

# ADS-115, ADS-116 10-BIT, 1 MHz SAMPLING A/D CONVERTERS

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#### **FEATURES**

- 10-Bit resolution
- · 1 MHz throughput
- 15 Megohm input impedance
- Includes fast Sample/Hold amplifier
- 3-State output, TTL and CMOS compatible

### GENERAL DESCRIPTION

DATEL's ADS-115 and ADS-116 combine a tracking Sample/Hold amplifier and a high-speed, 10- bit Analog-to-Digital converter, all in one small device. The ADS-115 and ADS-116 provide digitized outputs of sinusoidal signals at up to one milion samples per second. By combining both the S/H and A/D circuits in one device, critical layout and thermal factors are overcome. This offers stable, high-bandwidth performance with negligible degradation over time.

The ADS-115 accepts unipolar inputs from 0 to +10 volts and the ADS-116 has a ±10 volt input range. Both models are offered in commercial (0 °C to +70 °C) and military (-55 °C to +125 °C) temperature ranges. For 12-bit applications, users should consider Datel's ADS-105/106 or the SHM-45/ADC-500 pair.

Linearity error is  $\pm 1/2$  LSB (0 °C to +70 °C) and the maximum gain temperature coefficient is  $\pm 35$  ppm of full scale range per degree Celsius. Maximum power required is 2.7 watts using  $\pm 15$  and  $\pm 5$  volt dc supplies.

The ADS-115/116 may be sampled up to 1 MHz. The input bandwidth of the S/H will accept up to 60 KHz (10V pk-pk) with -60 dB typical harmonic distortion.

These miniature hybrid sampling A/D converters are ideal for fast data acquisition and control systems, array processing, DSP applications, imaging, medical scanning, signal processing, acoustic, resonance, and vibration analyzers.

For high reliability versions of either the ADS-115 or the ADS-116, contact the factory.

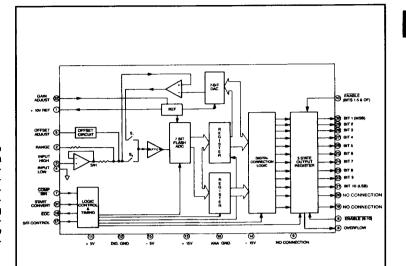
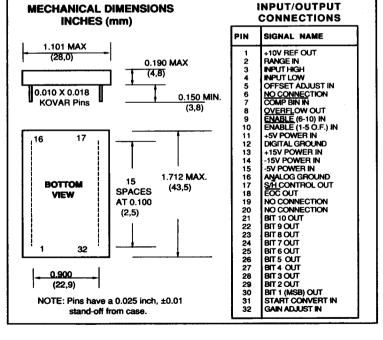


Figure 1. ADS-115, ADS-116 Simplified Block Diagram



# ABSOLUTE MAXIMUM RATINGS

PARAMETERS	LIMITS	UNITS
+15V Supply (Pin 13)	0 to +18	Volts dc
-15V Supply (Pin 14)	0 to -18	Volts dc
+5V Supply(Pin 11)	-0.5 to +7	Volts dc
-5V Supply (Pin 15)	+0.5 to -7	Volts do
Digital inputs	-0.3 to +6	Volts dc
(Pins 7, 9, 10, and 31)		
Analog input	-15 to +15	Volts dc
Lead temp. (10 sec.)	300	° C max.

## **FUNCTIONAL SPECIFICATIONS**

The following specifications apply over the operating temperature range and power supply range unless otherwise indicated.

INPUTS	MIN.	TYP.	MAX.	UNITS
Analog Signal Range ADS-115 ADS-116	- 1	0 to +10 ±10	<u>-</u>	Volts Volts
Input Impedance Resistance Capacitance	5	15 5	_ 7	M Ohms pF
input Bias Current	-	±20	±500	nA
Logic Levels: Logic 1 Logic 0	2.0 –	- -	_ 0.8	Volts Volts
Logic Loading: Logic 1 Logic 0	-	: -	2.5 -100	μ <b>Α</b> μ <b>Α</b>
SAMPLE/HOLD SPECIFICATIONS	MIN.	TYP.	MAX.	UNITS
Slew Rate Aperture Delay Time Aperture Uncertainty	1-1	90 20	<del>-</del> -	V/μSec. nSec.
(Jitter)	_	±100	_	pSec.
S/H Acquisition Time to 0.048% (10V step)	-	-	500	nSec.
Sinusoidal Input	-	-	395	nSec.

