

Application Note

Sensorless Brushless DC Motor Control with the Z16FMC MCU

AN031102-0311

Abstract

This application note discusses the closed loop control of a 3-phase brushless direct current (BLDC) motor using the Z16FMC Family of Microcontrollers (MCUs). The Z16FMC product family is designed specifically for motor control applications, featuring an on-chip integrated array of application-specific analog and digital modules. The result is fast and precise fault control, high system efficiency, on-the-fly speed/torque and direction control, as well as ease of firmware development for customized applications.

This document further discusses ways in which to implement a sensorless feedback control system using a Phase Locked Loop along with back EMF sensing. Test results are based on using the Z16FMC Modular Development System (MDS) module, a 3-phase motor control (MC) application board and a 3-phase 24 VDC, 30 W, 3200 RPM BLDC motor with internal Hall sensors.



Note: The source code files associated with this application note, AN0311-SC01 and AN0311-SC02, were tested with version 4.12 of ZDSII for ZNEO MCUs. Subsequent releases of ZDSII may require you to modify the code supplied with this application note.

The sample project included in ZDSII v4.12 and the firmware in the Rev D (or earlier) version of the Z16F28200KITG Development Kit were preprogrammed with AN0311-SC01. The source code files contained in AN0226-SC02 are enhanced versions of AN0311-SC01 that allow users to easily change parameters to accommodate differing motor types.

Revision History

Each instance in the following table reflects a change to this document from its previous version. For more details, refer to the corresponding pages or appropriate links provided in the table.

Date	Revision Level	Description	Page Number
Mar 2011	05	AN0311-SC02 source code added to encompass multiple motor types; correction to TimerPrescale data in PLL flow, Figure 17.	<u>1, 22</u>
Dec 2010	01	Original issue.	All

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Features

The power-saving features of this Z16FMC application include:

- Smooth S-curve motor start-up with reduced starting current
- Sensorless (back-EMF) control using Phase Locked Loop feedback
- Microcontroller-based overcurrent protection
- Selectable speed or torque setting
- Selectable speed or torque control
- Selectable control of motor direction
- UART Interface for PC control
- LED for max speed indication
- LED for motoring running indication
- LED for fault indication

Discussion

The Z16FMC Series Flash microcontrollers are based on Zilog's advanced ZNEO 16-bit CPU core. The Z16FMC MCU family of devices set a new standard of performance and efficiency with up to 20 MIPS performance at 20 MHz. It supports 16-bit internal and external bus widths and provides near-single-cycle instruction execution.

The Z16FMC's external interface allows seamless connection to external memory and peripherals. A 24-bit address bus and a selectable 8-bit or 16-bit data bus allows parallel access up to 16MB. Up to 128KB internal Flash memory is accessible by the CPU, 16 bits at a time, to improve processor throughput. Up to 4KB of internal RAM provides storage of data, variables and stack operations.

Figure 1 displays a block diagram of the Z16FMC MCU architecture.

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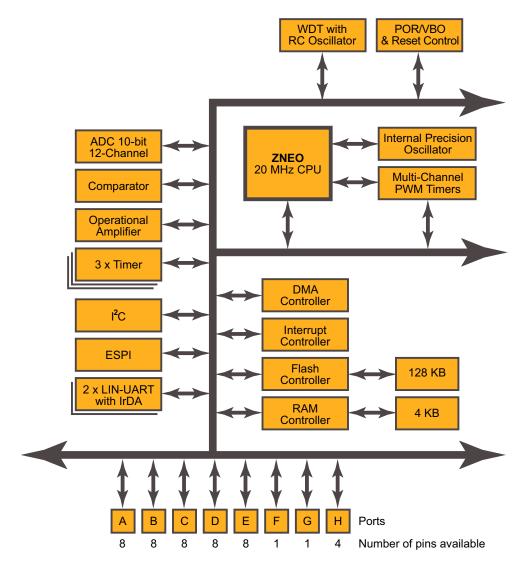


Figure 1. The Z16FMC MCU Architecture

In each of the Z16FMC products, the novel device architecture allows for realization of a number of enhanced control features:

- Time Stamp for Speed Control
- Integrated Operational Amplifier
- Multi-Channel PWM Timer

Time Stamp for Speed Control

Most microcontrollers use at least one dedicated comparator to detect the zero crossing of the input AC voltage signal so that the output driving pulses can be synchronized and

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adjusted to properly regulate the motor speed. An alternative approach based on Zilog's motor control MCU eliminates the need for this comparator by instead employing an analog to digital converter (ADC) in conjunction with a timer. In this case, the ADC samples the AC line voltage, with the timer running in the background.

