ISO-9001 CERTIFIED BY DSCC



HIGH POWER OPERATIONAL AMPLIFIER

M.S.KENNEDY CORP.

4707 Dey Road Liverpool, N.Y. 13088

FEATURES:

- Operates In Class AB Or Class C Mode
- Low Cost
- High Voltage Operation : 150V
- Low Quiescent Current := 4.0 mA Typ. In Class "C"
- Mode
- High Output Current : 5A Min.
- No Second Breakdown
- High Speed : 27V/µS Typ.
- External Compensation For Optimum Gain-Bandwidth

DESCRIPTION:

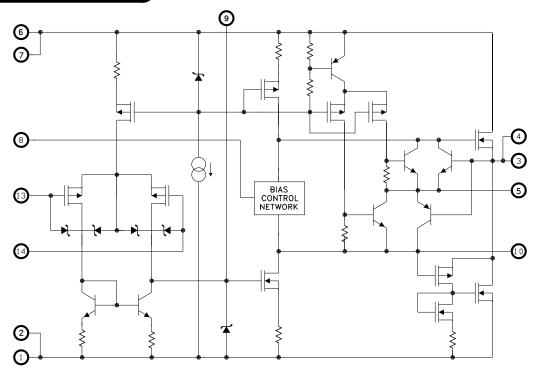


(315) 701-6751



The MSK 153 is a high power monolithic MOSFET operational amplifier ideally suited for high power amplification and magnetic deflection applications. With a total supply voltage rating of 150 volts and 5A of available output current, the MSK 153 is also an excellent low cost choice for motor drive circuits. The MOSFET output frees the MSK 153 from secondary breakdown limitations and power dissipation is kept to a minimum with a typical quiescent current rating of 4.0 mA in class "C" mode. Power saving class "C" mode is enabled by the user externally. The MSK 153 is packaged in a hermetically sealed 14 pin power dip which has two external compensation pins.

EQUIVALENT SCHEMATIC



TYPICAL APPLICATIONS

- PA Audio
 Magnetic Deflection
- Motor Drive
- Noise Cancellation

PIN-OUT INFORMATION

- 1 -Vcc
- 2 -Vcc
- 3 Output Drive
- 4 Output Drive
- 5 Current Sense
- 6 + Vcc
- 7 + Vcc

- 13 Inverting Input
- 12 N/C
- 11 N/C
- 10 Comp 2
- 9 Comp 1
 - 8 Quiescent Current Adjust

ABSOLUTE MAXIMUM RATINGS

| Vcc ② | Total Supply Voltage | .150V |
|--------|--------------------------------|---------|
| ± Іоит | Output Current (within S.O.A.) | . ±5A |
| VIND | Input Voltage (Differential) | _±16V |
| Vin | Input Voltage (Common Mode) | _ ± Vcc |
| ΤJ | Junction Temperature | .150°C |

| Тsт | Storage Temperature65°C to +150°C |
|-----|-----------------------------------|
| TLD | Lead Temperature |
| Tc | Case Operating Temperature |
| | (MSK153B) |
| | (MSK153) |
| Rтн | Thermal Resistance (DC) |
| | Junction to Case 1.4°C/W |

ELECTRICAL SPECIFICATIONS

| Devementer | Test Conditions ① | Group A | MSK153B | | В | MSK153 | | | |
|---|-----------------------|----------|---------|------------------|------|--------|------------------|------|-----------|
| Parameter | | Subgroup | Min. | Тур. | Max. | Min. | Тур. | Max. | Units |
| STATIC | | | | | | | | | |
| Supply Voltage Range ② ④ | | - | ±15 | ±50 | ±75 | ±15 | ±50 | ±75 | V |
| | | 1 | - | ±30 | ±50 | - | ± 30 | ±50 | mA |
| Quiescent Current | VIN=OV | 2 | - | ±20 | ±45 | - | - | - | mA |
| | | 3 | - | ± 35 | ±65 | - | - | - | mA |
| Quiescent Current (Class 'C') | VIN=0V | 1 | - | ± 4 | ± 6 | | ± 4 | ± 6 | mA |
| INPUT | | | | | | | | | |
| Offset Voltage | VIN=0V | 1 | - | ±5 | ±10 | - | ±5 | ±15 | mV |
| Offset Voltage Drift ④ | VIN=0V | 2,3 | - | ±10 | ±50 | - | ±10 | - | µV/°C |
| Offset Voltage vs ±Vcc ④ | VIN=0V | 1 | - | ±8 | ±15 | - | ±8 | ±15 | $\mu V/V$ |
| Input Bias Current ④ | Vcm=0V | 1,3 | - | ±20 | ±100 | - | ±20 | ±100 | pА |
| | | 2 | - | - | ±50 | - | - | - | nA |
| Input Impedance ④ | (DC) | - | - | 10 ¹¹ | - | - | 10 ¹¹ | - | Ω |
| Input Capacitance ④ | | - | - | 5 | - | - | 5 | - | pF |
| Common Mode Rejection ④ | $V_{CM} = \pm 30 VDC$ | - | 90 | 106 | - | 90 | 106 | - | dB |
| Noise | 10KHz BW | - | - | 10 | - | - | 10 | - | μVrms |
| OUTPUT | | | | | | | | | |
| Output Voltage Swing | $IOUT = \pm 5A Peak$ | 4 | ±40 | ±42 | - | ±40 | ±42 | - | V |
| Output Current | Vout = MAX | 4 | ±5 | ±5.5 | - | ±5 | ± 5.5 | - | А |
| Power Bandwidth ④ | Vout = 80Vpp | - | - | 66 | - | - | 66 | - | KHz |
| Settling Time to 0.1% ③④ 10V Step | | - | - | 2 | - | - | 2 | - | μS |
| Capacitive Load (4) $Av = + 1V/V$ Cc = 68pF | | - | 10 | - | - | 10 | - | - | nF |
| TRANSFER CHARACTERISTICS | | | | | | | | | |
| Slew Rate Cc = Open | | 4 | 20 | 27 | - | 20 | 27 | - | V/µS |
| Open Loop Voltage Gain ④ F = 15Hz | | 4 | 94 | 106 | - | 94 | 106 | - | dB |

