

## **AP8804** LOW VOLTAGE PULSED LED DRIVER

## Description

The AP8804 is a low voltage linear LED driver capable of operating from supplies as low as 1.8V.

The AP8804 has an output current tolerance of 10% at 25°C at 250mA and 15% at the full rated current of 350mA.

The maximum output saturation voltage at the full rated current is 200mV.

## **Features**

- 100mV Maximum output saturation voltage •
- TDFN small outline package •
- 10% Output current tolerance
- 0 to 70°C Junction temperature range
- Available in "Green" Molding Compound • (No Br, Sb) with lead free Finish/RoHS Compliant (Note1)

# **Typical Application Circuit**



1. EU Directive 2002/95/EC (RoHS). All applicable RoHS exemptions applied, see EU Directive 2002/95/EC Annex Notes. Notes:



**Pin Assignments** 

**Top View** 

AP8804 Document number: DS32050 Rev. 2 - 5





## **Absolute Maximum Ratings**

Symbol	Parameter	Rating	Unit	
Vcc	Supply Voltage	-0.3 to +8	V	
V <sub>OUT</sub>	Output Voltage	(Relative to GND)	v v	
V <sub>BIAS</sub>	Bias Input Voltage	-0.3 to +0.3 (Relative to GND)	V	
TJ	Junction Temperature	150	C°	
T <sub>ST</sub>	Storage Temperature	-65 to +150	C°	

These are stress ratings only. Operation outside the absolute maximum ratings may cause device failure. Operation at the absolute maximum rating for Note: extended periods may reduce device reliability.

Semiconductor devices are ESD sensitive and may be damaged by exposure to ESD events. Suitable ESD precautions should be taken when handling and transporting these devices.

## **Recommended Operating Conditions**

Symbol	Parameter	Min	Max	Unit
V <sub>CC</sub>	Supply Voltage	1.8	6	V
I <sub>BIAS</sub>	BIAS Input Current	-0.33	0	mA
Ι <sub>Ο</sub>	Output Current	50	350	mA
T <sub>A</sub>	Operating Ambient Temperature Range	0	70	°C

## Package Thermal Data

Thermal Resistance	Package		Unit
Junction-to-Ambient, $\theta_{JA}$	TDFN2020-6	80	°C 11/
Junction-to-Case, $\theta_{JC}$	TDFN2020-6	20	C/VV

## Electrical Characteristics (Under Operating Conditions)

Parameter	Conditions		TA	Min	Тур.	Max	Unit
DC parameters	·						
Pige voltage	$V_{CC}$ = 1.8V, $R_{BIAS}$ = 3.05k $\Omega^2$		25	900	1040	1200	mV
bias voltage			FT	860		1240	
Output current accuracy <sup>3</sup>	$V_{CC} = 2.0V$ to 6V, V	$V_{OUT} = V_{CC} - 1.8V, R_{BIAS} = 4.1k\Omega$	25	225	250	275	mA
	$V_{CC} = 2.0V, V_{OUT} =$	= 0.2V R <sub>BIAS</sub> = 3.05kΩ, I <sub>OUT</sub> =	25			15	0/
	350mA		FT			20	70
Saturated_output current	$V_{aa} = 2.0V_{ab} V_{ab} = -$	$V_{CC} = 2.0V, V_{OUT} = 0.5V R_{BIAS} = 3.05k\Omega I_{OUT} = 350mA$				10	0/2
tolerance <sup>5</sup>	VCC - 2.0V, VOUT -					15	70
	$V_{CC}$ = 2.0V, $V_{OUT}$ = 0.15V $R_{BIAS}$ = 4.1k $\Omega$ , $I_{OUT}$ = 250mA		25			15	%
			FT			20	
Current Gain $(l_{a}, -l_{a})^{3,4}$	$V_{CC} = 2V, V_{OUT} = 0.2V,$ $50\mu A < I_{BIAS} < 350\mu A$		25		1000		
Current Gain (IOUT/IBIAS)			25		1000		
Quiescent current	V <sub>CC</sub> = 6V, R <sub>BIAS</sub> = Open Circuit		FT		100	150	μA
Supply current (Active)	$V_{CC} = 2V, R_{BIAS} = 3.05 k\Omega, V_{OUT} = 0.5 V$		FT		5.5	9	mA
Output Leakage (Off)	$Vcc = 6V, R_{BIAS} \ge 2M\Omega, V_{OUT} = 6V$		25		<1	10	uA
Output Leakage (Disabled)	$Vcc = 0V, V_{OUT} = 6V$		25		<1	100	nA
AC parameters	•						
Rise time (10% to 90%) <sup>5</sup>	$V_{CC} = 2.0V,$	$R_{BIAS} = O/C$ to $3.05k\Omega$			0.2		
Fall time (90% to 10%) <sup>5</sup>	$V_{OUT} = 0.5V,$	$R_{BIAS} = 3.05 k\Omega$ to O/C			0.2		1
Propagation delay(low to high) <sup>5</sup>	$V_{\rm CC} = 2.0 V,$	$R_{BIAS} = O/C$ to $3.05k\Omega$			0.4		μs
Propagation delay(high to low) <sup>5</sup>	$V_{OUT} = 0.5V,$	$R_{BIAS} = 3.05 k\Omega$ to O/C			0.2		1