When the ADC samples the line voltage's zero crossing, it reads the timer count and writes the result to a register. As a result, the timers are cued for the output Pulse Width Modulation (PWM) pulses to efficiently regulate the speed of the motor. This time stamp approach results in a very simple and cost-effective solution for smooth operation of the motor in steady state.

Integrated Operational Amplifier

Appliance controllers almost invariably monitor motor speed by sensing the current through the windings, using sensor and sensorless techniques in conjunction with the ADC. Ordinarily, sampling instances by the ADC are synchronized by the MCU. With this process, an external operational amplifier is often used to convert the current signal to a voltage signal; the ADC next samples the voltage signal and outputs the result to the processor. The processor then synthesizes the PWM outputs to control motor speed.

In the case of the Z16FMC Family of Microcontrollers, an on-chip integrated operational amplifier eliminates the requirement for an external component, thereby reducing overall system cost.

Multi-Channel PWM Timer

The Z16FMC features a flexible PWM module with three complementary pairs - or six independent PWM outputs - supporting deadband operation and fault protection trip input. These features provide multiphase control capability for a variety of motor types and ensure safe operation of the motor by providing immediate shutdown of the PWM pins during a fault condition.

Theory of Operation

In a brushless DC motor, the rotor is comprised of permanent magnets while the stator windings are similar to those in poly-phase motors. For a detailed discussion of BLDC motor fundamentals, as well as closed-loop control using sensorless techniques, refer to the Motor Control Electronics Handbook by: Richard Valentine, McGraw-Hill, NY, 1998.

In sensor-based control applications, the Hall elements are integrated and are used to detect the position of the rotor for drive synchronization. In contrast, sensorless control employs the detection of Back EMF signals which are generated (induced) by specific phase windings to synchronize the timing of the control loop.

A block diagram of the BLDC motor control system based on the Z16FMC MCU is shown in Figure 2. In a 3-phase commutation arrangement, at any given instance, only two phases are energized. The back EMF voltage is in turn generated in the unenergized phase winding, and the zero crossing of this induced voltage is detected for synchronization of the subsequent closed-loop control events. As discussed earlier, the innovative time stamp feature of the Z16FMC MCU provides for robust, efficient implementation of this critical sensing function without the requirement for an additional comparator.

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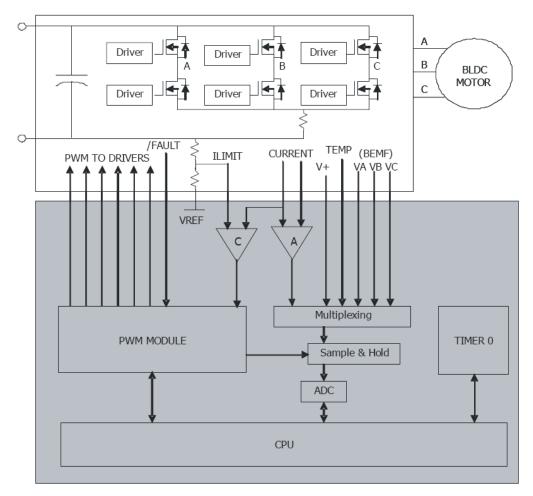


Figure 2. . A 3-Phase BLDC Motor Control System

The algorithm for Back EMF sensing is based on an implementation of a Phase Locked Loop (PLL), as shown in Figures 3 and 4 and described in Appendix C, Back EMF Sensing Phase Locked Loop. This algorithm is especially advantageous during startup, resulting in a very smooth increase in the motor speed as well as nearly instantaneous reversal of direction of rotation on command, as outlined below.

With a conventional approach during the start-up sequence, power is applied to the windings in order to place the rotor in a known starting position, followed by commutation and start of Back EMF sensing and control. In contrast, the PLL-based approach implemented herein makes it possible to lock the Back EMF signal from the onset of the start-up phase without the requirement for initial placement of the rotor in a specific position. Moreover, this approach significantly reduces any erratic movement of the motor during start-up or a reversal of direction.

During normal operation following the start-up period, phase torque/current mode control is achieved via sensing of the voltage generated across a sense resistor in the motor drive

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circuit. This voltage is routed to the on-chip integrated ADC, after which data processing by the CPU, based on a predefined computational algorithm, results in the regulation of the PWM commutation signal period(s).

As discussed earlier, another key feature of the Z16FMC MCU is the direct coupling of the on-chip integrated comparator to the PWM module to enable fast, cycle-by-cycle shutdown during an overcurrent fault event. Oscilloscope-generated waveforms representing this sequence of events are shown in Figure 5.

