

M74HC423A

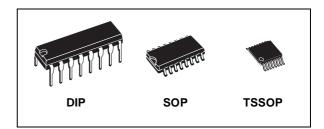
DUAL RETRIGGERABLE MONOSTABLE MULTIVIBRATOR

- HIGH SPEED : $t_{PD} = 22 \text{ ns} (TYP.) \text{ at } V_{CC} = 6V$
- LOW POWER DISSIPATION: STAND BY STATE : I_{CC}=4µA (MAX.) at T_A=25°C ACTIVE STATE : I_{CC} =700µA (TYP.) at V_{CC} = 6V
- HIGH NOISE IMMUNITY: $V_{NIH} = V_{NIL} = 28 \% V_{CC}$ (MIN.)
- SYMMETRICAL OUTPUT IMPEDANCE: $|I_{OH}| = I_{OL} = 4mA (MIN)$
- BALANCED PROPAGATION DELAYS: $t_{PLH} \cong t_{PHL}$
- WIDE OPERATING VOLTAGE RANGE: V_{CC} (OPR) = 2V to 6V
- WIDE OUTPUT PULSE WIDTH RANGE : t_{WOUT} = 120 ns ~ 60 s OVER AT V_{CC} = 4.5 V
- PIN AND FUNCTION COMPATIBLE WITH **74 SERIES 423A**

DESCRIPTION

The M74HC423A is an high speed CMOS MONOSTABLE MULTIVIBRATOR fabricated with silicon gate C²MOS technology.

There are two trigger inputs, A INPUT (negative edge) and B INPUT (positive edge). These inputs are valid for slow rising/falling signals, (tr = tf = 1)sec). After triggering the output maintains the MONOSTABLE state for the time period



ORDER CODES

PACKAGE	TUBE	T & R
DIP	M74HC423AB1R	
SOP	M74HC423AM1R	M74HC423ARM13TR
TSSOP		M74HC423ATTR

determined by the external resistor Rx and capacitor Cx.

The pulse width constant is $K \cong 1$.

Taking CLR low breaks this MONOSTABLE STATE. If the next trigger pulse occurs during the MONOSTABLE period it makes the MONOSTABLE period longer. Limit for values of Cx and Rx :

Cx : NO LIMIT

 $Rx : V_{cc} < 3.0V 5K\Omega$ to $1M\Omega$

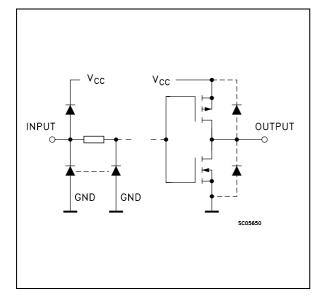
 $V_{cc} \geq 3.0V \ 1K\Omega \text{ to } 1M\Omega$

All inputs are equipped with protection circuits against static discharge and transient excess voltage.

1A VCC (1)1Ā (<u>13)</u> 1Q Ħ (2) B IREXT/CEXT 1B 1 B (3) 1 CLR 1 CLR 1 CX (14) 1 CEXT (4) 1Q сх 1RX/CX (15) 10 1 Q RX/CX ____₹ (9) 20 20 12 24 (5) _{2Q} (10) 2B 2 CEXT 6 2 CLR 2CLR (11) 2CX (6) <u>(12)</u> 2Q сх 2 REXT/CEXT 2 B 2RX/CX (7) RX/CX GND 0 2 A LC13070 July 2001 1/12

