

ML13146 Low Power DC - 1.8 GHz Transmitter

LOW POWER INTEGRATED TRANSMITTER FOR ISM BAND APPLICATIONS

SEMICONDUCTOR TECHNICAL DATA

Legacy Device: Motorola MC13146

The ML13146 is an integrated RF transmitter targeted at ISM band applications. It features a 50 Ω linear Mixer with linearity control, voltage controlled oscillator, divide by 64/65 dual modulus Prescaler and Low Power Amplifier (LPA). Together with the receiver chip (ML13145) and either baseband chip (MC33410 or MC33411A/B), a complete 900 MHz cordless phone system can be implemented. This device may be used in applications up to 1.8 GHz.

- Low Distortion LPA: P_{out} 1dB Compression Point 10 dBm
 High Mixer Linearity: IIP3 = 10 dBm
- 50Ω Mixer Input Impedance
- Differential Open Collector Mixer Output
- Low Power 64/65 Dual Modulus Prescaler (ML12054 type)
- 2.7 to 6.5 V Operation, Low Current Drain (25 mA @ 2.0 GHz)
- Powerdown Mode: <60 μA
- Usable up to 1.8 GHz
- Operating Temperature Range $T_A = -20^{\circ}$ to 70° C

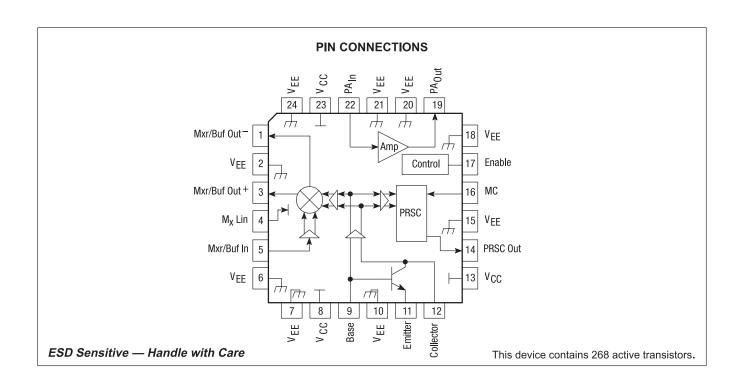


LQFP 24 = -9PPLASTIC PACKAGE **CASE 977**

CROSS REFERENCE/ORDERING INFORMATION

PACKAGE **MOTOROLA** LANSDALE LQFP 24 MC13146FTA ML13146-9P

Note: Lansdale lead free (Pb) product, as it becomes available, will be identified by a part number prefix change from ML to MLE.



MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Power Supply Voltage	V _{CC} (max)	7.0	Vdc
Junction Temperature	T _J (max)	150	°C
Storage Temperature Range	T _{stg}	-65 to 150	°C

NOTES: 1. Maximum Ratings are those values beyond which damage to the device may occur. Functional operation should be restricted to the limits in the Recommended Operating Conditions, Electrical Characteristics tables or Pin Descriptions section.

2. Meets Human Body Model (HBM) ≤100 V and Machine Model (MM) ≤25 V.

RECOMMENDED OPERATING CONDITIONS

Characteristic	Symbol	Min	Тур	Max	Unit
Power Supply Voltage (TA = 25°C)					
	VCC	2.7	_	6.5	Vdc
	VEE	_	0	_	Vdc
RF Frequency Range	fRF	1.0	_	2500	MHz
Ambient Temperature Range	TA	-20	_	70	°C
Maximum Input Signal Level	PIF				
– with no damage		_	– 10	_	dBm
 with minor performace degradation 		_	15	_	dBm

TRANSMITTER DC ELECTRICAL CHARACTERISTICS (T_A = 25°C, V_{CC} = 3.6 Vdc, no input signal, unless otherwise noted)

