## IRS2153(1)D(S)PbF

#### SELF-OSCILLATING HALF-BRIDGE DRIVER IC

**Product Summary** 

#### **Features**

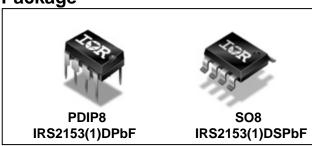
- Integrated 600 V half-bridge gate driver
- C<sub>T</sub>, R<sub>T</sub> programmable oscillator
- 15.4 V Zener clamp on V<sub>CC</sub>
- Micropower startup
- Non-latched shutdown on C<sub>T</sub> pin (1/6th V<sub>CC</sub>)
- Internal bootstrap FET
- Excellent latch immunity on all inputs and outputs
- +/- 50 V/ns dV/dt immunity
- ESD protection on all pins
- 8-lead SOIC or PDIP package
- Internal deadtime

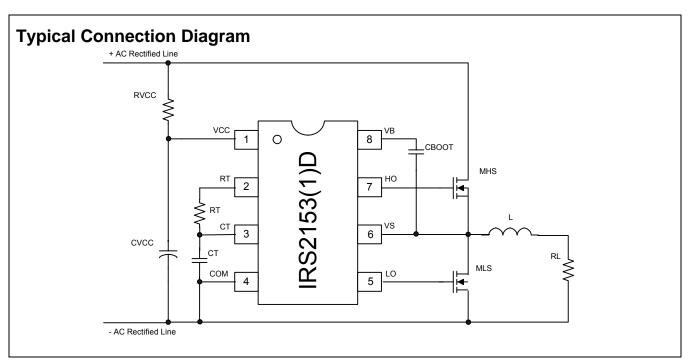
V <sub>OFFSET</sub>	600 V Max
Duty cycle	50%
Driver source/sink current	180 mA/260 mA typ.
$V_{clamp}$	15.4 V typ.
Deadtime	1.1 μs typ. (IRS2153D) 0.6 μs typ. (IRS21531D)

#### **Description**

The IRS2153(1)D is based on the popular IR2153 self-oscillating half-bridge gate driver IC using a more advanced silicon platform, and incorporates a high voltage half-bridge gate driver with a front end oscillator similar to the industry standard CMOS 555 timer. HVIC and latch immune CMOS technologies enable rugged monolithic construction. The output driver features a high pulse current buffer stage designed for minimum driver cross-conduction. Noise immunity is achieved with low di/dt peak of the gate drivers.

#### **Package**







#### **Absolute Maximum Ratings**

Absolute maximum ratings indicate sustained limits beyond which damage to the device may occur. All voltage parameters are absolute voltages referenced to COM, all currents are defined positive into any lead. The thermal resistance and power dissipation ratings are measured under board mounted and still air conditions.

	Parameter			
Symbol	Definition	Min.	Max.	Units
$V_{B}$	High side floating supply voltage	-0.3	625	
Vs	High side floating supply offset voltage	V <sub>B</sub> - 25	V <sub>B</sub> + 0.3	V
$V_{\text{HO}}$	High side floating output voltage	V <sub>S</sub> – 0.3	V <sub>B</sub> + 0.3	]
$V_{\text{LO}}$	Low side output voltage	-0.3	V <sub>CC</sub> + 0.3	
I <sub>RT</sub>	R <sub>T</sub> pin current	-5	5	mA
$V_{RT}$	R <sub>T</sub> pin voltage	-0.3	V <sub>CC</sub> + 0.3	
V <sub>CT</sub>	C <sub>T</sub> pin voltage	-0.3	V <sub>CC</sub> + 0.3	V
Icc	Supply current (Note 1)		20	
I <sub>OMAX</sub>	Maximum allowable current at LO and HO due to external power transistor Miller effect.	-500	500	mA
dV <sub>S</sub> /dt	Allowable offset voltage slew rate	-50	50	V/ns
P <sub>D</sub>	Maximum power dissipation @ T <sub>A</sub> ≤ +25 °C, 8-Pin DIP		1.0	
P <sub>D</sub>	Maximum power dissipation @ T <sub>A</sub> ≤ +25 °C, 8-Pin SOIC		0.625	W
R <sub>thJA</sub>	Thermal resistance, junction to ambient, 8-Pin DIP		85	
R <sub>thJA</sub>	Thermal resistance, junction to ambient, 8-Pin SOIC		128	°C/W
TJ	Junction temperature	-55	150	
Ts	Storage temperature	-55	150	°C
TL	Lead temperature (soldering, 10 seconds)		300	

