

## TACHOMETER DRIVE CIRCUIT

### DESCRIPTION

The CS-189 is specifically designed for use with air-core meter movements. The IC (see Block Diagram, Figure 1) comprises charge pump circuitry for frequency-to-voltage conversion, a shunt regulator for stable operation, a function generator, and sine and cosine amplifiers. The buffered sine and cosine outputs will typically sink or source 20mA.

### CHARGE PUMP

The input frequency is buffered through a transistor, then applied to the charge pump for frequency-to-voltage conversion (see Functional Diagram, Figure 2). The charge pump output voltage,  $E\emptyset$ , will range from 2.1V with no input ( $\emptyset=0^\circ$ ) to 7.1V at  $\emptyset=270^\circ$ . The charge that appears on  $C_T$  is reflected to  $C_{OUT}$  through a Norton amplifier. The frequency applied at Pin 10 charges and discharges  $C_T$  through  $R_1$  and  $R_2$ .  $C_{OUT}$  reflects the charge as a voltage across resistor  $R_T$ .

### FUNCTION GENERATOR/SINE AND COSINE AMPLIFIERS

The output waveforms of the sine and cosine amplifiers are derived by On-Chip Amplifier/Comparator circuitry. The various trip points for the circuit (i.e. 90°, 180°, 270°) are determined by an internal resistor divider connected to the voltage regulator. The voltage  $E\emptyset$  is compared to the divider network by the function generator circuitry. Use of an external zener reference at Pin 1 allows both sine and cosine amplifiers to swing positive and negative with respect to this reference. The output magnitudes and directions have the relationship as shown in Figures 3A, 3B and 3C.

NOTE: Pin connections referenced are for the 14L DIP

### BLOCK DIAGRAM

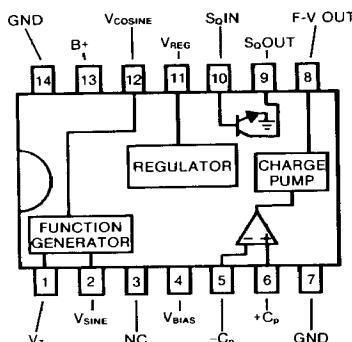


FIGURE 1

Cherry Semiconductor Corporation, 2000 South County Trail, East Greenwich, Rhode Island 02818  
(401) 885-3600



### FEATURES:

- Single supply operation
- On-Chip regulation
- 20mA output drive capability
- Wide temperature range
- Accuracy = .05% / °C

### APPLICATIONS:

- Tachometers
- Speedometer
- F/V conversion
- Windspeed & direction
- Flowmeter control

### PIN CONNECTIONS

#### 20L SOIC WIDE

VZ	1	20	B+
V <sub>SINE</sub>	2	19	V <sub>COS</sub>
V <sub>BIA</sub> S	3	18	V <sub>REG</sub>
GND	4	17	GND
GND	5	16	GND
GND	6	15	GND
GND	7	14	GND
-CP	8	13	SQIN
+CP	9	12	SQOUT
NC	10	11	F-V OUT

#### 14L DIP

Vz	1	14	GND
V <sub>SINE</sub>	2	13	B+
NC	3	12	V <sub>COSINE</sub>
V <sub>BIA</sub> S	4	11	V <sub>REG</sub>
-CP	5	10	SQIN
+CP	6	9	SQOUT
GND	7	8	F-V OUT

(TOP VIEW)

### ABSOLUTE MAXIMUM RATINGS

Operating Temperature .....	-30°C to +100°C	Supply Voltage .....	20V
Storage Temperature .....	-65°C to +150°C	Power Dissipation .....	2W@ 25°C (Derate 16mW/°C above 25°C)