Characteristic	Symbol	Min	Тур	Max	Unit
Total Supply Current (Enable = V _{CC})	I _{total}	15	18	21	mA
Power Down Current (Enable = V _{EE})	l _{total}	-	30	100	μА
MC Current Input (High)	l _{ih}	70	100	130	μА
MC Current Input (Low)	IiI	-130	-100	- 70	μА
Input high voltage	Vih	V _{CC} - 0.4	_	_	V
Input low voltge	Vil	_	_	0.4	V
Input Current	l _{in}	-50	_	50	μА

$\textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{Vdc}, \ \text{per Test Circuit shown in} \\ \textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{Vdc}, \ \text{per Test Circuit shown in} \\ \textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{Vdc}, \ \text{per Test Circuit shown in} \\ \textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{Vdc}, \ \text{per Test Circuit shown in} \\ \textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{Vdc}, \ \text{per Test Circuit shown in} \\ \textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{Vdc}, \ \text{per Test Circuit shown in} \\ \textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{Vdc}, \ \text{Per Test Circuit shown in} \\ \textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{Vdc}, \ \text{Per Test Circuit shown in} \\ \textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{Vdc}, \ \text{Per Test Circuit shown in} \\ \textbf{TRANSMITTER AC ELECTRICAL CHARACTERISTICS} \ (T_{A} = 25 ^{\circ}\text{C}, \ V_{CC} = 3.6 \ \text{Vdc}, \ \text{Enable} = 3.6 \ \text{V$ Figure 1, unless otherwise noted)

Characteristics	Input Pin	Measure Pin	Symbol	Min	Тур	Max	Unit
Amplifier Output Power (with external matching) @ 950 MHz; P _{in} = –19 dBm	PA _{in}	PA _{out}	PA_Po	-4.5	-3.3	-2.1	dBm
Amplifier 1.0 dB Compression Point (@ 950 MHz = fIF_out)	PA _{in}	PA _{out}	P1dBC.Pt.	_	8.0	_	dBm
Amplifier Output Harmonics (with external matching) @ 950 MHz; P _{in} = –19 dBm	PA _{in}	PA _{out}	_				dBc
2nd 3rd			PA – 2f PA – 3f	–25 –35	–37 –52	_ _	
Low Power Amplifier Power Gain @ 950 mA (matching required)				_	16	_	dB
Mixer/Buffer Output (@ 950 MHz = f_{OSC} ; Mixer input (Pin 5) pulled through 270 Ω resistor)		Buf_out+	P _{Mx} /Buf_out	– 19	-18	– 17	dBm
PLL Setup Time [Note 1]	MC	PRSCout	T _{PLL}	-	10	-	nS
Mixer Input Third Order Intercept Point			IIP3	_	10	-	dBm
VCO Phase Noise (@ 10 kHz offset)		Buf_out+		_	-80	_	dBc/Hz
Prescalar Output Level (10 k 8.0 pF Load)		PRSCout		450	_	600	mVpp

NOTES: 1. MC input (50%) to PRSC_{out} rising output (50%) for proper modulus selection.

2. Typical performance parameters indicate the potential of the device under ideal operation conditions.

RF In 2.2 p ⊥ 100 n $\stackrel{\perp}{+}$ 100 pF 8.2 nH 3 ਤ੍ਰੇ 3.3 nH ≥ 200 2.2 p RF Out 21 24 23 22 20 19 \overline{A} +18 Amp 100 p 68 k Control 17 ─ Enable Mixer/Buf Out 56 k ∕√√√ → MC 50 1.0 n PRSC 15 1.0 n 5 ^/\/ 270 O PRSC Out 13 8 9 10 11 12 100 p 1.8 nH 1.5 p 2.2 p ± 100 n 100 p 2.2 p= D1* * MMBV809L 100 k ≤ $1.0\,\mu$ 470 nH € V_{CC} Tuning Volts

