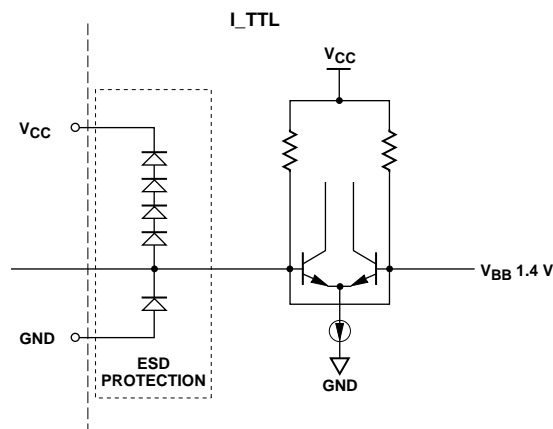
Figure 9. LVTTL input simplified circuit schematic (for RFCT).
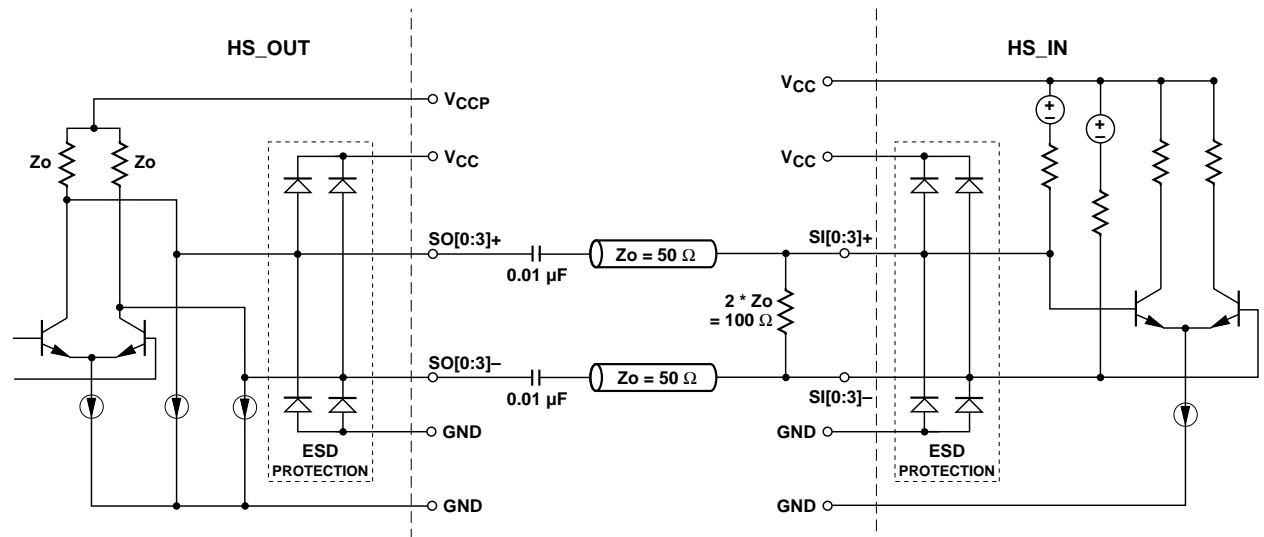
NOTES:
1. HS_IN INPUTS SHOULD NEVER BE CONNECTED TO GROUND AS PERMANENT DAMAGE TO THE DEVICE MAY RESULT.
2. CAPACITORS MAY BE PLACED AT THE SENDING END OR THE RECEIVING END.

Figure 10. HS_OUT and HS_IN simplified circuit schematic.