Note 1: This IC contains a zener clamp structure between the chip  $V_{CC}$  and COM which has a nominal breakdown voltage of 15.4 V. Please note that this supply pin should not be driven by a DC, low impedance power source greater than the  $V_{CLAMP}$  specified in the Electrical Characteristics section.



#### **Recommended Operating Conditions**

For proper operation the device should be used within the recommended conditions.

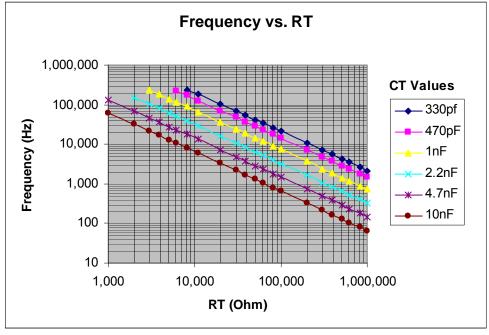
	Parameter			
Symbol	Definition	Min.	Max.	Units
V <sub>BS</sub>	High side floating supply voltage	V <sub>CC</sub> - 0.7	V <sub>CLAMP</sub>	
Vs	Steady state side floating supply offset voltage	-3.0 (Note 2)	600	V
Vcc	Supply voltage	V <sub>CCUV+</sub> +0.1 V	V <sub>CC</sub> CLAMP	
Icc	Supply current	(Note 3)	5	mA
TJ	Junction temperature	-40	125	°C

- **Note 2:** It is recommended to avoid output switching conditions where the negative-going spikes at the  $V_S$  node would decrease  $V_S$  below ground by more than -5 V.
- **Note 3:** Enough current should be supplied to the  $V_{CC}$  pin of the IC to keep the internal 15.6 V zener diode clamping the voltage at this pin.

#### **Recommended Component Values**

Parameter				
Symbol	Component	Min.	Max.	Units
R⊤	Timing resistor value	1		kΩ
Ст	C <sub>T</sub> pin capacitor value	330		pF

 $V_{BIAS}$  ( $V_{CC}$ ,  $V_{BS}$ ) = 14 V,  $V_{S}$ =0 V and  $T_{A}$  = 25 °C, CLO = CHO = 1 nF.



For further information, see Fig. 12.



#### **Electrical Characteristics**

 $V_{\text{BIAS}}$  ( $V_{\text{CC}}$ ,  $V_{\text{BS}}$ ) = 14 V,  $C_{\text{T}}$  = 1 nF,  $V_{\text{S}}$ =0 V and  $T_{\text{A}}$  = 25 °C unless otherwise specified. The output voltage and current ( $V_{\text{O}}$  and  $I_{\text{O}}$ ) parameters are referenced to COM and are applicable to the respective output leads: HO or LO. CLO = CHO = 1 nF.