2.  $R_{\text{BIAS}}$  connected to GND see Figure 1

3. All active output current measurements are pulsed tests of 8us duration, 10% duty cycle

4. See Typical Characteristics for a graph of Output current vs. R<sub>BIAS</sub> value

5. See Figure 2

Notes:





## Figure 1: Application Circuit



## Figure 2: Timing Waveforms







## **Typical Characteristics**









# **Typical Characteristics (Cont.)**





## **PIN Descriptions**

NAME	I/O	Pin #	FUNCTION	
NC	-	1	No connection. May be grounded or left floating.	
GND	Power	2	ND or the most negative terminal of the LED Driver. Also carries the LED return current.	
BIAS	I	3	The output of an internal voltage reference (1.0V approx.). Connect a resistor $R_{BIAS}$ from his pin to the external control input. The LED Driver amplifies the $R_{BIAS}$ resistor current by a factor of nominally1000. Open circuiting this pin or driving this pin to >1.25V will urn off the output of the AP8804. Avoid unnecessary stray capacitance at this pin, and connect $R_{BIAS}$ as close as possible to it.	
V <sub>CC</sub>	Power	4	The supply input pin of the LED Driver. This pin supplies the internal circuitry of the IC which is typically 80uA + 2.5% of the LED output current. It is advisable to decouple this pin with a 100nF capacitor to GND.	
NC	-	5	No connection. May be grounded or left floating.	
OUT	0	6	The output of the LED Driver, normally connected to the cathode of the LED. The normal working range of this pin is 0.1V (0.1V above GND pin potential) to 6V. It is permissible for the voltage on this pin to be higher than $V_{CC}$ .	



## **Applications Information**

#### Setting the Output Current

The AP8804 is controlled by an input current at the BIAS pin. When an LED or load resistance is connected between a positive supply and OUT, the load current  $I_{OUT}$  is proportional to the input current at the BIAS input of the AP8804. The current gain, A<sub>I</sub> is nominally -1000. The voltage at the BIAS pin is controlled to nominally +1.0V. The resistor R<sub>BIAS</sub> sets the input current according to the input voltage, V<sub>IN</sub>, as in Figure 3.

#### Figure 3



Therefore the input voltage, V<sub>IN</sub>, determines the output current according to this equation:

 $I_{OUT} = A_I * (V_{IN} - V_{BIAS}) / R_{BIAS}$  if  $V_{IN} \le V_{BIAS}$ 

and  $I_{OUT}~=~0~$  if  $V_{IN} \geq V_{BIAS}$ 

Note that these equations are approximate and a more accurate value of current can be found using the graph of  $I_{OUT}$  vs  $R_{BIAS}$ . Select  $R_{BIAS}$  to give the required  $I_{OUT}$ .

For example, if  $R_{BIAS} = 3k$ , an input of 0V will give an output current of nominally 350mA, while an input of about +1.1V or greater will give an output current equal to the leakage value which is typically less than 1µA.