11

V_{CC} (SERDES) = 3.3 V V_{CC} (SERDES) = 3.3 V V_{CC} (MAC)

V_{DDQ} (MAC)

HDMP-1685

- V_{CC}
- VREFR
- RX[0:3][0:4] / RC[0:3][0:1]
- TX[0:3][0:4]
- TC
- VREFT

0.1 µF R_1 R_2

RS = 50 Ω UNTERMINATED

RS = 50 Ω USE TERMINATION, IF NECESSARY, TO DELIVER PROPER VOLTAGE SWINGS AT TX[0:4]

V_{DDQ} (MAC) R_1 R_2 0.1 µF

MAC

- V_{CC}
- VDDQ
- VREFT
- DATAIN
- DATAOUT
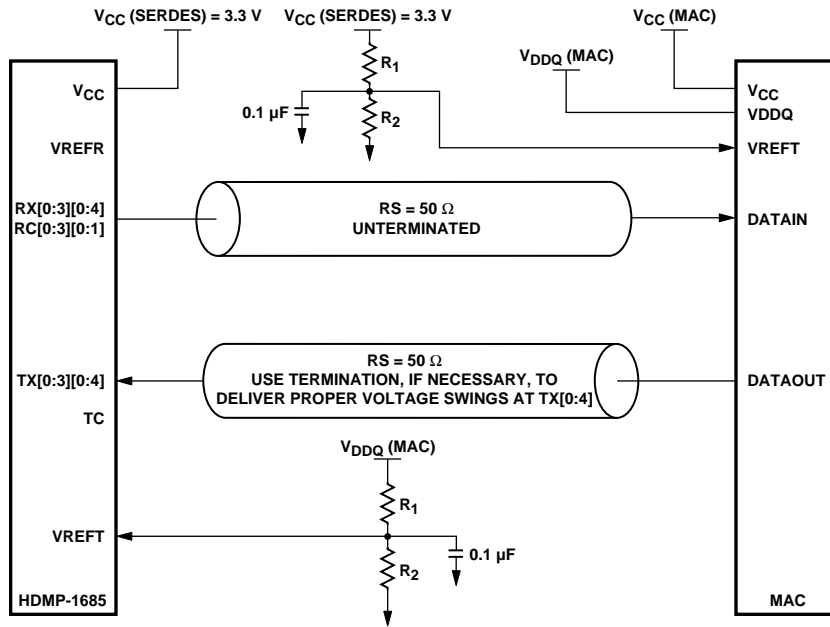
NOTE: VREFR ON EACH DEVICE MAY BE USED TO DRIVE VREFT ON THE OTHER DEVICE INSTEAD OF USING THE CONFIGURATION ABOVE. VREFR SHOULD BE BYPASSED WITH 0.1 µF IN THIS CASE. IF USED, R_1 AND R_2 SHOULD BE 500-1000 Ω. 1% RESISTORS SHOULD BE USED FOR R_1 AND R_2. WHEN USING THE CONFIGURATION ABOVE, VREFT TO THE MAC SHOULD BE SET TO 1.25 V NOMINAL. USING THIS VALUE CENTERS VREFR RELATIVE TO THE RX[0:3][0:4] OUTPUT SWINGS PROVIDED BY THE HDMP-1685A.

Figure 11. O-SSTL_2 and I-SSTL_2 simplified circuit schematic.

	01	02	03	04	05	06	07	08	09	10	11	12	13	14	15	16	17
A	GND	VCC	GND	SYN1	RC10	RX10	RX14		VCR1	VCC	GND	RX21	VCR2		GND	VCR2	GND
B			GND	RC11	RX11	VCR1		GND	SYN2	RX20	RX22				SYN3	RC30	RC31
C	RX04			VCR0	VCR1	RX12	GND			RC20	VCR2	RX23		VCC	VCR3	GND	RX30
D	RX00	RX01	VCR0	GND	GND	RX13	VRFR			RC21	GND	RX24		RX31	RX32	RX33	RX34
E	RC00	RC01	RX02	RX03										VCR3	GND		
F		SYN0	VCR0	GND													GND
G			GND	VCC				GND						TX22	TX21	TX20	VCR3
H	TX14							GNDA								TX24	TX23
J	TX10	TX11	TX12	TX13				VCCP						VCC			
K			GND	GND				VCC						GND	GND	VCC	TC
L			GND	VCC				VCCA						TX33	TX32	TX31	TX30
M	TX04							VCR									TX34
N	TX00	TX01	TX02	TX03										PLUP	VCC		
P	VRFT			GND	SO0−	GND	SO1−	GND	CAP0	GND	SO2+	GND	SO3+				
R	RFCT		GND	VCC	SO0+	GND	SO1+	GNDA	CAP1	GND	SO2−	GND	SO3−	VCC	GND	VCC	SYNC
T		GND	VCC	VCC	VCP0	GND	VCP1	GND	VCCA	GND	VCP2	GND	VCP3	VCC	GND	GND	
U	GND	GND	SIO−	SI0+	VCC	SI1−	SI1+	GND	GND	SI2−	SI2+	GND	SI3−	SI3+	GND	GND	GND