NOTES:

- (1) Unless otherwise noted Cc = 10pF, Rc = $1.0K\Omega$, $\pm Vcc = \pm 50VDC$. (2) Derate maximum supply voltage $0.5V/^{\circ}C$ below Tc = $\pm 25^{\circ}C$. No derating is needed above Tc = $25^{\circ}C$. (3) Av = -10V/V measured in false summing junction circuit. (4) Devices shall be capable of meeting the parameter, but need not be tested. Typical parameters are for reference only.
- (5) Industrial grade devices shall be tested to subgroups 1 and 4 unless otherwise requested.
- (a) Midistrial grade devices shall be tested to subgroups 1 and 4 diffess otherwise requirements (b) Military grade devices ('B' suffix) shall be 100% tested to subgroups 1,2,3 and 4. (c) Subgroup 5 and 6 testing available upon request. (c) Subgroup 1,4 $Tc = +25 \circ C$ Subgroup 2,5 $Tc = +125 \circ C$ Subgroup 2,5 $Tc = +125 \circ C$

Subgroup 3,6 $T_A = -55 \,^{\circ}C$

APPLICATION NOTES

CURRENT LIMIT (SEE TYPICAL CONNECTION DIAGRAM)

A value of current limit resistance can be calculated as follows:

Where:

RCL is the current limit resistor value

ICL is the current limit

 $0.05 \ * \ \text{ICL}$ is the voltage dropped in the current limit path across internal impedances other than the actual current limit resistor

0.83 volts is the voltage drop that must be developed across the current limit connections to activate the current limit circuit

The maximum practical value of current limit resistance is 16 ohms. The current limit resistor will decrease available output voltage swing in the following manner:

VR is the reduction in output voltage swing due to the current limit resistor. It is recommended the user limit output current to a value as close to the required output current as possible, without clipping output voltage swing. Current limit will vary with case temperature. Refer to the typical performance curves to predict current limit drift. If current limit is not required replace the resistor with a short.

STABILITY

The MSK 153 has sufficient phase margin when compensated for unity gain to be stable with capacitive loads of at least 10nF. However, it is recommended that the parallel sum of the input and feedback resistor be 1000 ohms or less for closed loop gains of ten or less to minimize phase shift caused by the R-C network formed by the input resistor, feedback resistor and input capacitance. The user can tailor the performance of the MSK 153 to their application using the external compensation pins. The graphs of small signal gain and phase as well as the graphs of slew rate and power response demonstrate the effect of various forms of compensation. The compensation capacitor must be rated at 150 volts working voltage if maximum power supply voltages are used. The compensation resistor and capacitor lead lengths must be kept as short as possible to minimize spurious oscillations. A high quality NPO capacitor is recommended for the compensation capacitor. An effective method of checking amplifier stability is to apply the worst case capacitive load to the output of the amplifier and drive a small signal square wave across it. If overshoot is less than 25%, the system will typically be stable.

INPUT PROTECTION

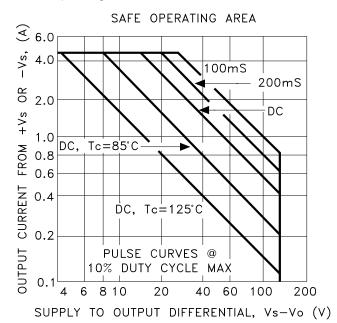
Input protection circuitry within the MSK 153 will clip differential input voltages greater than 16 volts. The inputs are also protected against common mode voltages up to the supply rails as well as static discharge. There are 300 ohm current limiting resistors in series with each input. These resistors may become damaged in the event the input overload is capable of driving currents above 1mA. If severe overload conditions are expected, external input current limiting resistors are recommended.