### ELECTRICAL CHARACTERISTICS ( $V_{CC}=13.1V$ , $T_A=25^\circ C$ )

PARAMETER	SYMBOL	PIN NO.	TEST CONDITIONS (SEE TEST CIRCUIT FIG. 1)	MIN.	TYP.	MAX.	UNITS
Supply Current (note 2)	$I_{CC}$	13	$V_{CC}=15V$		54	72	mA
			$V_{CC}=13.1V$		45	60	
			$V_{CC}=11.3V$		38	60	
Regulated Voltage	$V_{REG}$	11	$I_{REG}=4.3mA$	7.8	8.5	9.2	V
			$I_{REG}=0$ to $5mA$		0.10	0.20	
Signal Input Current	$I_{IN}$	10		0.1	2.0	4.0	mA
Saturation Voltage	$V_{SAT}$	9	$I_9=5mA$ , $I_{10}=0.5mA$		0.20	0.40	V
Leakage Current	$I_9$		$V_9=16V$ , $V_{10}=0V$			10	$\mu A$
Input Current	$I_5$	5					
Input Current	$I_6$	6	Zero input to pin 6		1	15	nA
F to V Output	E MIN	8	$V_{10}=0$ (zero input), $\theta=0^\circ$	1.9	2.1	2.3	V
	E MAX		$V_{cos}=0$ (note 1), $\theta=270^\circ$	6.6	7.1	7.6	
Linearity	%	8	$E_\theta$ vs. Frequency $V_{cos}=0$ (note 1), $\theta=270^\circ$			$\pm 1.5$	%
$V_{SINE}$ at $\theta=0^\circ$	$V_{SINE}$	2	$V_{10}=0$ (zero input), $\theta=0^\circ$	-0.35	0	+0.35	V
MAX $V_{SINE} +$			$V_{cos}=0$ (note 1), $\theta=90^\circ$	4.0	4.5	5.0	
MAX $V_{SINE} -$			$V_{cos}=0$ (note 1), $\theta=270^\circ$	-4.0	-4.5	-5.0	
Coil Drive Current	$I_{SINE+}$	2	$V_{cos}=0$ (note 1), $\theta=90^\circ$		20	25	mA
	$I_{SINE-}$		$V_{cos}=0$ (note 1), $\theta=270^\circ$		20	25	
MAX $V_{cos} +$	$V_{cos}$	12	$V_{10}=0$ (zero input), $\theta=0^\circ$	4.0	4.5	5.0	V
MAX $V_{cos} -$			$V_{SINE}=0$ (note 1), $\theta=180^\circ$	-4.0	-4.5	-5.0	
Coil Drive Current	$I_{cos+}$	12	$V_{10}=0$ (zero input), $\theta=0^\circ$		20	25	mA
	$I_{cos-}$		$V_{SINE}=0$ (note 1), $\theta=180^\circ$		20	25	
External Voltage Ref.	$V_z$	1		5.13	5.40	5.67	V

Note 1:  $V_{SINE}$  measured Pin 2 to Pin 1.  $V_{cos}$  measured Pin 12 to Pin 1. All other voltages specified are measured to ground.

Note 2: Max PWR dissipation  $\leq V_{CC} \times I_{CC} - (V_2 I_{SINE} + V_{12} I_{cos})$ .

### FUNCTION GENERATOR OUTPUT ( $\theta$ ): $V_{CC}=13.1V$ , $T_A=25^\circ C$

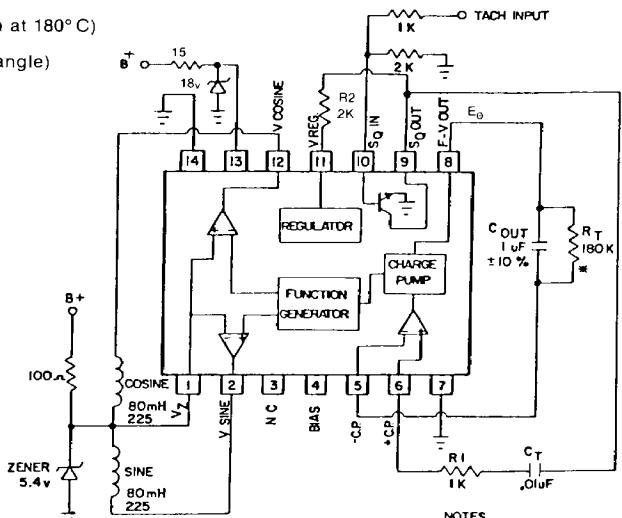
$$\theta = \text{ArcTan} \left( \frac{V_{SINE}}{V_{cos}} \right) \quad (\text{Measured angle after calibration at } 180^\circ C)$$

For  $\theta_A = 45^\circ, 90^\circ, 135^\circ, 180^\circ, 225^\circ, 270^\circ$ , (Desired angle)  
 $(\theta_A - \theta_M) \leq 4.0^\circ$