PIN CONNECTION AND IEC LOGIC SYMBOLS

INPUT AND OUTPUT EQUIVALENT CIRCUIT



PIN DESCRIPTION

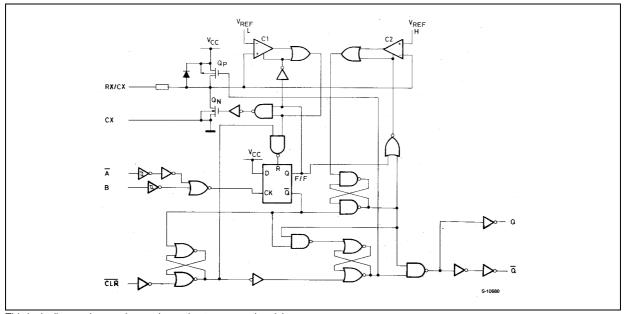
PIN No	SYMBOL	NAME AND FUNCTION
1,9	$1\overline{A}, 2\overline{A}$	Trigger Inputs (Negative Edge Triggered)
2, 10	1B, 2B	Trigger Inputs (Positive Edge Triggered)
3, 11	1 <u>CLR</u> 2 CLR	Direct Reset (Active LOW)
4, 12	1 <u>Q</u> , 2 <u>Q</u>	Outputs (Active Low)
7	2R _X /C _X	External Resistor Capacitor Connection
13, 5	1Q, 2Q	Outputs (Active High)
14, 6	1C _X 2C _X	External Capacitor Connection
15	1R _X /C _X	External Resistor Capacitor Connection
8	GND	Ground (0V)
16	Vcc	Positive Supply Voltage

TRUTH TABLE

	INPUTS		ουτι	PUTS	NOTE			
Ā	В	CLR	Q	Q	NOTE			
7	Н	Н			OUTPUT ENABLE			
Х	L	Н	L	Н	INHIBIT			
Н	Х	Н	L	Н	INHIBIT			
L		Н			OUTPUT ENABLE			
Х	Х	L	L	Н	INHIBIT			

X : Don't Care

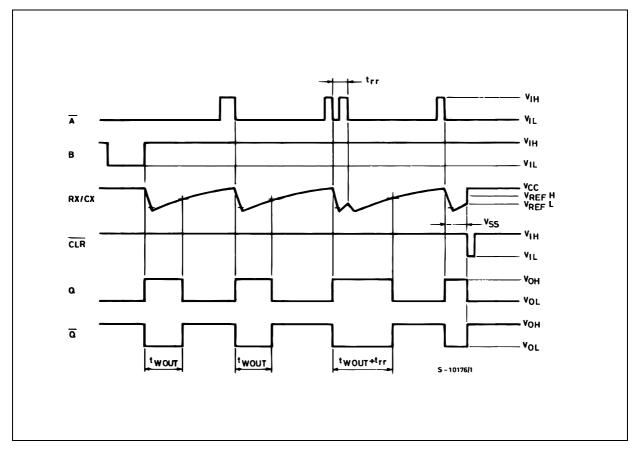
SYSTEM DIAGRAM



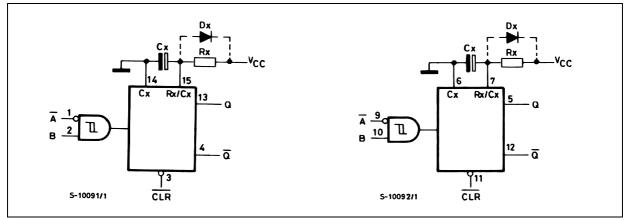
This logic diagram has not be used to estimate propagation delays

2/12

TIMING CHART



BLOCK DIAGRAM



Cx, Rx, Dx are external components.
(2) Dx is a clamping diode.
The external capacitor is charged to Vcc in the stand-by-state, i.e. no trigger. When the supply voltage is turned off Cx is discharged mainly trough an internal parasitic diode(see figures). If Cx is sufficiently large and Vcc decreases rapidly, there will be some possibility of damaging the I.C. with a surge current or latch-up. If the voltage supply filter capacitor is large enough and Vcc decrease slowly, the surge current is automatically limited and damage to the I.C. is avoided. The maximum forward current of the parasitic diode is approximately 20 mA. In cases where Cx is large the time taken for the supply voltage to fall to 0.4 Vcc can be calculated as follows :

 $t_{f} \ge (Vcc - 0.7) \times Cx/20mA$ In cases where t_{f} is too short an external clamping diode is required to protect the I.C. from the surge current.