In conjunction with the integrated on-chip hardware blocks, the 3-phase BLDC motor control software developed herein allows for ease of programming to achieve the desired closed-loop control characteristics. The routines that enable the sensing of the motor's Back EMF and current are all interrupt-driven. It is critical that the highest interrupt priority is assigned to the Back EMF sensing event for subsequent synchronization of the commutation events. In this case, Timer 0 is used for the Time Stamp function as well as for updating the commutation period, if necessary.

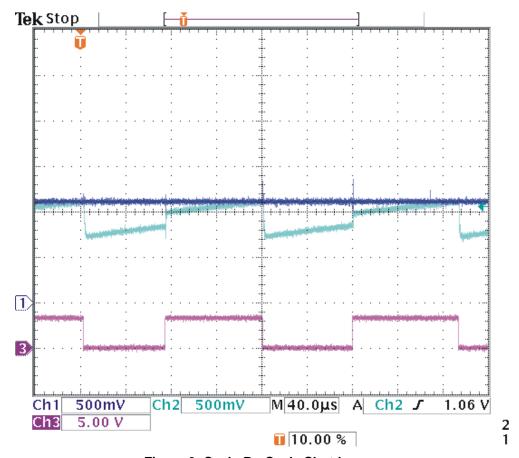


Figure 3. Cycle-By-Cycle Shutdown

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Testing

This section provides information about how to run the code and demonstrate this application, including the equipment used to build the implementation, its configuration and the results of its testing.

Equipment Used

The following equipment is used for the setup; the first five items are contained in the Z16FMC Series Development Kit (Z16FMC28200KITG).

- Zilog Z16FMC MDS Module (99C1299-001)
- Zilog 3-Phase Motor Control Application Board (99C0960-001)
- Opto-Isolated USB Smart Cable to connect the PC to the Z16FMC Series Development Board
- LINIX 3 Phase BLDC motor 24VDc, 30W 3200RPM (45ZWN24-30)
- 5 V DC power supply for the Z16FMC MDS Module
- 24 V DC power supply for BLDC motor
- Digital Oscilloscope or Logic Analyzer

Hardware Setup

Figures 4 and 5 illustrate the application hardware connections.

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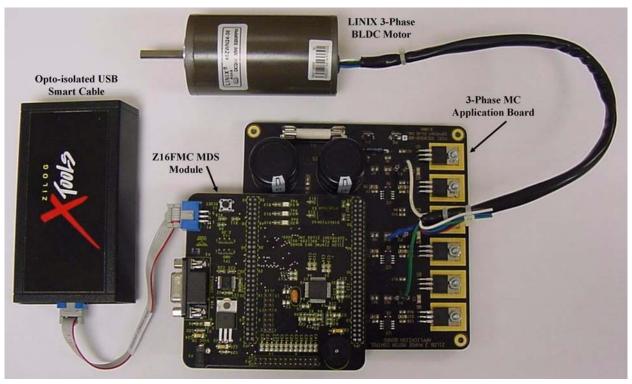


Figure 4. Z16F2800100ZCOG and Motor Control Application Board

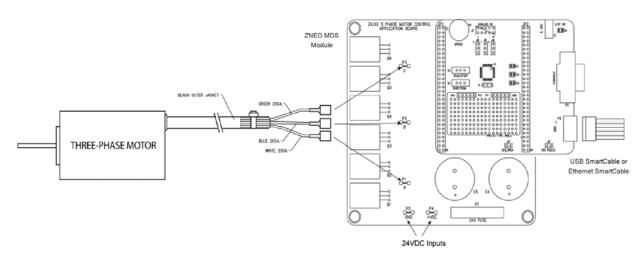


Figure 5. 3-Phase BLDC Hardware Setup with Z16FMC MDS Module

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Figure 6 displays the proper port settings in the HyperTerminal emulation program.



Figure 6. HyperTerminal Display Settings

Procedure

Observe the following steps to test the 3-Phase Sensorless BLDC Motor Control Application demo program to the Z16FMC MDS Module.

- 1. Install the ZDSII ZNEO version 4.12 or newer software on your PC.
- 2. Connect the Opto-Isolated USB Smart Cable to the PC.
 - To install the driver of the Opto-Isolated USB Smart Cable, refer to the installation guide of the Opto-Isolated USB Smart Cable included in the Z16FMC Series Development Kit.
- 3. Connect the Hardware Setup as shown in Figure 7.
- 4. Power up the Z16FMC MDS board using the 5 VDC adapter included in the kit.
- 5. Open the AN0311-SC01 project in ZDSII for ZNEO.