Figure 1. Test Circuit

PIN FUNCTION DESCRIPTION

Pin	Symbol/Type	PIN FUNCTION DESCRIPTION Description	Description
1, 3	Mxr/Buf Out–, Mxr/Buf Out+	Mxr/Buf Out-	Mixer/Buffer Outputs The Mixer/Buffer is a differential open collector configuration which designed to use over a wide frequency range for up conversion as well as direct conversion. Differential to single—ended circuit configuration and matching options are discussed in the Circuit Description section. 6.0 dB of additional Mixer gain can be achieved by conjugately matching the outputs at the desired RF frequency.
2	VEE		VEE, Negative Supply This pin is VEE supply for the mixer IF output. In the application PC board this pin is tied to a common VEE trace with other VEE pins.
4	Mx Lin	Mx Lin	Mixer Linearity Control The mixer linearity control circuit accepts approximately 0 to 200 μA control current to set the dynamic range of the mixer. An Input Third Order Intercept Point, IIP3 of 17 dBm may be achieved at 200 μA of control current.
5	Mxr/Buf I n	Mxr/Buf In	Mixer/Buffer Input The mixer input impedance is broadband 50 Ω for applications up to 2.4 GHz.
6, 7, 18, 24	VEE	VEE VEE VEE VEE VEE VEE	VEE, Negative Supply These pins are substrate connections on the IC. In the application PC board these pins are tied to a common VEE trace with other VEE pins.
8	VCC	Vcc 9 Base	V _{CC} , Supply Voltage Two V _{CC} pins are provided for the Local Oscillator and LO Buffer Amplifier. The operating supply voltage range is from 2.7 Vdc to 6.5 Vdc. In the PCB layout, the V _{CC} trace must be kept as wide as feasible to minimize inductive reactances along the trace. V _{CC} should be decoupled to V _{EE} at the IC pin.
9	Base	VEE 1 =	On-board VCO Transistor The transistor has the emitter, base, collector, V _{CC}
10	VEE	Emitter	and V _{EE} pins available. Internal biasing which is compensated for stability over temperature is
11	Emitter	Emitter 2.0mA 500 μA	provided. It is recommended that the base pin is pulled up to V _{CC} through an RFC chosen for the particular oscillator center frequency. The
12	Collector	Collector	application circuit shows a Colpitts oscillator configuration.

PIN FUNCTION DESCRIPTION (continued)

Pin	Symbol/Type	Description	Description
13	VCC	13	V _{CC} , Supply Voltage
14	PRSC Out	PRSC Out	Prescaler Output The prescaler output provides 500 mVpp drive to the Fin Pin of a PLL synthesizer. Conjugately matching the interface will increase the drive delivered to the PLL input.
15	VEE	15 VCC	V _{EE} , Negative Supply
16	MC	16 MC	Dual Modulus Control Current Input This requires a current input of typically 200 μApp.
17	Enable	17 TO K	Transmitter Enable Enable the transmitter by pulling the pin up to V _{CC} .
19	PA _{out}	 	PA Out The output is an open collector of the cascode transistor low power amplifier (LPA); it is externally biased. The output may be conjugately matched with a shunt L, and series L and C network.
20, 21	VEE	PA _{out} 20 VEE = 21	VEE, Negative Supply VEE pin is taken to an ample dc ground plane through a low impedance path. The path should be kept as short as possible. A two sided PCB is implemented so that ground returns can be easily made through via holes.
22	PA _{in}	V _{EE} 22 V _{ref2} V _{ref1} V _{ref}	PA In The input is the base of the common emitter transistor. Minimum external matching is required to optimize the input return loss and gain.
23	Vcc		V _{CC} , Positive Supply V _{CC} pin is taken to the incoming positive battery or regulated dc voltage through a low impedance trace on the PCB. It is decoupled to V _{EE} ground at the pin of the IC.