FUNCTIONAL DESCRIPTION

STAND-BY STATE

The external capacitor,Cx, is fully charged to Vcc in the stand-by state. Hence, before triggering, transistor Qp and Qn (connected to the Rx/Cx node) are both turned-off. The two comparators that control the timing and the two reference voltage sources stop operating. The total supply current is therefore only leakage current.

TRIGGER OPERATION

Triggering occurs when :

1 st) A is "LOW" and B has a falling edge;

2 nd) B is "HIGH" and A has a rising edge;

3 rd) A is "LOW" and B is HIGH and C1 has a rising edge;

After the multivibrator has been retriggered comparator C1 and C2 start operating and Qn is turned on. Cx then discharges through Qn. The voltage at the node R/C external falls.

When it reaches V_{REFL} the output of comparator C1 becomes low. This in turn reset the flip-flop and Qn is turned off.

At this point C1 stops functioning but C2 continues to operate.

The voltage at R/C external begins to rise with a time constant set by the external components Rx, Cx.

Triggering the multivibrator causes Q to go high after internal delay due to the flip-flop and the gate. Q remains high until the voltage at R/C external rises again to V_{REFH} . At this point C2

output goes low and O goes low. C2 stop operating. That means that after triggering when the voltage R/C external returns to V_{REFH} the multivibrator has returned to its MONOSTABLE STATE. In the case where $Rx \cdot Cx$ are large enough and the discharge time of the capacitor and the delay time in the I.C. can be ignored, the width of the output pulse tw (out) is as follows :

$tW(OUT) = Cx \cdot Rx$

RE - TRIGGERED OPERATION

When a second trigger pulse follows the first its effect will depend on the state of the multivibrator. If the capacitor Cx is being charged the voltage level of R/C external falls to V_{REFL} again and Q remains High i.e. the retrigger pulse arrives in a time shorter than the period Rx \cdot Cx seconds, the capacitor charging time constant. If the second trigger pulse is very close to the initial trigger pulse it is ineffective; i.e. the second trigger must arrive in the capacitor discharge cycle to be ineffective; Hence the minimum time for a second trigger to be effective depends on Vcc and Cx

RESET OPERATION

CL is normally high. If CL is low, the trigger is not effective because Q output goes low and trigger control flip-flop is reset.

Also transistor Op is turned on and Cx is charged quickly to Vcc. This means if CL input goes low the IC becomes waiting state both in operating and non operating state.

<u>\</u>

Symbol	Parameter	Value	Unit
V _{CC}	Supply Voltage	-0.5 to +7	V
VI	DC Input Voltage	-0.5 to V _{CC} + 0.5	V
Vo	DC Output Voltage	-0.5 to V _{CC} + 0.5	V
I _{IK}	DC Input Diode Current	± 20	mA
I _{ОК}	DC Output Diode Current	± 20	mA
۱ _۵	DC Output Current	± 25	mA
I _{CC} or I _{GND}	DC V _{CC} or Ground Current	± 50	mA
PD	Power Dissipation	500(*)	mW
T _{stg}	Storage Temperature	-65 to +150	°C
ΤL	Lead Temperature (10 sec)	300	°C

ABSOLUTE MAXIMUM RATINGS

Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation under these conditions is not implied (*) 500mW at 65 °C; derate to 300mW by 10mW/°C from 65°C to 85°C

RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter		Value	Unit
V _{CC}	Supply Voltage		2 to 6	V
VI	Input Voltage		0 to V _{CC}	V
Vo	Output Voltage		0 to V _{CC}	V
T _{op}	Operating Temperature		-55 to 125	°C
	Input Rise and Fall Time	$V_{CC} = 2.0V$	0 to 1000	ns
t _r , t _f		$V_{CC} = 4.5V$	0 to 500	ns
		$V_{CC} = 6.0V$	0 to 400	ns
Сх	External Capacitor	·	NO LIMITATION	pF
Rx	External Resistor	Vcc < 3V	5K to 1M	Ω
IXX		Vcc <u>></u> 3V	1K to 1M	52