Symbol	Definition		Тур	Max	Units	Test Conditions		
Low Voltage Supply Characteristics								
V <sub>CCUV</sub> +	Rising V <sub>CC</sub> undervoltage lockout threshold	10.0	11.0	12.0				
V <sub>CCUV</sub> -	Falling V <sub>CC</sub> undervoltage lockout threshold	8.0	9.0	10.0	V			
Vccuvhys	V <sub>CC</sub> undervoltage lockout hysteresis	1.6	2.0	2.4				
Iqccuv	Micropower startup V <sub>CC</sub> supply current		130	170		V <sub>CC</sub> ≤ V <sub>CCUV</sub> -		
Iqcc	Quiescent V <sub>CC</sub> supply current		800	1000	μA			
Icc	V <sub>CC</sub> supply current		1.8		mA	$R_T = 36.9 \text{ k}\Omega$		
V <sub>CC CLAMP</sub>	V <sub>CC</sub> zener clamp voltage	14.4	15.4	16.8	V	$I_{CC} = 5 \text{ mA}$		
Floating	Supply Characteristics							
I <sub>QBS</sub>	Quiescent V <sub>BS</sub> supply current		60	80	μΑ			
V <sub>BSUV+</sub>	V <sub>BS</sub> supply undervoltage positive going threshold	8.0	9.0	9.5	V			
V <sub>BSUV</sub> -	V <sub>BS</sub> supply undervoltage negative going threshold	7.0	8.0	9.0	V			
I <sub>LK</sub>	Offset supply leakage current			50	μΑ	V <sub>B</sub> = V <sub>S</sub> = 600 V		
Oscillato	r I/O Characteristics		•					
	Oscillator fraguerov	18.4	19.0	19.6	- kHz	$R_T = 36.5 \text{ k}\Omega$		
f <sub>OSC</sub>	Oscillator frequency	88	88 93 100	100		$R_T = 7.15 \text{ k}\Omega$		
d	R <sub>T</sub> pin duty cycle		50		%	f <sub>o</sub> < 100 kHz		
Іст	C <sub>T</sub> pin current		0.02	1.0	μА			
$I_{CTUV}$	UV-mode C <sub>T</sub> pin pulldown current	0.20	0.30	0.6	mA	V <sub>CC</sub> = 7 V		
$V_{\text{CT+}}$	Upper C <sub>T</sub> ramp voltage threshold		9.32					
V <sub>CT</sub> -	Lower C <sub>T</sub> ramp voltage threshold		4.66		V			
V <sub>CTSD</sub>	C <sub>T</sub> voltage shutdown threshold	2.2	2.3	2.4				
$V_{RT}$ +	High-level R <sub>T</sub> output voltage, V <sub>CC</sub> - V <sub>RT</sub>		10	50		I <sub>RT</sub> = -100 μA		
VRI+	Thigh-level IX output voltage, vcc - vRI		100	300		$I_{RT} = -1 \text{ mA}$		
$V_{RT ext{-}}$	Low-level R⊤ output voltage		10	50		I <sub>RT</sub> = 100 μA		
VKI-	Low-level RT output voltage		100	300		I <sub>RT</sub> = 1 mA		
$V_{RTUV}$	UV-mode R <sub>T</sub> output voltage		0	100		$V_{CC} \leq V_{CCUV}$		
V	SD mode P- output voltage V V		10	50	mV	I <sub>RT</sub> = -100 μA, V <sub>CT</sub> = 0 V		
$V_{RTSD}$	SD-mode R <sub>T</sub> output voltage, V <sub>CC</sub> - V <sub>RT</sub>		100	300		I <sub>RT</sub> = -1 mA, V <sub>CT</sub> = 0 V		



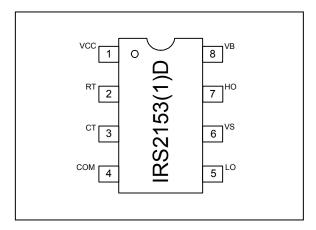
#### **Electrical Characteristics**

 $V_{BIAS}$  ( $V_{CC}$ ,  $V_{BS}$ ) = 14 V,  $C_T$  = 1 nF,  $V_S$ =0 V and  $T_A$  = 25 °C unless otherwise specified. The output voltage and current ( $V_o$  and  $I_o$ ) parameters are referenced to COM and are applicable to the respective output leads: HO or LO. CLO = CHO = 1 nF.