# **APPLICATIONS**

- High-speed Data Acquisition and Control Systems
- Array Processing, Vibration and Resonance/ transient Analysis
- · Medical Imaging and Scanning
- · Communications Signal Processing
- Spectrum and Noise Analyzers
- · Video Processing Systems

PERFORMANCE	MIN.	TYP.	MAX.	UNITS
Integral Nonlinearity				
+25 °C 0 °C to +70 °C	-	_	±1/2 ±1/2	LSB LSB
-55 °C to +125 °C	-	-	±1	LSB
Integral Nonlinearity Tempco	_	_	±10	ppm/°C
Differential Nonlinearity				:
+25 °C 0 °C to +70 °C	_	-	±1/2 ±1/2	LSB LSB
-55 °C to +125 °C	<del>-</del> .	±1/2	±3/4	LSB
Differential Nonlinearity Tempco	_	_	±7.5	ppm/ °C
Full-Scale Absolute				
Accuracy +25 °C	_	±2	±6	LSB
0 °C to +70 °C -55 °C to +125 °C	-	±4 ±8	±9 ±15	LSB LSB
		10	1 113	LSB
ADS-115 Unipolar Zero Error, +25 °C Unipolar Zero	_	±1/4	±1	LSB
Error Tempco		±13	±25	ppm/°C
	-	-''	123	ppin/ C
ADS-116 Bipolar Offset	·	1		
Error, +25 °C Bipolar Offset	-	±2	±5	LSB
Error Tempco	-	±17.5	±35	ppm/ °C
Bipolar Zero Error, +25 °C	_	±1/4	±1	LSB
Bipolar Zero Error Tempco		±13	±25	ppm/°C
Gain Error	_	±2	±5	LSB
Gain Error Tempco		±17.5	±35	ľ
·	-	±17.5		ppm/°C
No missing codes (For 12 binary bits)	Guara	anteed o	ver operatin	g temperature
OUTPUTS	MIN.	TYP.	MAX.	UNITS
Logic Levels: Logic 1 Logic 0	2.4	1 1	 0.4	Volts Volts
Logic Loading: Logic 1 Logic 0	1.1	-	- 160 6.4	μ <b>Α</b> m <b>A</b>
Internal Reference: +Voltage, +25° C Tempco External current	9.98 - -	10 ±5 -	10.02 ±30 1.5	Volts dc ppm/ °C mA
Output Coding: ADS-115 (Pin 7 High) (Pin 7 Low)			Straight bin plementary	
ADS-116 (Pin 7 High) (Pin 7 Low)	(		Offset bina mentary of	ry fset binary

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DYNAMIC PERFORMANCE	MIN.	TYP.	MAX.	UNITS
Conversion Rate:				
+25 °C	1 1	-		MHz ·
0 °C to +70 °C	1	_	-	MHz
-55 °C to +125 °C	1 1	-	-	MHz
A/D Conversion Time:				
0 °C to +70 °C	- 1	-	480	nSec.
-55 °C to +125 °C	-	- '	510	nSec.
Total Harmonic				
Distortion:	1 1			
DC to 110 KHz,	1 1			
(Vin = <5V pk-pk)	1 1		1	
+25 °C	-55	-60	- 1	dB
-55 °C to +125 °C	-52	-56	- 1	dB
DC to 60 KHz,	1 1			
(Vin = 10V pk-pk)				-10
+25 °C	-55	-60	- 1	dB
-55 °C to +125 °C	-52	-56	-	dB
DC to 30 KHz,				
(Vin = 20V pk-pk)			ŀ	dB
+25 °C	-55	-60	-	dB dB
-55 °C to +125 °C	-52	-56	-	aв

#### TECHNICAL NOTES

- Use external potentiometers to remove system errors or to reduce any small initial errors to zero. Use a 20K trimming potentiometer for gain adjustment with the wiper tied to pin 32 (ground pin 32 for operation without adjustments). Use a 20K trimming potentiometer with the wiper tied to pin 5 for zero/offset adjustment (leave pin 5 open for operation without adjustment).
- Rated performance requires using good high frequency circuit board layout techniques. The analog and digital grounds are not connected internally. Avoid groundrelated problems by connecting the digital and analog grounds to one point, the ground plane beneath the converter.

Do not connect these grounds together at the power supply terminals when the power supplies are located some distance from the ground plane. Due to the inductance and resistance of the power supply return paths, return the analog and digital ground separately to the power supplies. This prevents contamination of the analog ground by noisy digital ground currents.