The Maximum allowable values of Cx and Rx are a function of leakage of capacitor Cx, the leakage of device and leakage due to the board layout and surface resistance. Susceptibility to externally induced noise may occur for $Rx > 1M\Omega$

DC SPECIFICATIONS

		٦	Test Condition				Value				
Symbol	Parameter	Parameter V _{CC}		т	T _A = 25°C			85°C	-55 to 125°C		Unit
		(Ŭ)		Min.	Тур.	Max.	Min.	Max.	Min.	Max.	
VIH	High Level Input	2.0		1.5			1.5		1.5		
	Voltage	4.5		3.15			3.15		3.15		V
		6.0		4.2			4.2		4.2		
V _{IL}	Low Level Input	2.0				0.5		0.5		0.5	
	Voltage	4.5				1.35		1.35		1.35	V
		6.0				1.8		1.8		1.8	
V _{OH}	High Level Output	2.0	I _O =-20 μΑ	1.9	2.0		1.9		1.9		
	Voltage	4.5	I _O =-20 μA	4.4	4.5		4.4		4.4		
		6.0	I _O =-20 μA	5.9	6.0		5.9		5.9		V
		4.5	I _O =-4.0 mA	4.18	4.31		4.13		4.10		
		6.0	I _O =-5.2 mA	5.68	5.8		5.63		5.60		
V _{OL}	Low Level Output	2.0	I _O =20 μA		0.0	0.1		0.1		0.1	
	Voltage	4.5	I _O =20 μA		0.0	0.1		0.1		0.1	
		6.0	I _O =20 μA		0.0	0.1		0.1		0.1	V
		4.5	l _O =4.0 mA		0.17	0.26		0.33		0.40	
		6.0	I _O =5.2 mA		0.18	0.26		0.33		0.40	
I _I	Input Leakage Current	6.0	$V_{I} = V_{CC} \text{ or } GND$			± 0.1		± 1		± 1	μΑ
Ι _Ι	R/C Terminal Off State Current	6.0	$V_{I} = V_{CC} \text{ or } GND$			± 0.1		± 1		± 1	μΑ
I _{CC}	Quiescent Supply Current	6.0	$V_I = V_{CC}$ or GND			4		40		80	μΑ
I _{CC'}	Active State	2.0	$V_{I} = V_{CC} \text{ or } GND$		45	200		260		320	μA
	Supply Current (1)	4.5	Pin 7 or 15		500	600		780		960	μΑ
		6.0	$V_{IN} = V_{CC}/2$		0.7	1		1.3		1.6	mA

(1) : Per Circuit

M74HC423A

AC ELECTRICAL CHARACTERISTICS ($C_L = 50 \text{ pF}$, Input $t_r = t_f = 6 \text{ns}$)

		Т	est Condition				Value				
Symbol	Parameter	v _{cc}		Т	A = 25°	C	-40 to	85℃	-55 to	125°C	Unit
		(V)		Min.	Тур.	Max.	Min.	Max.	Min.	Max.	
t _{TLH} t _{THL}	Output Transition	2.0			30	75		95			
	Time	4.5			8	15		19			ns
		6.0			7	13		16			
t _{PLH} t _{PHL}	Propagation Delay	2.0			102	210		265			
	Time	4.5			29	42		53			ns
	(Ā, B - Q, Q)	6.0			22	36		45			
t _{PLH} t _{PHL}	Propagation Delay	2.0			68	160		200			
		4.5			20	32		40			ns
	$(\overline{CLR} - Q, \overline{Q})$	6.0			16	27		34			
t _{WOUT}	Output Pulse Width	2.0	0 400 5		1.7						
		4.5	Cx = 100 pF Rx = 10KΩ		1.4						μs
	6.0	10122		1.3							
		2.0	00.4E		10						
		4.5	Cx = 0.1μF Rx = 100KΩ		9.5						ms
		6.0	1001022		9.5						
∆t _{WOUT}	Output Pulse Width Error Between Circuits in Same Package				±1						%
t _{W(H)}	Minimum Pulse	2.0				75		95			
t _{W(L)}	Width	4.5				15		19			ns
		6.0				13		16			
t _{W(L)}	Minimu <u>m P</u> ulse	2.0				75		95			
	Width (CLR)	4.5				15		19			ns
		6.0				13		16			
t _{rr}	Minimum Retrigger	2.0	Cx = 100 pF		325						
	Time	4.5	Cx = 100 pr Rx = 10K Ω		108						ns
		6.0			78						
		2.0	Cx = 0.1μF		5						
		4.5	$Cx = 0.1\mu F$ Rx = 100K Ω		1.4						μs
		6.0	100122		1.2						