CIRCUIT DESCRIPTION

General

The ML13146 consists of a low power amplifier, a 50 Ω linear mixer with linearity control, divide by 64/65 dual modulus prescaler and LPA. This device is designated for use as the low power transmitter in analog and digital FM systems such as UHF and 800 MHz Special Mobile Radio (SMR), UHF Family Radio Services, PCS and 902 to 928MHz cordless telephones. It features a mixer linearity control to preset or auto program the mixer dynamic range, an enable function and a wideband mixer output so the IC may be used either as an up converter or for a direct conversion source. Additional details are covered in the Pin by Pin Description which shows the equivalent internal circuit and external circuit requirements.

Current Regulation/Enable

The device features temperature compensating, voltage independent current regulators which are controlled by the enable function in which "high" powers up the IC.

Mixer: General

The mixer is a double–balanced four quadrant multiplier biased class AB allowing for programmable linearity control via an external current source. An input third order intercept point of 20 dBm has been achieved. The mixer has a 50 Ω single–ended RF input and open collector differential outputs. An onboard Local Oscillator transistor has the emitter, base and collector pinned out to implement a low phase noise VCO in various configurations. Additionally, a buffered prescaler output is provided for operation with a low frequency synthesizer. For direct conversion applications the input of the mixer may be terminated to ground through a 120 to 330 Ω resistor.

Local Oscillator/Voltage Control Oscillator

The on–chip transistor operates with coaxial transmission line or LC resonant elements to over 1.8 GHz. Biasing is done with a temperature/voltage compensated current source in the emitter. A RFC from V_{CC} to the base is recommended. The transistor can be operated in the classic Colpitts, Clapp, or Hartley configuration. The application circuit (Figure 8) depicts a parallel resonant VCO which can cover the entire 902 to 928 MHz frequency band with phase noise of approximately –80 dBc/Hz at a 10 kHz offset (see Figure 2). For this configuration, the LO will be driven with approximately 100 mVrms, and the frequency of oscillation can be approximated by:

$$F_{OSC} = \frac{1}{\left(2\pi\sqrt{\left(\frac{C1\ C2}{C1 + C2}\right)\left(\frac{C3\ Cv}{C3 + Cv} + 3.6\ pF\right)(L1 + 1.8\ nH)}\right)}$$

where Cv is the equivalent capacitance of the varactor at the control voltage.

For higher frequency operation, a series tuned oscillator configuration is recommended. Table 1 contains the S-parameters for the VCO transistor in a common collector configuration. This information is useful for designing a VCO at other operating frequencies or for various other oscillator topologies.

The output power (at Mix/Buf Out) can be varied by adjusting the value of R5 as illustrated in Figures 3 and 4. Figure 5 shows the typical operating window for the prescaler.

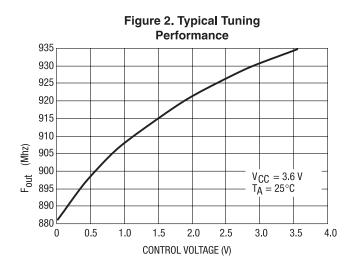


Figure 3. Mixer/Buffer Output versus 1st LO Input

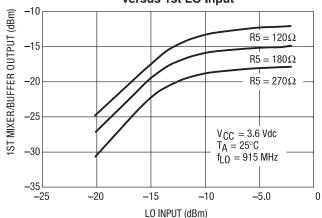
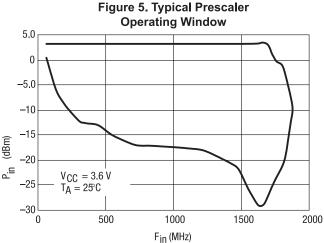


Figure 4. Test Circuit for Figure 3.



Mixer/Buffer Input

The Mixer/Buf In pin is a broadband, 50 Ω input used to drive the IF port of the mixer (see Table 2, S11parameters). The Mixer/Buf In pin can be used in one of three modes:

- 1. A IF signal can be applied to this pin and up—converted to the desired RF frequency.
- 2. A resistor can be connected to ground, controlling the RF output power.
- A resistor can be connected to V_{CC}, disabling the entire mixer.