Figure 12. Pinout of HDMP-1685A (top view).

12

Filtering Schematic

HDMP-1685A
GUIDELINES FOR
DECOUPLING CAPACITOR
PLACEMENTS/CONNECTIONS

V_CC 10 μF *

0.1 μF (A2, V_CC)
0.1 μF (A10, V_CC)
0.1 μF (A13, A16)

C4 VCR
C5 VCR
D3 VCR
F3 VCR
G4 V_CC
0.1 μF

L4 V_CC
R4 V_CC
0.1 μF

C11 VCR
C14 V_CC
C15 VCR
E14 VCR
0.1 μF

G17 VCR
0.1 μF

J14 V_CC
K16 V_CC
0.1 μF

N15 V_CC
0.1 μF

R16 V_CC

B7 VCR
A9 VCR
A10 V_CC
A13 VCR
A16 VCR

T3 V_CC
T4 V_CC
T5 VCP
U5 V_CC
T7 VCP
T9 V_CCA
T11 VCP
T13 VCP
T14 V_CC
R14 V_CC

0.1 μF   0.1 μF   0.1 μF

V_CC
1 μH
0.1 μF   + 10 μF   10 μF   +   0.1 μF

* PLACEMENT NOT CRITICAL. INDICATES
  THE NEED FOR ADDITIONAL LOW FREQUENCY
  CAPACITIVE DECOUPLING.

⊞ OPTIONAL – PROVIDES INCREASED LOW
  FREQUENCY DECOUPLING.

	01	02	03	04	05	06	07	08	09	10	11	12	13	14	15	16	17
	0.1 μF						0.1 μF		10 μF*						0.1 μF		
A	GND	V_CC	GND						VCR	V_CC	GND		VCR		GND	VCR	GND
B			GND				VCR		GND								
C				VCR	VCR		GND				VCR			V_CC	VCR	GND	
D			VCR	GND	GND						GND						
E					0.1 μF									VCR	GND		
F			VCR	GND									0.1 μF				GND
G			GND	V_CC												VCR	
H																	
J														V_CC			
K			GND	GND									0.1 μF	GND	GND	V_CC	
L			GND	V_CC													
M																	
N					0.1 μF									V_CC			
P				GND		GND		GND		GND		GND					
R			GND	V_CC		GND		GND		GND		GND		V_CC	GND	V_CC	
T		GND	V_CC	V_CC	VCP	GND	VCP	GND	V_CCA	GND	VCP	GND	VCP	V_CC	GND	GND	
U	GND	GND			V_CC			GND	GND			GND			GND	GND	GND

* 10 μF PLACEMENT NOT CRITICAL – INDICATES
  NEED FOR LOW-FREQUENCY BYPASS CAPACITANCE

HDMP-1685A

0.1 μF (right side, G)
0.1 μF (right side, P)

0.1 μF   0.1 μF   0.1 μF   0.1 μF

10 μF   → TO V_CCA PI-FILTER (SEE SCHEMATIC)