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6. From the main.c source file, choose the following mode for the Motor Control application:

#define LOOP_SELECT_VALUE 1u // 0u = torque loop, 1u = speed loop

- 7. Compile the application and download the code to the Z16FMC MDS module.
- 8. Stop the debug mode from the IDE; disconnect the Opto-Isolated USB Smart Cable and switch off the power supply to the Z16FMC development board.
- 9. Connect the 24V DC supply source to the MC application board.
- 10. Ensure that the RUN/STOP switch on the Z16FMC development board is in the STOP position.
- 11. First apply power to the Z16FMC development board supply, then apply power via the 24V supply to the MC application board.
- 12. Set the RUN/STOP switch on the Z16FMC development board to RUN.
- 13. Additionally, observe the following points:
 - If SPEED mode is selected, the speed of the motor can be varied by adjusting the potentiometer on the Z16FMC development board.
 - If TORQUE mode is selected, the motor speed is decreased with application of force on the shaft of the motor.
 - The direction of rotation of the motor is set by changing the position of the direction switch on the Z16FMC development board.

You can now add your application software to the main program to experiment with additional functions.



Note: While debugging your code, ensure that the Opto-Isolated USB Smart Cable controls the reset pin of the MCU. After debugging and running your code, detach the Opto-Isolated USB Smart Cable from P3 of the Z16F28200KITG to free the Reset pin and apply a power cycle to reset the MCU from debug mode.

Result

This Motor Control application was tested with the Z16FMC MDS board connected to Zilog's 3-phase motor control application board.

The BLDC motor specifications are:

Manufacturer: Linix

Motor type: 3-wire, 3-phase brushless DC motor

Voltage rating: 24 V

Power rating: 30W

Maximum speed of rotation: 3200RPM

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Observations that we noted about speed and torque are indicated in Table 1.

Table 1. Speed and Torque Observations

Function	Description	Observation	
Speed	RUN/STOP switch in STOP position	Motor is in idle mode.Yellow LED blinks.	
	 RUN/STOP switch in RUN position. Direction switch set to a clockwise rotation. Potentiometer R10 at minimum value. 	 Motor starts rotating. When the motor is loaded mechanically by holding the shaft, speed initially decreases then picks up gradually; current increases. Green LED blinks. 	
	 RUN/STOP switch in RUN position. Direction switch set to a clockwise rotation. Potentiometer R10 at maximum value. 	Motor starts rotating with no load.Green LED blinks.	
	 RUN/STOP switch in RUN position. Direction switch set to a counterclockwise rotation. Potentiometer R10 at minimum value. 	 Motor starts rotating at a speed of 1280 RPM. When the motor is loaded mechanically by holding the shaft, speed initially decreases then picks up gradually; current increases. Green LED blinks. 	
	 RUN/STOP switch in RUN position. Direction switch set to a counterclockwise rotation. Potentiometer R10 at maximum value. 	 Motor begins rotating at a speed of 3890 RPM with no load. Green LED blinks. 	
Torque	 RUN/STOP switch in RUN position. Direction switch set to a clockwise direction. 	 Motor begins rotating at a speed of 1250 RPM at no load. Motor stops rotating upon holding the shaft Constant current consumption of 60mA. Green LED blinks. 	
	 RUN/STOP switch in RUN position. Direction switch set to a counterclockwise direction. 	 Motor starts rotating at a speed of 1250 RPM at no load. Motor stops rotating upon holding the shaft. Constant current consumption of 60mA. Green LED blinks. 	
	RUN/STOP switch in STOP position.	Motor is in idle mode. Yellow LED blinks.	

Summary

This Application Note described the closed-loop control of a sensorless BLDC motor using the advanced on-chip integrated features of the Z16FMC MCU. The Z16FMC product line is ideally suited for such applications, providing for a seamless start-up of the motor from an idle mode to full operational speed, on-the-fly reversal of the direction of rotation, an extremely fast fault detection cycle and a lower total solution cost. These features, along with the powerful ZNEO CPU core and some of the best development tools

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available in the industry, result in less complex board designs and reduced design cycle time.

References

The following documents are associated with the Z16FMC Series of Motor Control MCUs; each is available for download on www.zilog.com.

- Z16FMC Series Motor Control MCU Product Specification (PS0287)
- Z16FMC Series Motor Control Development Kit User Manual (UM0234)
- Z16FMC Series Motor Control Development Kit Quick Start Guide (QS0079)
- Zilog Developer Studio II ZNEO User Manual (UM0171)
- ZNEO CPU Core User Manual (UM0188)
- Sensorless Brushless DC Motor Control with Z8 Encore! MC Microcontrollers Application Note (AN0226)

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Appendix A. Schematics

Figures 7 and 8 show basic block and MCU schematics, respectively, for the Z16FMC Motor Control MDS Module.