Symbol	Definition	Min	Тур	Max	Units	Test Conditions
Gate Dri	ver Output Characteristics					
V <sub>OH</sub>	High-level output voltage		V <sub>CC</sub>			L 0 A
V <sub>OL</sub>	Low-level output voltage		COM		V	I <sub>O</sub> = 0 A
V <sub>OL_UV</sub>	UV-mode output voltage		COM			$I_O = 0 A,$ $V_{CC} \le V_{CCUV}$
t <sub>r</sub>	Output rise time		120	220		
t <sub>f</sub>	Output fall time		50	80	ns	
t <sub>sd</sub>	Shutdown propagation delay		350			
t <sub>d</sub>	Output deadtime (HO or LO) (IRS2153D)	0.65	1.1	1.75	μS	
t <sub>d</sub>	Output deadtime (HO or LO) (IRS21531D)	0.35	0.6	0.85	μS	
I <sub>O+</sub>	Output source current		180		mA	
I <sub>O-</sub>	Output sink current		260		IIIA	
Bootstra	p FET Characteristics					
$V_{B\_ON}$	V <sub>B</sub> when the bootstrap FET is on		13.7		V	
I <sub>B_CAP</sub>	V <sub>B</sub> source current when FET is on	40	55		mA	C <sub>BS</sub> =0.1 uF
I <sub>B_10V</sub>	V <sub>B</sub> source current when FET is on	10	12		111/4	V <sub>B</sub> =10 V

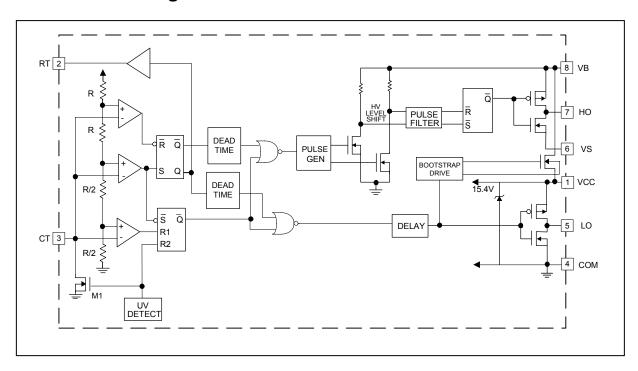
# International TOR Rectifier

## **Lead Definitions**



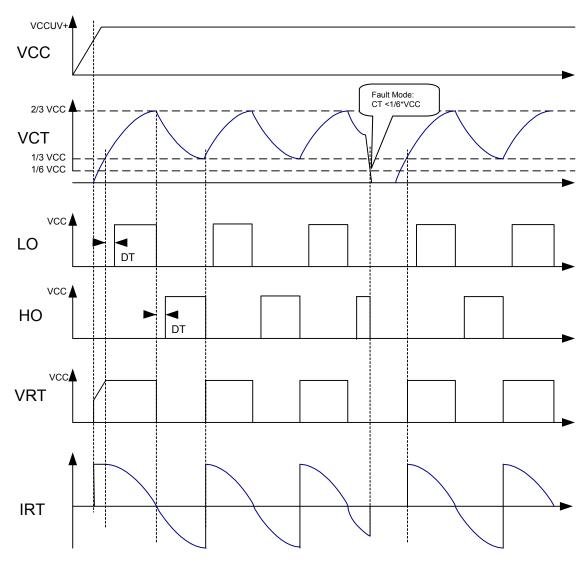
	Lead					
Symbol	Description					
V <sub>CC</sub>	Logic and internal gate drive supply voltage					
R <sub>T</sub>	Oscillator timing resistor input					
Ст	Oscillator timing capacitor input					
COM	IC power and signal ground					
LO	Low-side gate driver output					
Vs	High voltage floating supply return					
НО	High-side gate driver output					
V <sub>B</sub>	High side gate driver floating supply					

## **Functional Block Diagram**



## **Timing Diagram**

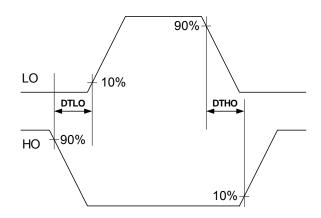
#### **Operating Mode**



#### **Switching Time Waveform**

# HO LO 10%

#### **Deadtime Waveform**





#### **Functional Description**

#### **Under-voltage Lock-Out Mode (UVLO)**

The under-voltage lockout mode (UVLO) is defined as the state the IC is in when  $V_{\rm CC}$  is below the turn-on threshold of the IC. The IRS2153(1)D under voltage lock-out is designed to maintain an ultra low supply current of less than 170  $\mu A$ , and to guarantee the IC is fully functional before the high and low side output drivers are activated. During under voltage lock-out mode, the high and low-side driver outputs HO and LO are both low.