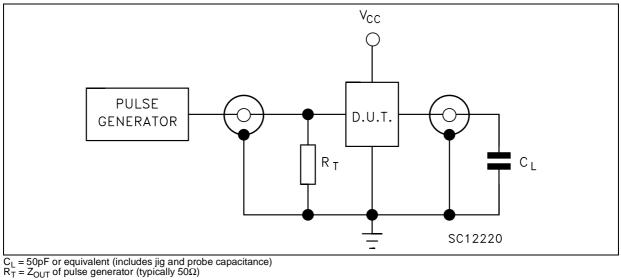
CAPACITIVE CHARACTERISTICS

		Te	Test Condition	Value							
Symbol	Parameter	v _{cc}		т	A = 25°	С	-40 to	85°C	-55 to	125°C	Unit
		(V)		Min.	Тур.	Max.	Min.	Max.	Min.	Max.	
C _{IN}	Input Capacitance	5.0			5	10		10		10	pF
C _{PD}	Power Dissipation Capacitance (note 1)	5.0			160						pF

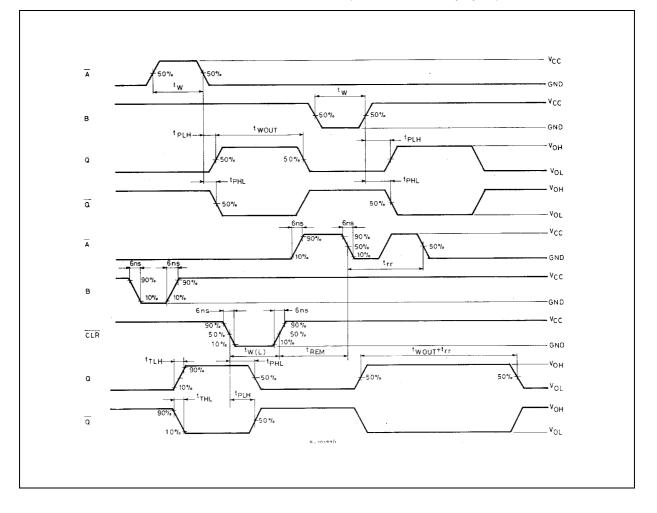
1) C_{PD} is defined as the value of the IC's internal equivalent capacitance which is calculated from the operating current consumption without load. (Refer to Test Circuit). Average operating current can be obtained by the following equation. $I_{CC(opr)} = C_{PD} \times V_{CC} \times f_{IN} + I_{CC}$ ' Duty/100 + Ic/2(per monostable) (I_{cc} ': Active Supply current) (Duty : %)

57

TEST CIRCUIT

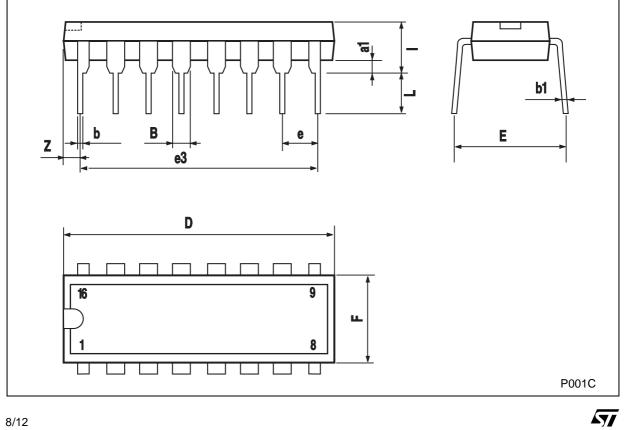


SWITCHING CHARACTERISTICS TEST WAVEFORM(f=1MHz; 50% duty cycle)