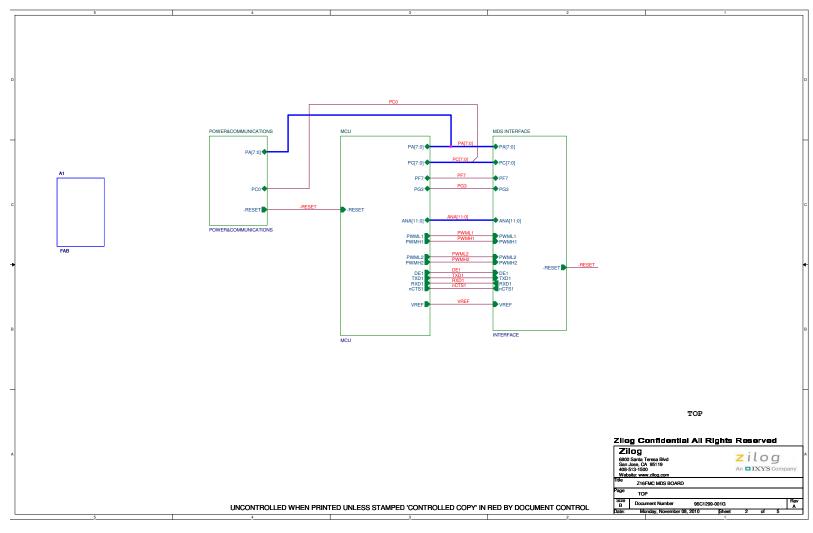


Figure 7. Z16FMC Motor Control MDS Module, #1 of 2

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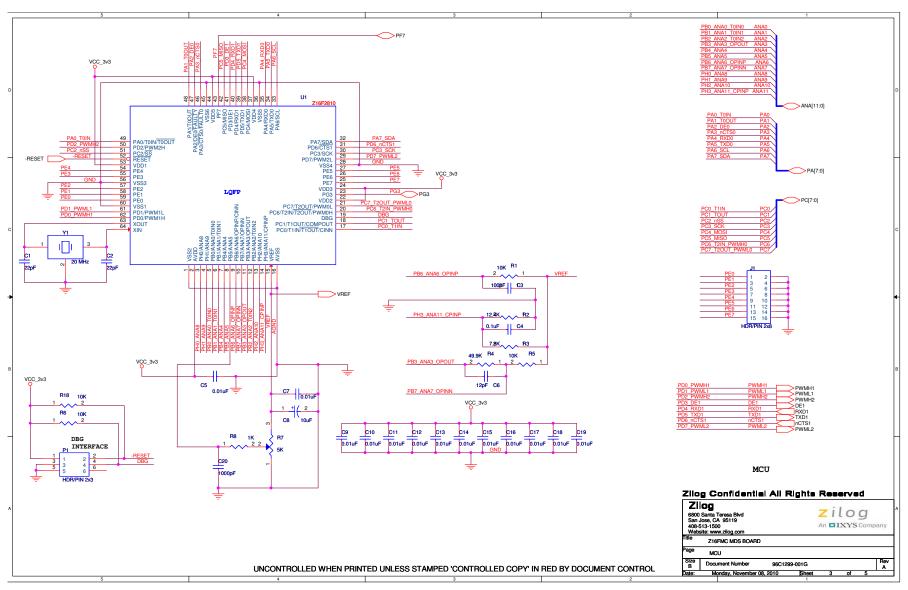


Figure 8. Z16FMC Motor Control MDS Module, #2 of 2

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Figure 9 shows the schematics for the MDS board's power and serial interfaces.