#### Supply voltage

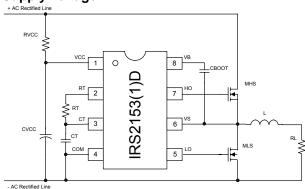


Fig. 1 Typical Connection Diagram

Fig. 1 shows an example of supply voltage. The start-up capacitor ( $C_{\text{VCC}}$ ) is charged by current through supply resistor ( $R_{\text{VCC}}$ ) minus the start-up current drawn by the IC. This resistor is chosen to provide sufficient current to supply the IRS2153(1)D from the DC bus.  $C_{\text{VCC}}$  should be large enough to hold the voltage at Vcc above the UVLO threshold for one half cycle of the line voltage as it will only be charged at the peak, typically 0.1 uF. It will be necessary for  $R_{\text{VCC}}$  to dissipate around 1 W.

The use of a two diode charge pump made of DC1, DC2 and CVS (Fig. 2) from the half bridge ( $V_S$ ) is also possible however the above approach is simplest and the dissipation in  $R_{VCC}$  should not be unacceptably high.

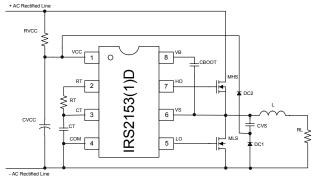


Fig. 2 Charge pump circuit

The supply resistor ( $R_{\text{VCC}}$ ) must be selected such that enough supply current is available over all operating conditions.

Once the capacitor voltage on  $V_{\text{CC}}$  reaches the start-up threshold  $V_{\text{CCUV+}}$ , the IC turns on and HO and LO begin to oscillate.

#### **Bootstrap MOSFET**

The internal bootstrap FET and supply capacitor ( $C_{\text{BOOT}}$ ) comprise the supply voltage for the high side driver circuitry. The internal boostrap FET only turns on when LO is high. To guarantee that the high-side supply is charged up before the first pulse on pin HO, the first pulse from the output drivers comes from the LO pin.

#### Normal operating mode

Once the  $V_{CCUV^+}$  threshold is passed, the MOSFET M1 opens, RT increases to approximately  $V_{CC}$  ( $V_{CC^-}V_{RT^+}$ ) and the external CT capacitor starts charging. Once the CT voltage reaches  $V_{CT^-}$  (about 1/3 of  $V_{CC}$ ), established by an internal resistor ladder, LO turns on with a delay equivalent to the deadtime (t<sub>a</sub>). Once the CT voltage reaches  $V_{CT^+}$  (approximately 2/3 of  $V_{CC}$ ), LO goes low, RT goes down to approximately ground ( $V_{RT^-}$ ), the CT capacitor discharges and the deadtime circuit is activated. At the end of the deadtime, HO goes high. Once the CT voltage reaches  $V_{CT^-}$  HO goes low, RT goes high again, the deadtime is activated. At the end of the deadtime, LO goes high and the cycle starts over again.

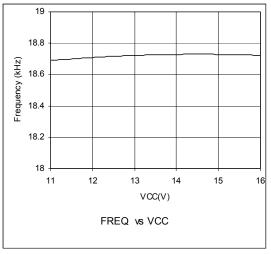
The following equation provides the oscillator frequency:

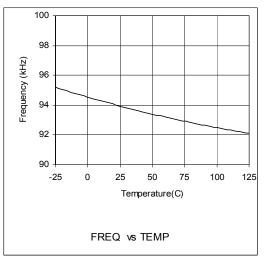
$$f \sim \frac{1}{1.453 \times RT \times CT}$$

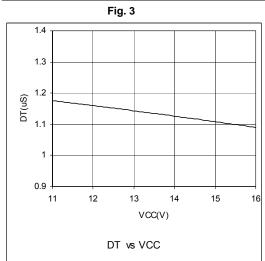
This equation can vary slightly from actual measurements due to internal comparator over- and under-shoot delays. For a more accurate determination of the output frequency, the frequency characteristic curves should be used (RT vs. Frequency, page 3).