		Plastic DIP	-16 (0.25) N	IECHANIC	AL DATA		
DIM.		mm.		inch			
DINI.	MIN.	ТҮР	MAX.	MIN.	TYP.	MAX.	
a1	0.51			0.020			
В	0.77		1.65	0.030		0.065	
b		0.5			0.020		
b1		0.25			0.010		
D			20			0.787	
E		8.5			0.335		
е		2.54			0.100		
e3		17.78			0.700		
F			7.1			0.280	
Ι			5.1			0.201	
L		3.3			0.130		
Z			1.27			0.050	

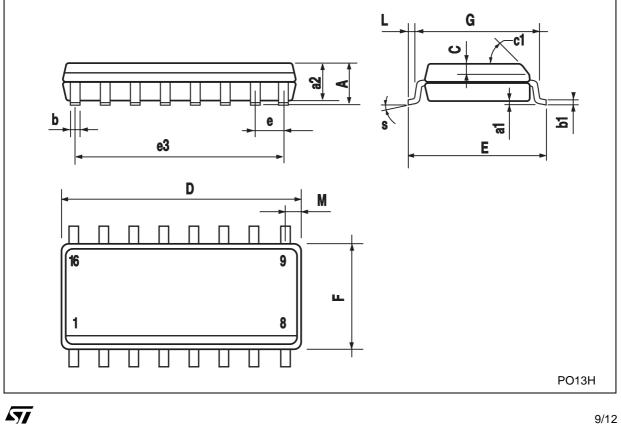
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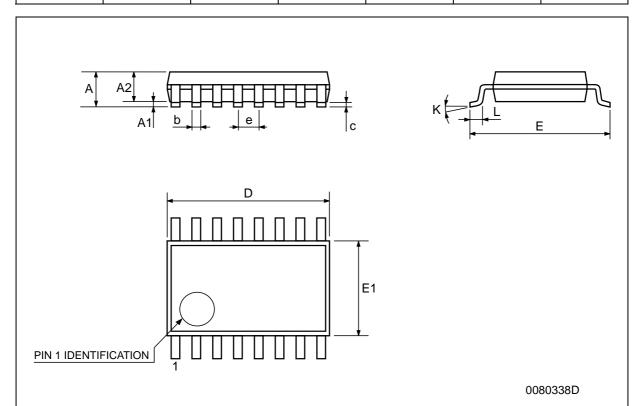
DIM.		mm.		inch					
	MIN.	ТҮР	MAX.	MIN.	TYP.	MAX.			
А			1.75			0.068			
a1	0.1		0.2	0.003		0.007			
a2			1.65			0.064			
b	0.35		0.46	0.013		0.018			
b1	0.19		0.25	0.007		0.010			
С		0.5			0.019				
c1		ł	45°	(typ.)					
D	9.8		10	0.385		0.393			
E	5.8		6.2	0.228		0.244			
е		1.27			0.050				
e3		8.89			0.350				
F	3.8		4.0	0.149		0.157			
G	4.6		5.3	0.181		0.208			
L	0.5		1.27	0.019		0.050			
М			0.62			0.024			





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					inch	
DIM.		mm.	i		Incn	
	MIN.	ТҮР	MAX.	MIN.	TYP.	MAX.
А			1.2			0.047
A1	0.05		0.15	0.002	0.004	0.006
A2	0.8	1	1.05	0.031	0.039	0.041
b	0.19		0.30	0.007		0.012
С	0.09		0.20	0.004		0.0089
D	4.9	5	5.1	0.193	0.197	0.201
Е	6.2	6.4	6.6	0.244	0.252	0.260
E1	4.3	4.4	4.48	0.169	0.173	0.176
е		0.65 BSC			0.0256 BSC	
К	0°		8°	0°		8°
L	0.45	0.60	0.75	0.018	0.024	0.030



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