- 3. Bypass all the analog and digital supplies to ground with a 4.7μF, 25V tantalum electrolytic capacitor in parallel with a 0.1μF ceramic capacitor. Bypass the +10V reference (pin 1) to ground (pin 16) also using a 4.7μF, 25V capacitor. The -5V dc supply is treated as an analog supply and analog ground (pin 16) should be treated as its return path for decoupling purposes.
- 4. Obtain straight binary/offset binary output coding by tying the COMP BIN signal (pin 7) to +5V dc or leaving it open. The device has an internal pull-up resistor on this pin. To obtain complementary binary or complementary offset binary output coding, tie the COMP BIN pin to ground. The COMP BIN signal is compatible to CMOS/TTL logic levels for those users desiring logic control of this function.

POWER	MIN.	TYP.	MAX.	UNITS
REQUIREMENTS				
Power Supply Range				
+15V dc Supply	+14.25	+15	+15.75	Volts dc
-15V dc Supply	-14.25	-15	-15.75	Volts dc
+5V dc Supply	+4.75	+5	+5.25	Volts dc
-5V dc Supply	-4.75	-5	-5.25	Volts dc
Supply Current				İ
+15V Supply	-	+40	+54	mA
-15V Supply	_	-30	-40	mA
+5V Supply *	_	+56	+90	mA
-5V Supply	-	-173	-210	mA
Power Dissipation	_	2.2	2.7	Watts
Supply Rejection	-	-	±0.01	%FSR/%V
PHYSICAL/ Environmental	MIN.	TYP.	MAX.	UNITS
ENVIRORMENTAL	MAIN.	IIF.	MAA.	01113
Operating Tempera- ture Range				
MC Models	0	_	+70	°C
MM Models	-55	_	+125	°C
Storage Tempera-				
ture Range	-65	_	+150	∘c
Weight	_	l –	0.42(12)	oz.(gram)
Package Type		<del>'                                    </del>		· · · · · · · · · · · · · · · ·
			cally sealed c	eramic DIP
Pin Type	0.010	x 0.018	inch Kovar	

- \* + 5V power usage at 1 TTL logic loading per data output bit.
- An overflow signal, pin 8, indicates when analog input signals are below or above the desired full-scale range. The overflow pin also has a three-state output and is enabled by pin 10 (Enable bits 1-5 & O.F.).
- The S/H Control signal, pin 17, goes low following the rising edge of a start convert pulse and high 30 nanoseconds minimum before EOC goes low. This indicates that the converter can accept a new analog input.
- Full-scale absolute accuracy refers to the unadjusted performance of the ADS-115/116. These figures may be improved substantially using external trim circuits.

#### THEORY OF OPERATION

This theory of operation describes the ADS-115/116's function in conjunction with its internal Sample/Hold amplifer for digitizing sinusoidal signals. The ADS-115/116 employs a subranging A/D conversion architecture with digital error correction. Also known as a two-step conversion method, this technique uses a single 7-bit flash A/D converter twice in the conversion process to yield a final resolution of 10 bits. Refer to the connection diagram, the block diagram, and the timing diagram as needed.

The ADS-115/116 guarantees a 1 MHz throughput rate when the START CONVERT pulse of 50 nanoseconds is provided at a 1 MHz rate. The 1 MHz rate is based upon sinusoidal input frequencies up to those specified in the harmonic distortion specifications. The acquisition time for pulse or dc level signals is longer and listed under the acquisition specifications (10V step).

The ADS-115/116 is in the sample mode when the S/H CONTROL pin is high (S/H is in high-state on power-up). The START CONVERT pulse should be given at a time delay equal to the desired acquisition time minus the 10 nanosecond delay from START CONVERT high to S/H CONTROL low. This

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## THEORY OF OPERATION (Cont.)

assures the sample-hold has the minimum required acquisition time for the particular application mode. Sinusoidal inputs being digitized by using a repetitive START CONVERT pulse will automatically give the appropriate acquisition time when continuously sampling.

Upon going into the hold mode, there will be a 75 nanosecond delay before EOC goes high and the A/D conversion begins. This consists of the remaining 40 nanoseconds of the START CONVERT (10 nanoseconds is part of the acquisition time) and a 35 nanosecond maximum delay from \$TART CONVERT low to EOC high. The hold mode settling time and input settling time requirements required are met by observing this timing.

After conversion is initiated, switch S1 of the ADC closes and S2 opens. The analog input, having been configured for the appropriate input range, is buffered and then digitized by the 7-bit flash ADC to determine the seven most significant bits. The seven bits of data are then stored in a register and provided to the input of a 7-bit digital-to-analog converter. This DAC has 13 bits of linearity.