#### Driving the AP8804

The input current is approximately  $1/R_{BIAS}$  when the driven end of  $R_{BIAS}$  is pulled to 0V or roughly output current divided by 1000. For a typical output current of up to about 300 mA, the input current will be  $300\mu$ A or less. Hence a low level CMOS or open-drain logic drive level is convenient to turn the LED on (logic LOW, 0V) and off (logic HIGH, 1V or greater).

By connecting a switch to ground in series with R<sub>BIAS</sub> (see figure 3) the device can be turned on and off. Typical applications for this include LED dimming as well as data interface applications.

An alternate way of turning the output of the AP8804 off is to drive the bias pin to greater than 1.1V. However direct drive to this pin is not recommended. Stray capacitance at the BIAS pin must be avoided for stability and the resistor R<sub>BIAS</sub> should be placed close to the BIAS pin. The best method is to drive the input through the series resistor R<sub>BIAS</sub> as shown from a CMOS or open drain logic signal.

Vcc pin quiescent current is typically 100uA and OUT pin leakage is typically less than 1uA when the device is turned off using the BIAS pin. BIAS pin controlled output switching times are less than 1us.

A further means of disabling the device is to turn off the supply to the Vcc pin. In this state, assuming the LED anode is still connected to the supply, the Vcc pin current is of course zero and the OUT pin leakage is typically less than 1nA. This method has the advantage of reducing the quiescent current from about 100µA to less than 1nA. When using this switching method, the time to settle to the desired current level may be several microseconds.





#### PCB Layout and Supply Conditioning

For a satisfactory pulse shape and AC stability, attention should be given to power supply decoupling. Connect a  $0.1\mu$ F to  $1\mu$ F X7R ceramic capacitor from Vcc to the common ground plane close to the Vcc pin. Position the capacitor close to the V<sub>CC</sub> pin and use local vias to ground the capacitor. Additional ground vias should be placed as close as possible to the central paddle connection for the sake of both the pulse performance and thermal heat-sinking. The ground plane is also required to minimize undesired common-impedance coupling between input and output.

Excessive stray capacitance at the BIAS pin affects stability. It is sufficient to place the input bias resistor close to the BIAS pin to minimize the PCB trace capacitance to ground.

The AP8804 has been designed to drive the cathode of a low voltage LED which has the anode connected to the same node as  $V_{CC}$  pin. Alternatively the anode may be connected to a higher voltage than  $V_{CC}$  within the limits given. However best LED current regulation will be obtained when the anode is connected to  $V_{CC}$ .

#### Power Dissipation and Operating Temperature

Although the performance of the AP8804 is not significantly affected by junction temperature, excessive power dissipation does affect the target output current. The typical variation of I\_Out with power dissipation is approximately +0.065%/mW. A typical remote control application uses an NEC protocol data-burst with a low impedance 3V supply and 1.6V drop (at 250mA) across the LED load. Under these conditions, the power dissipation related error in I\_Out will be approximately 3% average during the burst.

## **Ordering Information**

	Packago		7"/13" Tape and Reel		
Device	ce Package Packaging Code		Quantity	Part Number Suffix	
AP8804SNG-7	SN	DFN2020-6	3000/ Tape & Reel	-7	

## **Marking Information**



A7 = Product Type Marking Code Y = Year 0 ~ 9 W = Week: A ~ Z : 1 ~ 26 Week; a ~ z : 27 ~ 52 week; z represents 52 and 53 week

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# **Package Outline Dimensions**



DFN2020-6					
Dim	Min	Max	Тур		
Α	0.57	0.63	0.60		
A1	0	0.05	0.03		
A3			0.15		
b	0.20	0.30	0.25		
D	1.95	2.075	2.00		
D2	1.45	1.65	1.55		
e			0.65		
E	1.95	2.075	2.00		
E2	0.76	0.96	0.86		
L	0.30	0.40	0.35		
All Dimonsions in mm					

# **Suggested Pad Layout**



Dimensions	Value (in mm)	
Z	1.67	
G	0.15	
X1	0.90	
X2	0.45	
Y	0.37	
С	0.65	





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