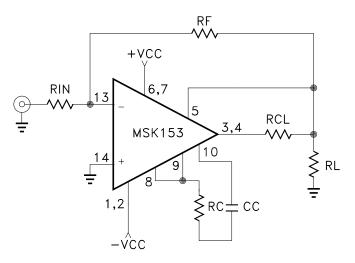
SAFE OPERATING AREA (SOA)

The MOSFET output stage of this power operational amplifier has two distinct limitations:

- 1. The current handling capability of the die metallization.
- 2. The junction temperature of the output MOSFET's.

NOTE: The output stage is protected against transient flyback. However, for protection against sustained, high energy flyback, external fast-recovery reverse biased diodes should be connected from the output to ground.



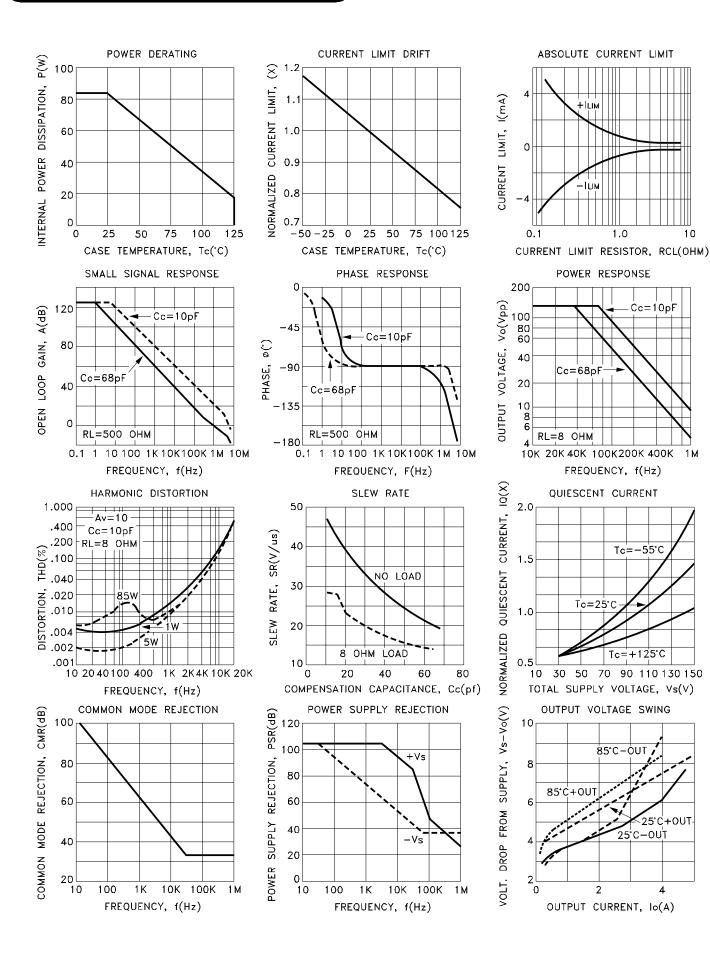


TYPICAL CONNECTION DIAGRAM

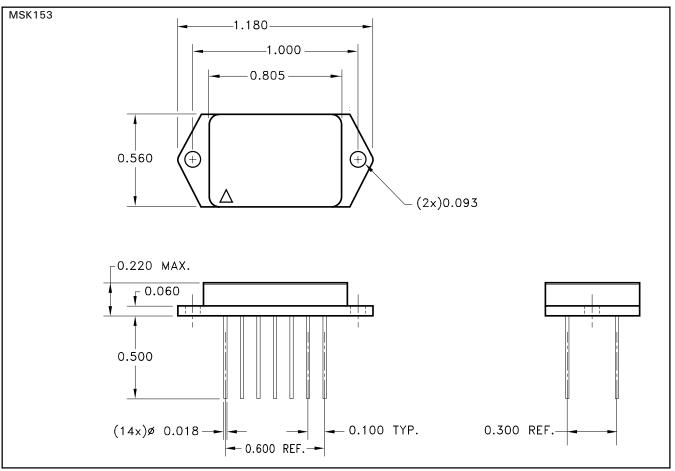
CLASS "C" MODE

The user can enable class "C" mode by simply connecting pin 8 to pin 9. This connection will disable the bias control network in the output stage. Quiescent current will drop to ± 4 mA typically and a small amount of crossover distortion will appear on the output waveform. This mode of operation is recommended for all low frequency applications where distortion is not a critical specification and quiescent power dissipation must be minimized. Disconnecting the pin 8 to pin 9 short will enable the bias control network and the output will be biased in class AB mode again. Pin 8 should be left unconnected when using class AB mode.(Do not ground or pull high)

TYPICAL PERFORMANCE CURVES



MECHANICAL SPECIFICATIONS



 $\label{eq:estimate} \begin{array}{l} \text{ESD TRIANGLE INDICATES PIN 1.} \\ \text{ALL DIMENSIONS ARE } \pm 0.010 \text{ INCHES UNLESS OTHERWISE LABELED.} \end{array}$

ORDERING INFORMATION

| Part Number | Screening Level | | | | |
|----------------|------------------------|--|--|--|--|
| MSK153 | Industrial | | | | |
| MSK153B | Military-Mil-PRF-38534 | | | | |

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