When the first pass finishes, S2 closes and S1 opens. The output of the DAC is then subtracted from the analog input. The result is a voltage difference between the first 7-bit digitization and the analog input. This voltage difference is amplified and converted by the 7-bit ADC. The result of this second conversion is then latched to determine the least 7 significant bits. The outputs from the two registers are then combined by the digital correction logic to produce a 10-bit word. EOC goes low indicating the conversion is complete, and the output passes to three-state output buffers.

Once the second step of the flash ADC is finished, the analog input can change <u>even</u> though the conversion cycle has not been completed (EOC going low). The S/H Control output goes high shortly before EOC goes low, indicating that the Sample/Hold is back sampling the input. This feature improves the overall throughput of the ADS-115/116.

Data from the previous conversion is valid up to 275 nanoseconds after the falling edge of the START CONVERT pulse. Data from the new conversion is valid a minimum of 25 nanoseconds before the EOC goes low and valid up to 275 nanoseconds after the falling edge of the next START CONVERT pulse. There is a 10 nanosecond maximum delay after the three-state output buffers are enabled before the data is valid at the device output.

The overall throughput of the ADS-115/116 for sinusoidal in-

puts consists of 355 nanoseconds from EOC low to START CONVERT high (logic delays and S/H control going low 30 nanoseconds earlier included); 50 nanoseconds for the hold and input settling time, and 500 nsec for the conversion process. The Sample/Hold control pin saves 30 nanoseconds by allowing the S/H to return to the sample mode 30 nanoseconds before the conversion is complete.

Combining the A/D and S/H in one device guarantees a total throughput of 1.0 microsecond maximum over the -55 °C to +125 °C temperature range for the complete system or a throughput rate of 1 MHz. Retriggering the START CONVERT pulse before EOC goes low will not initiate a new conversion.

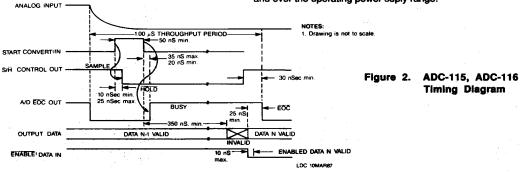
The practical results of this means that single-channel (non-multiplexed) full-scale inputs with spectral content not exceeding 60 KHz for 10V peak-peak signals may be digitized up to a 1 MHz rate. This will give about 16 data samples per cycle to 10-bit accuracy and linearity. This type of resolution is ideal for capture of signals with a broad spectral content of 60 KHz and below (10V peak-to-peak signals). Because many samples may be taken in a short sampling interval, the ADS-115/116 are ideal for computer-aided spectral analysis of a complex-frequency signal. The 10-bit performance is ideal for larger fast fourier transform sizes of 256 to 1024 points by reducing frequency binning resolution noise.

For multi-channel multiplexed signals, the very high multiplexing rate allows larger numbers of channels while still retaining moderate bandwidth per channel. Users requiring higher input bandwidth or faster acquisition times should review Datel's ADS-21 or ADS-22 Sampling A/D converters.

Table 1. ADC-115, ADC-116 Timing Specifications

TIMING	MIN.	TYP.	MAX.	UNITS
Start Convert Pulse				
Width:	50	-	_	nSec.
Start Convert Low	ł			
to EOC High Delay	20	- 1	35	nSec.
Start Convert Low	1			
to Previous Data		· ·		
Invatid	275	'-	_	nSec.
Data Valid Before				
EOC Goes Low	25	-	-	nSec.
Enable to Output	1.14			
Data Valid Delay	-	-	10	nSec.
EOC Low to Start	355	t – I	l . <del>-</del>	nSec.
Convert High (Sinusoidal Inputs)	•			

Note: Table 1 applies over the operating temperature range and over the operating power suply range.



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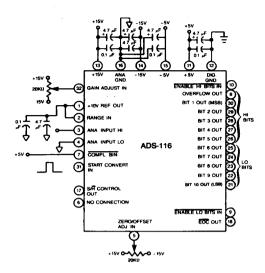


Figure 3. ADS-116 Typical External Connections, ± 10V dc

#### SYSTEM CALIBRATION PROCEDURE

Remove system errors or the small initial errors as follows:

Connect the converter per Figure 3 and Table 2 for the appropriate full-scale input range (FSR). The data outputs should be connected to LED's to observe the resulting data values.

Table 2. Input Connections

INPUT VOLTAGE RANGE	INPUT PIN	JUMP PIN 2 TO PIN:
0 to +10V dc	3	No connection
±10V dc	3	1 (+10V Ref.)