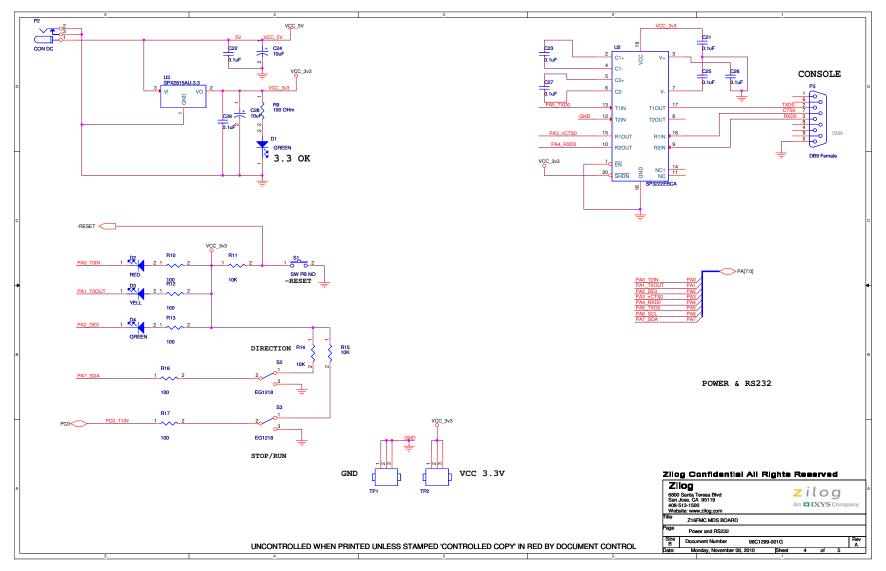


Figure 9. Z16FMC MDS Board Power and RS-232 Connections

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Figure 10 displays MDS interface schematics.

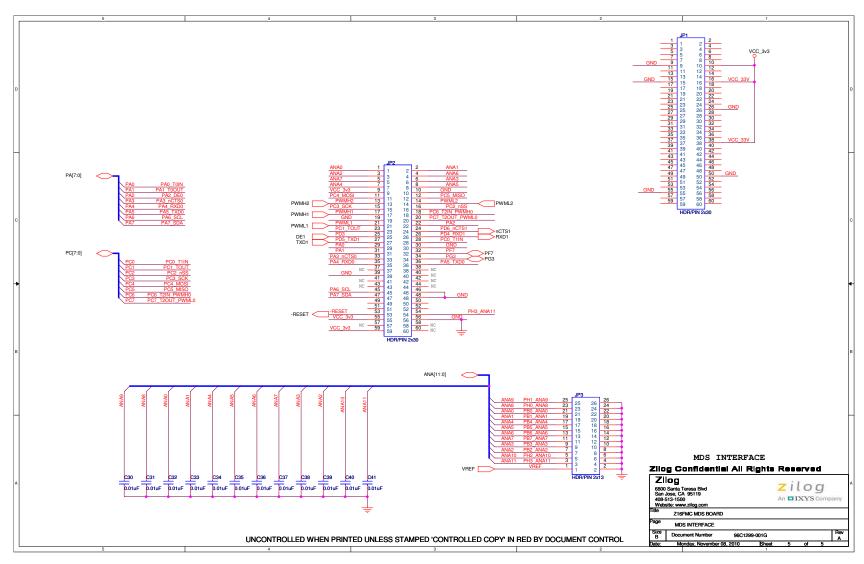


Figure 10. MDS Board Interfaces

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Figures 11 through 13 display the schematics for the 3-Phase Motor Control Application Board.

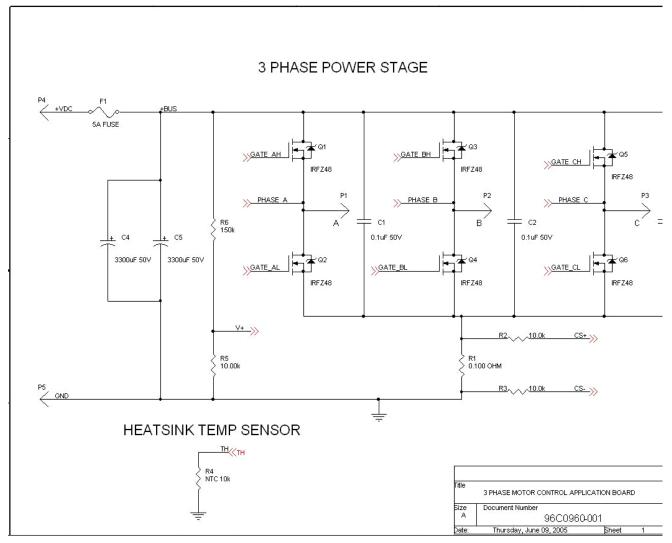


Figure 11. 3-Phase Motor Control Application Board, #1 of 3

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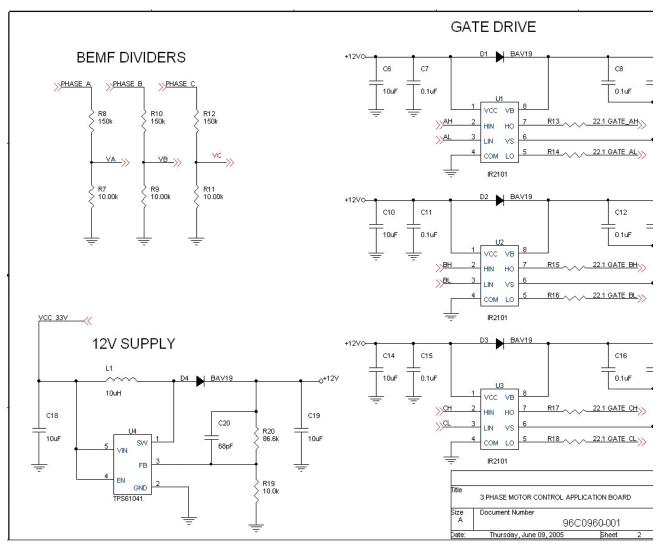