The linear gain of the Mixer/Buf when used as a buffer is approximately -5.0 to -8.0 dB.

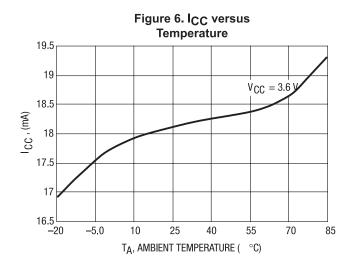
Mixer/Buffer Outputs

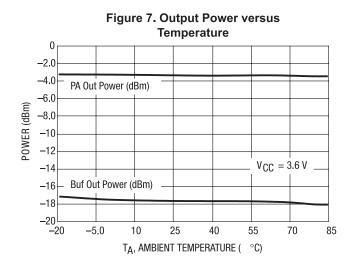
The mixer outputs (Mixer/Buf Out + and Mixer/Buf Out -) are balanced, open collector. A shunt resistor of 200 Ω minimum to V_{CC} is recommended for stability.

The outputs can be used as a single-ended driver or connected in a balanced-to-unbalanced configuration. If the single-ended driver configuration is used, the unused output must be tied directly to V_{CC}. For the balanced-to-unbalanced configuration, an additional 3.0 to 6.0 dB of power gain can be achieved. Conjugate matching is easily accomplished to the desired load by the addition of a shunt and series element (see Table 2, S22 parameters).

Low Power Amplifier (LPA)

The LPA is internally biased at low supply current (approximately 2.0 mA emitter current) for optimal low power operation, yielding a 10 dBm 1.0 dB output power compression point. Input and output matching may be achieved at various frequencies using few external components (see Table 3 S–parameters). Typical power gain is 16 dB with the input/output conjugately matched to the source/load impedance. A minimum 200 Ω shunt resistor from the output to V_{CC} is recommended for stability.





RF In 2.2 p \pm 100 pF 100 p 10 nH 🗟 ± 10 n ± 1.0 n **§** 200 ਤ੍ਰੇ 3.3 nH 2.2 p RF Out 24 23 22 21 20 19 $\overline{}$ $\overline{\mathcal{A}}$ 1.0 n 100 p 18 Amp 56 k Control ─ Enable <u>| 10</u>0 р Mixer/Buf Out 56 k ∕√√√ 3 16 → MC 51 1.0 n PRSC 15 1.0 n 5 14 ^/\\ 270 → PRSC Out 6 13 8 9 10 12 11 1.8 nH 100 p 4.7 p C2 4.7 p C1 $\pm 10 \,\mathrm{n}$ ± 100 p 4.3 p C3 D1* <u>+</u> 10 μ * MMBV809L 10 k ≤ 100 p 1.0 k≤ VCC(3.6 V)Tuning Volts