13

HDMP-1685A TRx I/O Definition

Name	Pin	Type	Signal
CAP0	P09	C	**Loop Filter Capacitor:** A loop filter capacitor for the internal PLLs must be connected across the CAP0 and CAP1 pins. (typical value = 0.1 μF)
CAP1	R09		
PLUP	N14	I-SSTL2	**Parallel Loopback Enable Input:** When set high, a high-speed serial signal from the transmitter section's serial output select block, reflecting TX data, is driven to the receiver section's serial input select block. RX data reflects this serial signal. Also when in parallel loopback mode, the SO [0:3]+/- outputs are held static at logic 1.
RFCT	R01	I-LVTTL	**LVTTL Reference Clock:** RFCT is a 125 MHz clock signal supplied to the IC.
RC00	E01	O-SSTL2	**Receiver Byte Clocks:** The receiver sections drive 125 MHz receive byte clocks RC [0:3] [0:1].
RC01	E02		
RC10	A05		
RC11	B05		
RC20	C10		
RC21	D10		
RC30	B16		
RC31	B17		
RX00	D01	O-SSTL2	**Data Outputs:** Four 5-pin data busses. RX [0:3] [0] are the first bits received.
RX01	D02		
RX02	E03		
RX03	E04		
RX04	C01		
RX10	A06		
RX11	B06		
RX12	C06		
RX13	D06		
RX14	A07		
RX20	B11		
RX21	A12		
RX22	B12		
RX23	C12		
RX24	D12		
RX30	C17		
RX31	D14		
RX32	D15		
RX33	D16		
RX34	D17		
SI0+	U04	HS_IN	**Serial Data Inputs:** High-speed inputs. Serial data are accepted from the SI [0:3]+/- inputs except when PLUP is high.
SI0-	U03		
SI1+	U07		
SI1-	U06		
SI2+	U11		
SI2-	U10		
SI3+	U14		
SI3-	U13		

Name	Pin	Type	Signal
SO0+ SO0- SO1+ SO1- SO2+ SO2- SO3+ SO3-	R05 P05 R07 P07 P11 R11 P13 R13	HS_OUT	**Serial Data Outputs:** High-speed outputs. These lines are active except when PLUP is high, in which case these outputs are held static at logic 1.
SYNC	R17	I-SSTL2	**Enable Byte Sync Input:** When high, turns on the internal byte sync functions to allow clock synchronization to a comma character of positive disparity (0011111XXX). When the line is low, the function is disabled and will not reset registers and clocks, or strobe the SYN [0:3] lines.
SYN0 SYN1 SYN2 SYN3	F02 A04 B10 B15	O-SSTL2	**Byte Sync Outputs:** Active high outputs. Used to indicate detection of a comma character of positive disparity (0011111XXX) when SYNC is enabled.
TC	K17	I-SSTL2	**Transmit Byte Clock:** This signal is used to latch transmit data for all channels into the IC.
TX00 TX01 TX02 TX03 TX04 TX10 TX11 TX12 TX13 TX14 TX20 TX21 TX22 TX23 TX24 TX30 TX31 TX32 TX33 TX34	N01 N02 N03 N04 M01 J01 J02 J03 J04 H01 G16 G15 G14 H17 H16 L17 L16 L15 L14 M17	I-SSTL2	**Data Inputs:** Four 5-pin data busses. TX [0:3] [0] are the first bits transmitted.
VREFT	P01	I-S	**TX Parallel Interface SSTL_2 Reference Voltage:** Voltage reference derived from 2 resistor network with V_{DDQ} (ASIC) as supply, as recommended in Figure 11.
VREFR	D07	O-S	**RX Parallel Interface SSTL_2 Reference Voltage:** Provided by HDMP-1685A. Drives the VREF input of the ASIC.

HDMP-1685A TRx I/O Definition, continued

Name	Pin	Type	Signal
V_CC	A02 A10 C14 G04 J14 K16 L04 N15 R04 R14 R16 T03 T04 T14 U05	S	**Power Supply:** Normally 3.3 volts. Used for logic, SSTL inputs, and LVTTL I/O.
VCCA	T09	S	**Analog Power Supply:** Normally 3.3 volts. Used to provide a clean supply line for the PLLs and high-speed analog cells.
VCR0 VCR1 VCR2 VCR3	C04 D03 F03 A09 B07 C05 A13 A16 C11 C15 E14 G17	S	**Rx SSTL2 Output Power Supply:** Normally 3.3 volts. Used for all SSTL2 receiver output buffer cells.
VCP0 VCP1 VCP2 VCP3	T05 T07 T11 T13	S	**High-Speed Output Supply:** Normally 3.3 volts. Used only for the last stage of the high-speed transmitter output cells (HS_OUT) as shown in Figure 10. Due to high current transitions, this V_{CC} should be well bypassed to a ground plane.
GNDA	R08	S	**Analog Ground:** Normally 0 volts. All GND pads on the chip are connected to one ground slug in the package, which then distributes these to GND balls.