Apply a pulse of 50 nanoseconds minimum to the START CON-VERT input (pin 31) at a rate of 500 KHz. This rate is chosen to reduce flicker if LED's are used on the outputs for calibration purposes.

#### 2. Zero Adjustments:

Apply a precision voltage reference source between the analog input (pin 3) and analog ground (pin 16). Use a very low-noise signal source for accurate calibration.

Adjust the output of the reference source per Tables 3 and 4 for the unipolar zero adjustment (+1/2 LSB) or the bipolar zero adjustment (zero + 1/2 LSB) for the appropriate full scale range. For unipolar operation, adjust the zero trimming potentiometer so that the output code flickers equally between 00 0000 0000 and 00 0000 0001 with the COMP BIN (pin 7) tied high or between 11 1111 1111 and 11 1111 1110 with COMP BIN tied low.

For bipolar operation, adjust the potentiometer such that the code flickers equally between 10 0000 0000 and 10 0000 0001 with the COMP BIN tied high or between 01 1111 1111 and 01 1111 1110 with COMP BIN tied low.

## 3. Full Scale Adjustment:

Set the output of the voltage reference used in step 2 to the value shown in the Tables 3 and 4 for the unipolar or bipolar gain adjustment (+F.S. -1 1/2 LSB) for the appropriate FSR. Adjust the gain trimming potentiometer so that the output code flickers equally between 11 1111 1110 and 11 1111 1111 1111 for COMP BIN (pin 7) tied high or between 00 0000 0001 and 00 0000 0000 for COMP BIN tied low.

To confirm proper operation of the device, vary the precision reference voltage source to obtain the output coding listed in the Tables 5 and 6.

Table 3. Zero and Gain Adjust for Unipolar Use

RANGE	ZERO ADJUST	GAIN ADJUST
UNIPOLAR FSR	(+1/2 LSB)	(+FS -1 1/2 LSB)
0 to +10V	+4.88 mV dc	+9.9927V dc

Table 4. Zero and Gain Adjust for Bipolar Use

RANGE	ZERO ADJUST	GAIN ADJUST
BIPOLAR FSR	(+1/2 LSB)	(+FS -1 1/2 LSB)
±10V	+9.77 mV	+9.9707 V

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Table 5. Output Coding for Uniplar Operation

UNIPOLAR	IIPOLAR INPUT RANGE OUTPUT CODING				
SCALE	Straight	Binary	Complen Straight		
	0 to +10V	MSB	LSB	MSB	LSB
+FS - 1 LSB	+9.99023V	1111 111	1 11	0000 0000	00
7/8 FS	+8.750V	1110 000	0 00	0001 111	1 11
3/4 FS	+7.500V	1100 000	0 00	0011 111	1 11
1/2 FS	+5.000V	1000 000	0 00	0111 111	1 11
1/4 FS	+2.500V	0100 000	00 0	1011 111	1 11
1/8 FS	+1.250V	0.010 000	0 00	1101 111	1 11
1 LSB	√+0.00977V	0000 000	0 00	* 1111 111	1 11
0	0.0000V	0000 0000	0 00	1111 111	1 11

Table 6. Output Coding for Bipolar Operation

BIPOLAR INPUT RANGE SCALE		OUTPUT ( Sinary			
. :	± 10V	MSB	LSB	MSB	LSB
+FS - 1 LSB	+9.9805V	1111 111	1 11	0000 000	0 00
+3/4 FS	+7.5000V	1110 000	0 00	0001 111	1 11
+1/2 FS	+5.0000V	1100 000	0 00	0011 111	1 11
0	0.00000V	1000 000	0 00	0111 111	1 11 📑
-1/2 FS	-5.0000V	0100 000	0 00	101,1.111	1 11
-3/4 FS	-7. <b>50</b> 00V	0010 000	0 00.	1101 111	
-FS + 1 LSB	-9.9805V	0000 000	0.00	- 1111 111	1 11
-FS	-10.000V	0000 000	0 00	1111 111	1 11

# ORDERING INFORMATION

ADS-115	10-Bit, 1 MHz	2	
ADS-116	Sampling A/D Converters		
MODEL	TEMPERATURE IN	PUT VOLTAGE RANGE	
ADS-115MC	0 °C to +70 °C	0 to +10V	
ADS-116MC	0 °C to +70 °C	-10 to +10V	
ADS-115MM	-55 °C to +125 °C	0 to +10V	
ADS-116MM	-55 °C to +125 °C	-10 to +10V	
Frimming Potention	meter:	TP 20K (2 required)	
	C board mounting can be o 31272-8 (Component Lea		

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