Figure 12. 3-Phase Motor Control Application Board, #2 of 3

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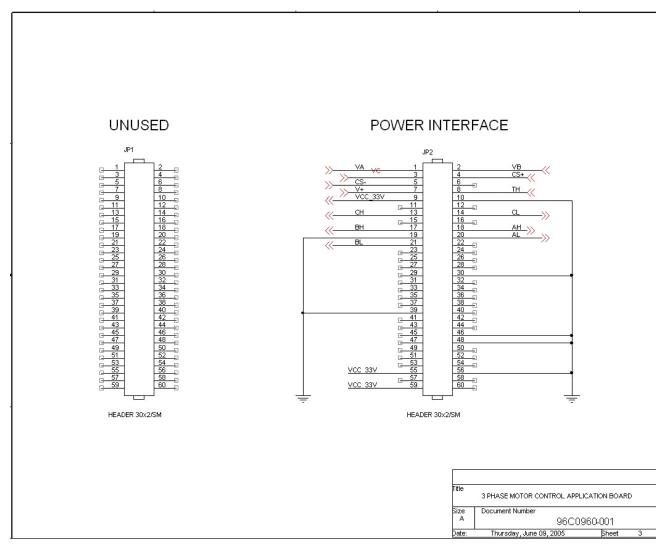


Figure 13. 3-Phase Motor Control Application Board, #3 of 3

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Appendix B. Flowcharts

This appendix displays flow charts that diagram the Main Function and the Read and Write APIs.

Figure 14 shows the main control loop.

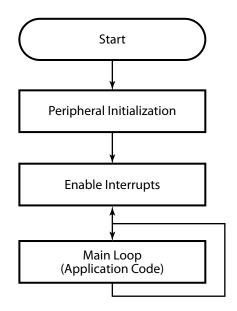


Figure 14. Initialization and Application Code Space

The Back EMF sensing loop is shown in Figure 15.

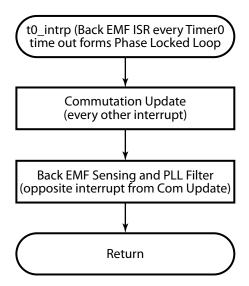


Figure 15. Initialization and Application Code Space

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A flow chart of the PWM loop is shown in Figure 16. This PWM loop can also be used for specific application code, such as communications or additional user interfaces.

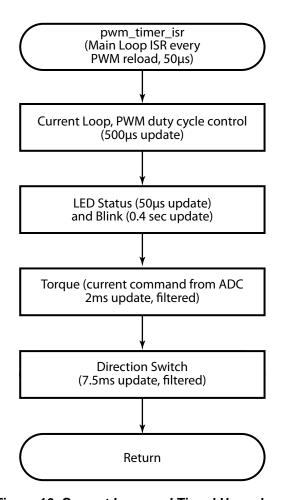


Figure 16. Current Loop and Timed Housekeeping

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Appendix C. Back EMF Sensing Phase Locked Loop

The Phase Locked Loop Back EMF algorithm, implemented to provide a smooth start-up of the motor, is shown in Figures 17 and 18. Additional details about the specific formulas in these figures are shown in Table 2; a description of these calculations follows.

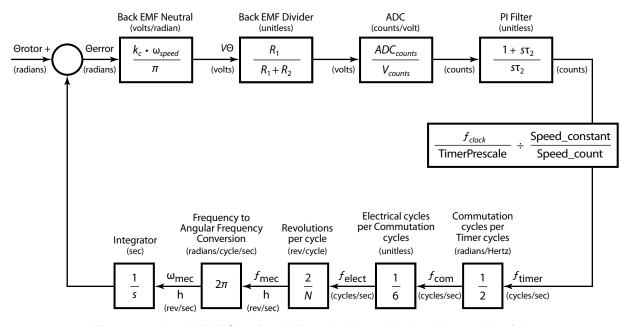


Figure 17. Back EMF Sensing Using the Phase Locked Loop Algorithm

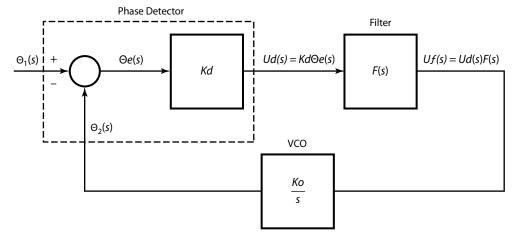


Figure 18. Proportional Integral (PI) Filter Representation for Back EMF Sensing

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Table 2. Back EMF Sensing Phase Locked Loop