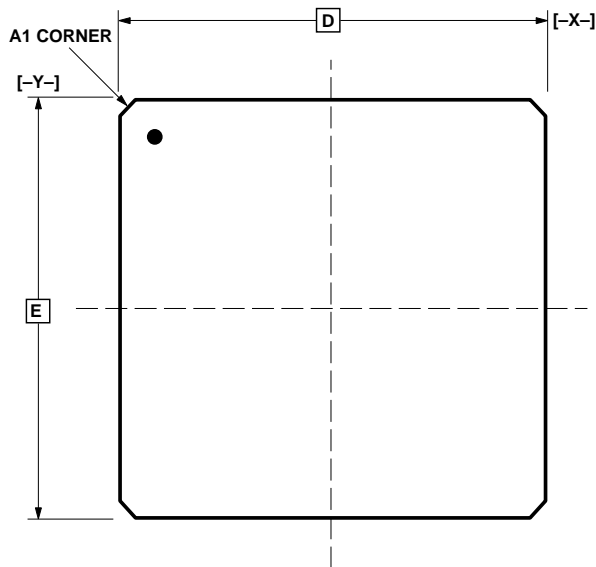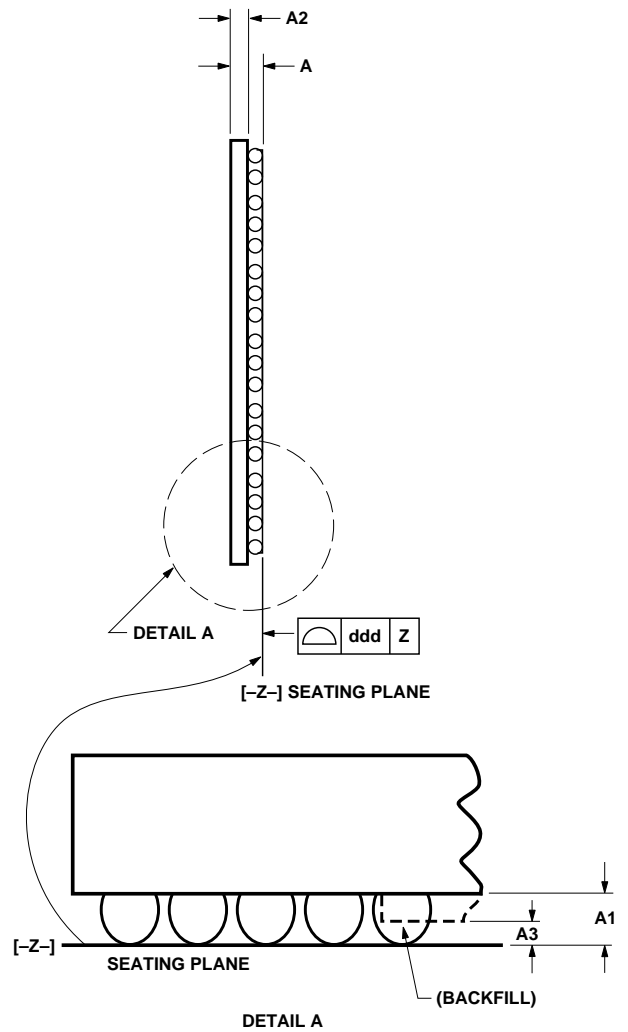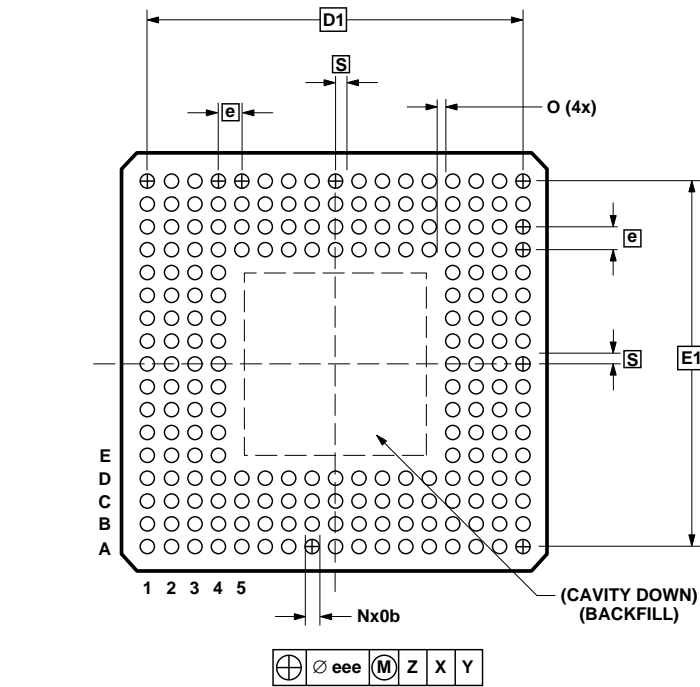| Name	Pin	Type	Signal
GND	A01	S	**Logic Ground:** Normally 0 volts. All GND pads on the chip are connected to one ground slug in the package, which then distributes these to GND balls.
	A03		
	A11		
	A15		
	A17		
	B04		
	B09		
	C07		
	C16		
	D04		
	D05		
	D11		
	E15		
	F04		
	F17		
	G03		
	K03		
	K04		
	K14		
	K15		
	L03		
	P04		
	P06		
	P08		
	P10		
	P12		
	R03		
	R06		
	R10		
	R12		
	R15		
	T02		
	T06		
	T08		
	T10		
	T12		
	T15		
	T16		
	U01		
	U02		
	U08		
	U09		
	U12		
	U15		
	U16		
	U17		