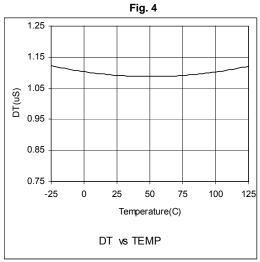
#### Shut-down

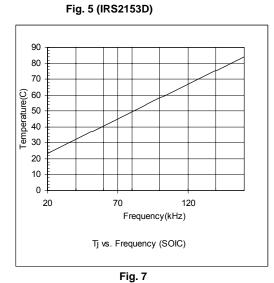
If CT is pulled down below VCTSD (approximately 1/6 of V<sub>CC</sub>) by an external circuit, CT doesn't charge up and oscillation stops. LO is held low and the bootstrap FET is off. Oscillation will resume once CT is able to charge up again to V<sub>CT</sub>.

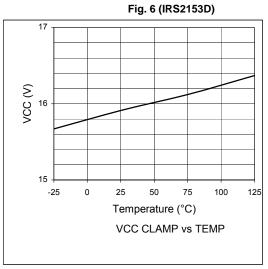


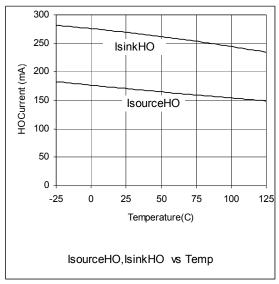


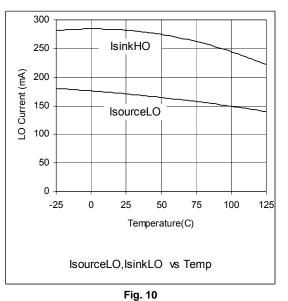


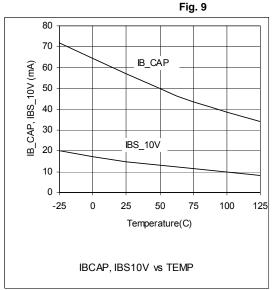












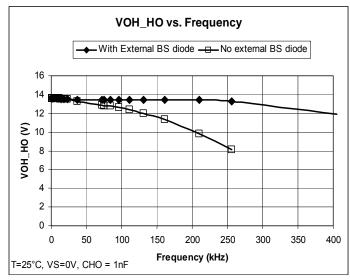


Fig. 11 Fig. 12

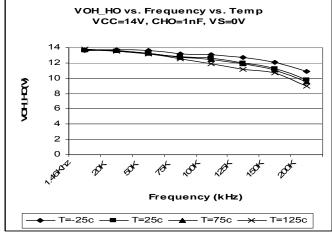
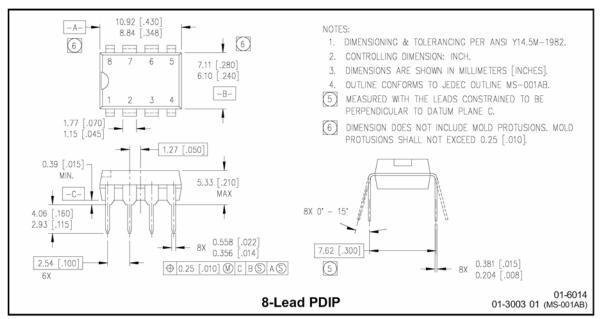
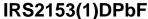
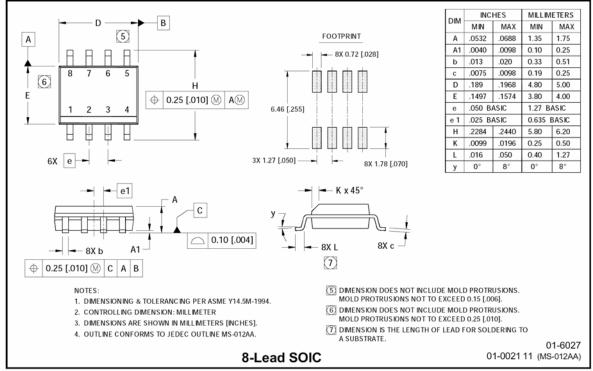