$D = \frac{R_1}{R_1 + R_2}$	Divider Ratio	
$K_d = \frac{K_e \cdot \omega speed \cdot D}{2\pi}$	(volts/rad) ωspeed = current speed of motor	
$F(s) = \frac{1 + s\tau_2}{s\tau_1}$	τ_1 = numeric constant τ_2 = numeric constant	
$K_{O} = \frac{ADC_{counts} \cdot fclock \cdot 2 \cdot 2\pi}{Vref \cdot Prescaler \cdot 6 \cdot N \cdot K_{speed}}$	(rad/sec/volt)	
$K_{speed} = \frac{(2 \cdot 2\pi \cdot fclock \cdot speed_count_max)}{(2 \cdot 6 \cdot N \cdot \omega_{max} \cdot Prescaler)}$	Speed_count_max = counts at max speed ω max = Maximum motor speed (rad/sec)	
$K_{O} = \frac{ADC_{counts} \cdot \omega_{max}}{Vref \cdot speed_{count_{max}}}$	(rad/sec/volt)	
$\omega n = \sqrt{Kd \cdot Ko \cdot \tau_1} = \omega_{\text{max}}$	Natural frequency	
$\zeta = \frac{\omega n \cdot \tau_2}{2} = 0.707$	Damping factor	
$H(s) = \frac{\theta_2(s)}{\theta_1(s)} = \frac{2 \cdot s \cdot \zeta \cdot \omega_n + \omega_n^2}{s^2 + 2 \cdot s \cdot \zeta \cdot \omega_n + \omega_n^2}$	Closed Loop Transfer Function	
$A_{oi}(s) = Kd \cdot F(s) \cdot \frac{Ko}{s} = Kd \cdot Ko \cdot \frac{1 + s\tau_2}{s^2 \tau_1}$		
$A_{ol}(s) = \frac{s \cdot Kd \cdot Ko \cdot \tau_2 + Kd + Ko}{s^2 \tau_1}$	Open Loop Gain	

We begin with the transfer function of the Proportional Integral (PI) Filter in the s-plane:

$$F(s) = \frac{Y(s)}{R(s)} = \frac{1 + s\tau_2}{s\tau_1}$$

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Next, by using the bilinear transform identity:

$$s = \frac{2}{T} \frac{z - 1}{z + 1}$$

where T = the sampling period, yields the following equation.

$$F(z) = \frac{Y(z)}{R(z)} = \frac{1 + \left(\frac{2z-1}{Tz+1}\right)\tau_2}{\left(\frac{2z-1}{Tz+1}\right)\tau_1}$$

When multiplying by:

$$Tz-T$$

the calculations that follow are:

$$F(z) = \frac{Y(z)}{R(z)} = \frac{Tz + T + (2\tau_2)(z - 2\tau_2)}{(2\tau_1)z - 2\tau_1}$$

$$zY(z)-Y(z) = \bigg(\frac{T+2\tau_2}{2\tau_1}\bigg)zR(z) + \bigg(\frac{T-2\tau_2}{2\tau_1}\bigg)R(z)$$

$$zY(z)-Y(z)\,=\,a_0zR(z)+a_1R(z)$$

where:

$$a_0 = \frac{T + 2\tau_2}{2\tau_1} \qquad \qquad a_1 = \frac{T - 2\tau_2}{2\tau_1}$$

and:

$$Y(z) = z^{-1}Y(z) + a_0R(z) + a_1z^{-1}R(z)$$

Collecting terms and dividing by z yields the following result:

$$y(n) = y(n-1) + a_0r(n) + a_1r(n-1)$$

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When writing this computation as a computer program, it takes the form of a recursive filter, with the coefficients A0 and A1:

$$Y0 = Y1 + A0 * R0 - A1 * R1$$

where:

- Y0 = Current output
- Y1 = Output at the last sample period
- R0 = Current ADC sample of Back EMF (phase voltage $-V_{BUS}/2$)
- R1 = Most recent sample of Back EMF from ADC
- A0 = a0
- $\bullet \quad A1 = -a1$

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