Figure 8. Applications Circuit

Legacy Applications Information

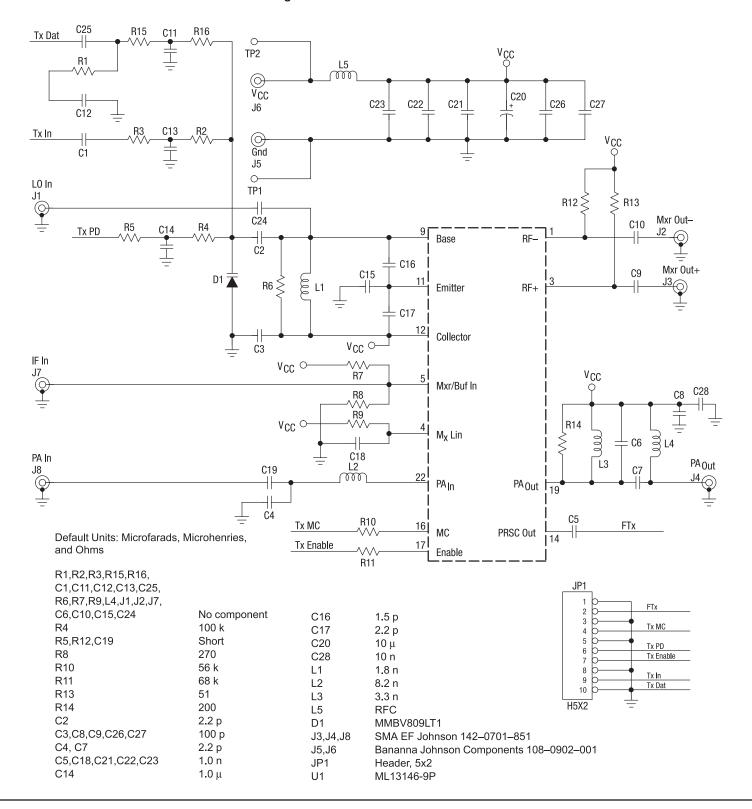
Evaluation PCB

The evaluation PCB is a versatile board which allows the ML13146 to be configured as a basic transmitter, or to characterize individual operating parameters.

The general purpose schematic and associated parts list for the PCB is given in Figure 9. This parts list build—up is identical to the Test Circuit illustrated in Figure 1, although parameters can very significantly due to differences in PCB parasitics. Figures 10, 11, and 12 show the actual PCB component, ground and solder sides, respectively.

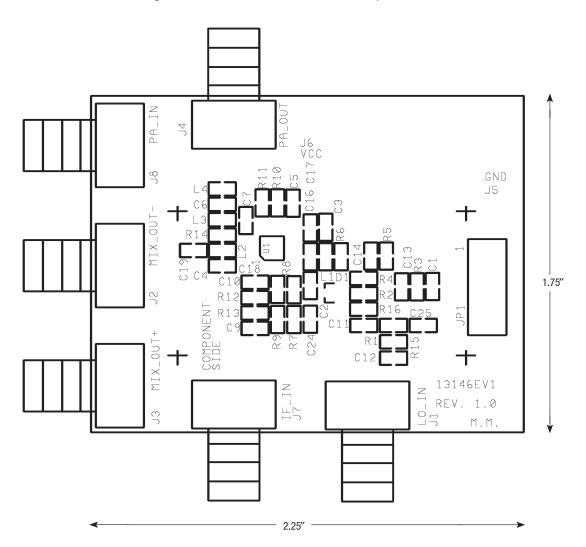
Please refer to AN1687/D and AN1691/D for additional details and applications for the device.

Figure 9. Evaluation PCB Schematic



Legacy Applications Information

Figure 10. ML13146 Evaluation PCB Component Side



Legacy Applications Information

Figure 11. ML13146 Evaluation PCB Ground Plane

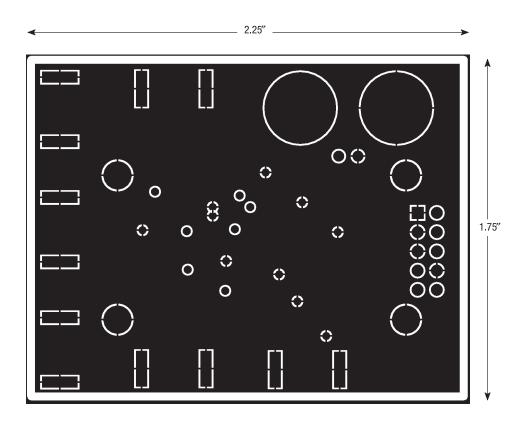


Figure 12. ML13146 Evaluation PCB Solder Side

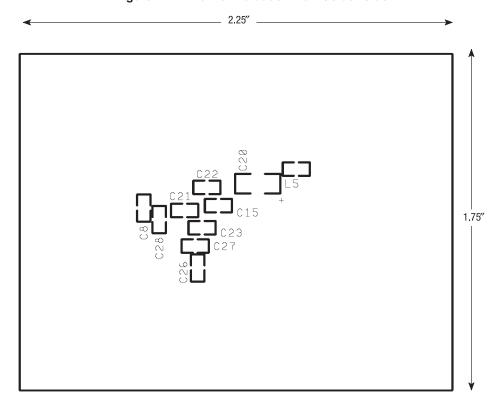


Table 1. VCO Transistor S–Parameters 3.6 Vdc; 50 Ω Load and Source Impedance; Common Collector