HDMP-1685A TRx I/O Definition, continued

Name	Pin	Type	Signal
NC	A08		These pins are connected to an isolated pad and have no functionality.
	A14		They may be left open, or LVTTL levels may be applied.
	B01		
	B02		
	B03		
	B08		
	B13		
	B14		
	C02		
	C03		
	C08		
	C09		
	C13		
	D08		
	D09		
	D13		
	E16		
	E17		
	F01		
	F14		
	F15		
	F16		
	G01		
	G02		
	H02		
	H03		
	H04		
	H14		
	H15		
	J15		
	J16		
	J17		
	K01		
	K02		
	L01		
	L02		
	M02		
	M03		
	M04		
	M14		
	M15		
	M16		
	N16		
	N17		
	P02		
	P03		
	P14		
	P15		
	P16		
	P17		
	R02		
	T01		
	T17		

Package Drawing

D1

S

e

O (4x)

A2

A

e

S

E1

E
D
C
B
A

1 2 3 4 5

Nx0b

(CAVITY DOWN)
(BACKFILL)

| ⊕ | ⌀ eee | Ⓜ | Z | X | Y |

DETAIL A

ddd | Z

[–Z–] SEATING PLANE

A1 CORNER

D

[–X–]

[–Y–]

E

[–Z–]

SEATING PLANE

A3

A1

(BACKFILL)

DETAIL A

DIMENSIONS IN MILLIMETERS

SYMBOL	MIN.	NOM.	MAX.
A	1.35	1.50	1.65
A1	0.60	0.65	0.70
A2	0.85	0.90	0.95
A3	0.15		
D		23.00 ± 0.20	
D1		20.32 BSC	
E		23.00 ± 0.20	
E1		20.32 BSC	
MD/ME		17	
N		208	
N1		4	
O	0.60		
b	0.60	0.75	0.90
e		1.27 ± 0.10	

TOLERANCE OF FORM AND POSITION

SYMBOL	MIN.	NOM.	MAX.
ddd			0.15
eee			0.30

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