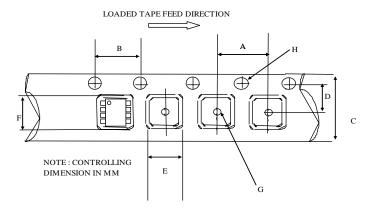
Fig. 13





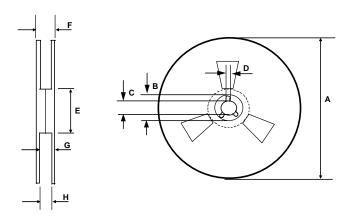


IRS2153(1)DSPbF



#### CARRIER TAPE DIMENSION FOR 8SOICN

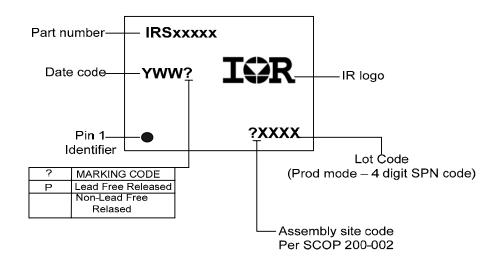
CARRIER TALE DIMENSION FOR 030101							
	Me	etric	Imperial				
Code	Min	Max	Min	Max			
Α	7.90	8.10	0.311	0.318			
В	3.90	4.10	0.153	0.161			
С	11.70	12.30	0.46	0.484			
D	5.45	5.55	0.214	0.218			
E	6.30	6.50	0.248	0.255			
F	5.10	5.30	0.200	0.208			
G	1.50	n/a	0.059	n/a			
Н	1.50	1.60	0.059	0.062			



#### REEL DIMENSIONS FOR 8SOICN

	Me	etric	Imp	erial
Code	Min	Max	Min	Max
Α	329.60	330.25	12.976	13.001
В	20.95	21.45	0.824	0.844
С	12.80	13.20	0.503	0.519
D	1.95	2.45	0.767	0.096
E F	98.00	102.00	3.858	4.015
	n/a	18.40	n/a	0.724
G H	14.50	17.10	0.570	0.673
Н	12.40	14.40	0.488 0.566	

#### PART MARKING INFORMATION



#### ORDER INFORMATION

8-Lead PDIP IRS2153DPbF

8-Lead PDIP IRS21531DPbF

8-Lead SOIC IRS2153DSPbF

8-Lead SOIC IRS21531DSPbF

8-Lead SOIC Tape & Reel IRS2153DSTRPbF

8-Lead SOIC Tape & Reel IRS21531DSTRPbF

## International TOR Rectifier

The SOIC-8 is MSL2 qualified.

This product has been designed and qualified for the industrial level.

Qualification standards can be found at <a href="www.irf.com">www.irf.com</a> <a href="http://www.irf.com">www.irf.com</a> <a href="http://www.irf.com">http://www.irf.com</a> <a href="http://www.irf.com">IR WORLD HEADQUARTERS: 233 Kansas St., El Segundo, California 90245, Tel: (310) 252-7105</a>

Data and specifications subject to change without notice. 6/27/2006



#### REVISION HISTORY

Date	Revision	Description of Change							
		The following	specs have be	een updated. Se	e PCN S02100	6			
				Previous			Rev. A		
2/10/06	A		Min	Тур	Max	Min	Тур	Max	
		VCCUV+	10.2	10.8	11.5	10.0	11.0	12.0	
		VCCUV-	8.3	8.8	9.4	8.0	9.0	10.0	
3/4/06	В	Addition of F	Addition of Fig 12.						
5/11/06	С	Formatting ch	Formatting changes.						
6/21/06	D	Updated Fig. 12 to add temperature test condition. Addition of Fig. 13. Addition of MSL2 statement.							
7/28/06	E		Added the specifications for RIS21531D. The only difference between IRS2153D and IRS21531D is the deadtime, 1.1 µs and 0.6 µs respectively (see page 5).						