Freq (MHz)	S11 Mag	S11 Ang	S21 Mag	S21 Ang	S12 Mag	S12 Ang	S22 Mag	S22 Ang
25	0.99	-1	0.88	0	0.01	44	0.10	- 7
50	0.99	-2	0.92	-1	0.02	61	0.09	-9
100	0.98	- 5	0.95	-2	0.04	70	0.07	-37
150	0.98	- 7	0.97	-3	0.06	73	0.07	-4 7
200	0.97	-10	1.04	-4	0.07	73	0.06	-86
300	0.95	-14	1.11	-8	0.10	71	0.09	-124
400	0.93	-19	1.23	-12	0.13	67	0.14	-149
450	0.92	-21	1.26	-14	0.15	66	0.15	– 155
500	0.91	-23	1.30	-16	0.16	65	0.17	-159
600	0.86	-28	1.35	-20	0.19	61	0.20	-167
750	0.79	-37	1.46	-25	0.24	57	0.26	-172
800	0.79	-39	1.48	-26	0.25	56	0.28	-174
850	0.77	-42	1.48	-28	0.26	54	0.29	– 177
900	0.74	-44	1.47	-31	0.28	52	0.28	-179
950	0.67	-49	1.53	-35	0.30	49	0.31	174
1000	0.61	-55	1.59	-38	0.33	47	0.34	171
1250	0.45	-81	1.61	-50	0.41	38	0.38	157
1500	0.35	-159	1.68	-67	0.53	16	0.38	134
1750	0.85	107	1.60	-100	0.57	– 15	0.33	97
2000	1.02	76	1.17	-117	0.47	-32	0.18	86
2250	1,25	76	1.13	-125	0.55	-38	0.19	89
2500	1.58	53	0.84	-150	0.56	-64	0.09	57

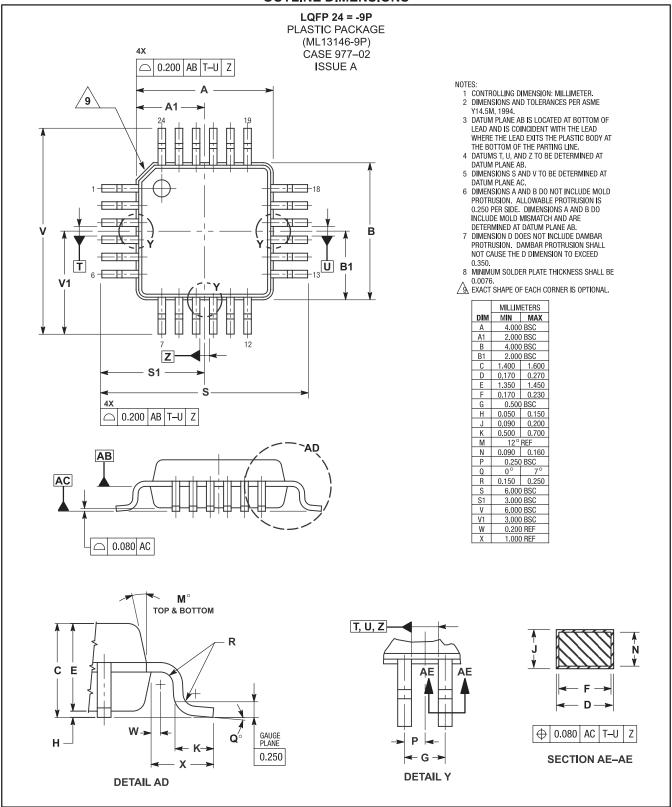
Table 2. Mixer Input/Output S–Parameters: 200 Ω Pull–Up Resistor