**TEMPERATURE SENSITIVITY:**  $V_{CC}=13.1V$   
 $\Delta \theta_{MT} = \theta_M$  ( $T=25^\circ C$ ) -  $\theta_M$  ( $-20^\circ C \leq T \leq +85^\circ C$ )  
 $(\Delta \theta_{MT}) \leq 3.5^\circ C$ ,  $-20^\circ C \leq T \leq +85^\circ C$

**VOLTAGE SENSITIVITY:**  $T_A=25^\circ C$   
 $\Delta \theta_{MV} = \theta_M$  ( $V_{CC}=13.1V$ ) -  $\theta_M$  ( $11.3V \leq V_{CC} \leq 15V$ )  
 $(\Delta \theta_{MV}) \leq 2^\circ$ ,  $11.3V \leq V_{CC} \leq 15V$

FIGURE 2  
FUNCTIONAL DIAGRAM  
(TEST CIRCUIT)



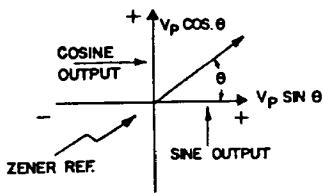


FIGURE 3A

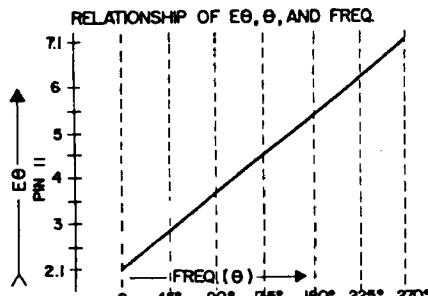


FIGURE 3B

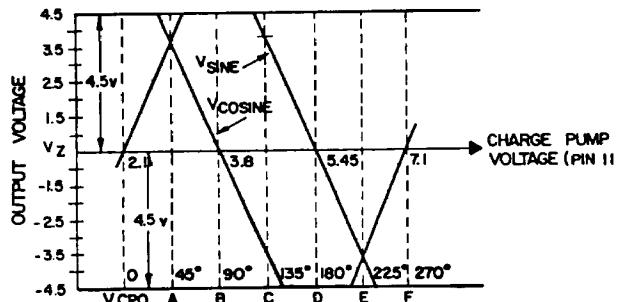


FIGURE 3C

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#### TACHOMETER APPLICATION

$$\frac{\text{RPM} \times \# \text{ OF CYL}}{2} = \text{Frequency}$$

$$E (\text{Pin 8}) = 2.1 + \text{Frequency} \times C_T \times R_T (V_{REG} = .7)$$

The above equations were used in calculating the following values, where  $E=7.1\text{V}$  at  $=270^\circ$  and  $C_T=.01\text{F}$ .

4 cylinder: Freq=200Hz,  $R_T=320\text{K}$

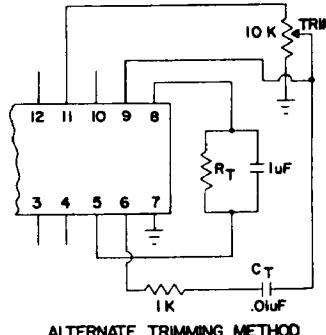
6 cylinder: Freq=300Hz,  $R_T=220\text{K}$

8 cylinder: Freq=400Hz,  $R_T=150\text{K}$

Typical values shown above apply to a nominal value of  $V_{REG}$  of 8.5 volts. It must be realized that trimming of  $R_T$  will be necessary to compensate for variations in regulator voltage from one unit to another.

An alternative to this adjustment is to replace  $R_T$  with a potentiometer, as shown in Figure 4.

Partial schematic shown in Figure 5 represents one method for use with DC applications instead of frequency.



ALTERNATE TRIMMING METHOD

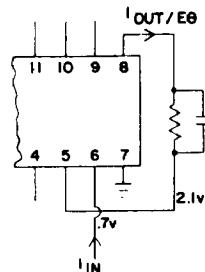


FIGURE 5

#### ORDERING INFORMATION

PART NUMBER	DESCRIPTION
CS-189DW	20 Lead SO Wide
CS-189N	14 Lead PDIP