Freq (MHz)	S11 Mag	S11 Ang	S21 Mag	S21 Ang	S12 Mag	S12 Ang	S22 Mag	S22 Ang
50	0.11	176.8	0.43	-4.2	0.001	38.7	0.60	-1.9
100	0.11	177.9	0.43	- 7.5	0.002	19.8	0.60	-3.5
200	0.11	179.4	0.42	-13.7	0.001	28.3	0.60	-6.7
300	0.10	179.5	0.42	-20.7	0.001	69.8	0.61	-9.9
400	0.10	177.2	0.42	-27.3	0.001	106.3	0.61	-13.2
450	0.11	174.9	0.41	-31.1	0.001	135.2	0.62	-14.8
500	0.10	177.7	0.42	-34.1	0.002	138.2	0.62	-16.6
600	0.09	174.3	0.42	-41.8	0.003	150.5	0.63	-20.0
700	0.09	167.2	0.41	-49.3	0.005	158.7	0.64	-23.5
750	0.08	162.8	0.41	-53.9	0.006	166.0	0.65	-25.2
800	0.08	156.6	0.40	-58.4	0.008	166.5	0.65	-26.9
850	0.06	152.3	0.40	-62.7	0.009	171.2	0.66	-28.7
900	0.05	145.2	0.39	-66.4	0.012	177.6	0.66	-30.3
950	0.04	131.1	0.38	-71.6	0.015	-179.7	0.67	-31.9
1000	0.02	101.1	0.38	-76.7	0.019	178.0	0.68	-33.7
1250	0.08	-41.5	0.27	-96.8	0.042	137.1	0.73	-43.2
1500	0.40	-87.6	0.24	-90.2	0.036	129.9	0.78	-53.3
1750	0.50	-144.1	0.30	-114.0	0.058	142.8	0.86	-63.8
2000	0.51	– 173.5	0.22	-133.0	0.174	151.6	0.96	-81.3

Table 3. LPA S–Parameters: 200 Ω Pull–Up Resistor

Freq (MHz)	S11 Mag	S11 Ang	S21 Mag	S21 Ang	S12 Mag	S12 Ang	S22 Mag	S22 Ang
200	0.76	-26.0	9.3	148.1	0.0006	73.3	0.60	-12.4
300	0.71	-37.5	8.5	135.2	0.0011	74.4	0.60	-18.5
400	0.67	-47. 2	7.6	124.5	0.0011	79.6	0.61	-24.6
450	0.64	- 51.7	7.2	118.6	0.0010	66.0	0.62	-28.3
500	0.62	-55.4	6.9	114.2	0.0011	45.4	0.62	-31.6
600	0.58	-63.7	6.3	105.3	0.0012	16.7	0.64	-38.8
700	0.54	- 72.1	5.6	95.2	0.0016	-20.9	0.66	-45.6
750	0.52	-74.6	5.4	91.8	0.0013	-36.9	0.66	-48.5
800	0.51	-77. 9	5.2	87.7	0.0023	-50.8	0.67	-52.6
850	0.49	-80.3	5.0	83.8	0.0033	-63.6	0.68	-56.1
900	0.49	-83.5	4.7	79.6	0.0044	- 78.7	0.68	-60.3
950	0.48	-85.4	4.5	77.2	0.0060	-90.3	0.68	-63.2
1000	0.48	-88.8	4.3	74.7	0.0082	-97.6	0.68	-65.8
1250	0.51	-102.7	3.7	58.8	0.0249	-136.6	0.73	-74.6
1500	0.48	-119.7	3.3	37.6	0.0273	172.0	0.90	-87.7
1750	0.47	-130.0	2.7	20.5	0.0290	166.5	0.97	-103.7
2000	0.51	-136.7	2.2	-1,1	0.0386	164.1	1.01	– 119.1

OUTLINE DIMENSIONS



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