

Evaluation Board for the TLC320AD50C DSP Analog Interface Circuit

User's Guide

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Preface

Read This First

About This Manual

This user's guide discusses the design, use, and performance of the TLC320AD50 Evaluation Module.

How to Use This Manual

This document contains the following chapters:

- ☐ Chapter 1 – Introduction
- ☐ Chapter 2 – AD50-EVM Design and Construction
- ☐ Chapter 3 – Setting Up the AD50-EVM
- ☐ Chapter 4 – Results Obtained With AD50 EVM
- ☐ Appendix A –Installing the AD50-EVM
- ☐ Appendix B –Programmable Logic
- ☐ Appendix C –Converting DSK+ Software for the AD50-EVM

Notational Conventions

This document uses the following conventions.

- ☐ Program listings, program examples, and interactive displays are shown in a special typeface similar to a typewriter's. Examples use a **bold version** of the special typeface for emphasis; interactive displays use a **bold version** of the special typeface to distinguish commands that you enter from items that the system displays (such as prompts, command output, error messages, etc.).

Here is a sample program listing:

```
0011 0005 0001      .field    1, 2
0012 0005 0003      .field    3, 4
0013 0005 0006      .field    6, 3
0014 0006           .even
```

Here is an example of a system prompt and a command that you might enter:

```
C:  csr -a /user/ti/simuboard/utilities
```

- In syntax descriptions, the instruction, command, or directive is in a **bold typeface** font and parameters are in an *italic typeface*. Portions of a syntax that are in **bold** should be entered as shown; portions of a syntax that are in *italics* describe the type of information that should be entered. Here is an example of a directive syntax:

.asect *"section name", address*

.asect is the directive. This directive has two parameters, indicated by *section name* and *address*. When you use *.asect*, the first parameter must be an actual section name, enclosed in double quotes; the second parameter must be an address.

- Square brackets (**[** and **]**) identify an optional parameter. If you use an optional parameter, you specify the information within the brackets; you don't enter the brackets themselves. Here's an example of an instruction that has an optional parameter:

LALK *16-bit constant [, shift]*

The LALK instruction has two parameters. The first parameter, *16-bit constant*, is required. The second parameter, *shift*, is optional. As this syntax shows, if you use the optional second parameter, you must precede it with a comma.

Square brackets are also used as part of the pathname specification for VMS pathnames; in this case, the brackets are actually part of the pathname (they are not optional).

- Braces (**{** and **}**) indicate a list. The symbol **|** (read as *or*) separates items within the list. Here's an example of a list:

{ ** | *+ | *- }*

This provides three choices: ***, **+*, or **-*.

Unless the list is enclosed in square brackets, you must choose one item from the list.

- Some directives can have a varying number of parameters. For example, the *.byte* directive can have up to 100 parameters. The syntax for this directive is:

.byte *value₁ [, ... , value_n]*

This syntax shows that *.byte* must have at least one value parameter, but you have the option of supplying additional value parameters, separated by commas.

Related Documentation From Texas Instruments

TLC320AD50C Data Manual, Literature number SLAS131

Data Acquisition Data Book, Literature number SLAD001

Data Converter Selection Guide, Literature number SLABE05

Operational Amplifiers Data Book Volume A, Literature number SLYD011

Operational Amplifiers Data Book Volume B, Literature number SLYD012

Rail-to-Rail Operational Amplifier Selection Guide,
Literature number SLOBE02

Single Supply Operational Amplifier Selection Guide,
Literature number SLOBE03

Mixed Signal Analog CD-ROM, Literature number SLYC005

TMS320C54x CPU and Peripherals, Literature number SPRU131

TMS320C54x Algebraic Instruction Set, Literature number SPRU179

TMS320C54x DSKplus User's Guide, Literature number SPRU191

Much useful software is available from the TI Internet site. The main TI Web site is at

<http://www.ti.com/>

Information on the DSK+ is at

<http://www.ti.com/sc/docs/dsps/tools/c54x/c54xdskp.htm>

DSK+ software can be downloaded from

<http://www.ti.com/sc/docs/dsps/tools/c54x/softsupp.htm>

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Introduction

This user's guide discusses the design of the AD50-EVM evaluation board and its use to demonstrate the performance of the TLC320AD50C (AD50) analog interface circuit.

The TLC320AD50C provides high-resolution low-speed signal conversion from digital-to-analog (D/A) and from analog-to-digital (A/D) using oversampling sigma-delta technology. This device consists of two serial synchronous conversion paths (one for each data direction) and includes an interpolation filter before the DAC and a decimation filter after the ADC. Other overhead functions provide on-chip timing and control. The sigma-delta architecture produces high-resolution analog-to-digital and digital-to-analog conversion at low system speeds and low cost.

The options and the circuit configurations of this device can be programmed through the serial interface. The options include reset, power down, communications protocol, serial clock rate, signal sampling rate, gain control, and test mode.

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The TLC320AD50 is an analog interface circuit (AIC) with many features that make it suitable for DSP-based applications

- ☐ Single 5-V supply or dual (5-V analog and 3.3-V digital) supplies
- ☐ 16-bit resolution sigma-delta ADC and DAC
- ☐ 85 dB (min) signal to noise
- ☐ Inherent antialiasing filtering and $\sin(x)/x$ compensation
- ☐ High input impedance
- ☐ Supports up to 4 devices on one serial interface
- ☐ Low operating power (175 mW max)
- ☐ Power down mode (20 mW max)
- ☐ Small package size

1.1 Scope of User's Guide

This application note discusses the design of the AD50-EVM evaluation board and its use to demonstrate the performance of the TLC320AD50C (AD50) analog interface circuit. The results come from measurements on a small number of samples. For specifications refer to the datasheet.

The AD50-EVM has two AD50 devices for stereo operation. Two AD50-EVMs can be configured as a four-channel system using a single serial interface. The AD50-EVM can be interfaced directly to the DSK+ DSP starter kit or other systems which have a compatible synchronous serial interface.

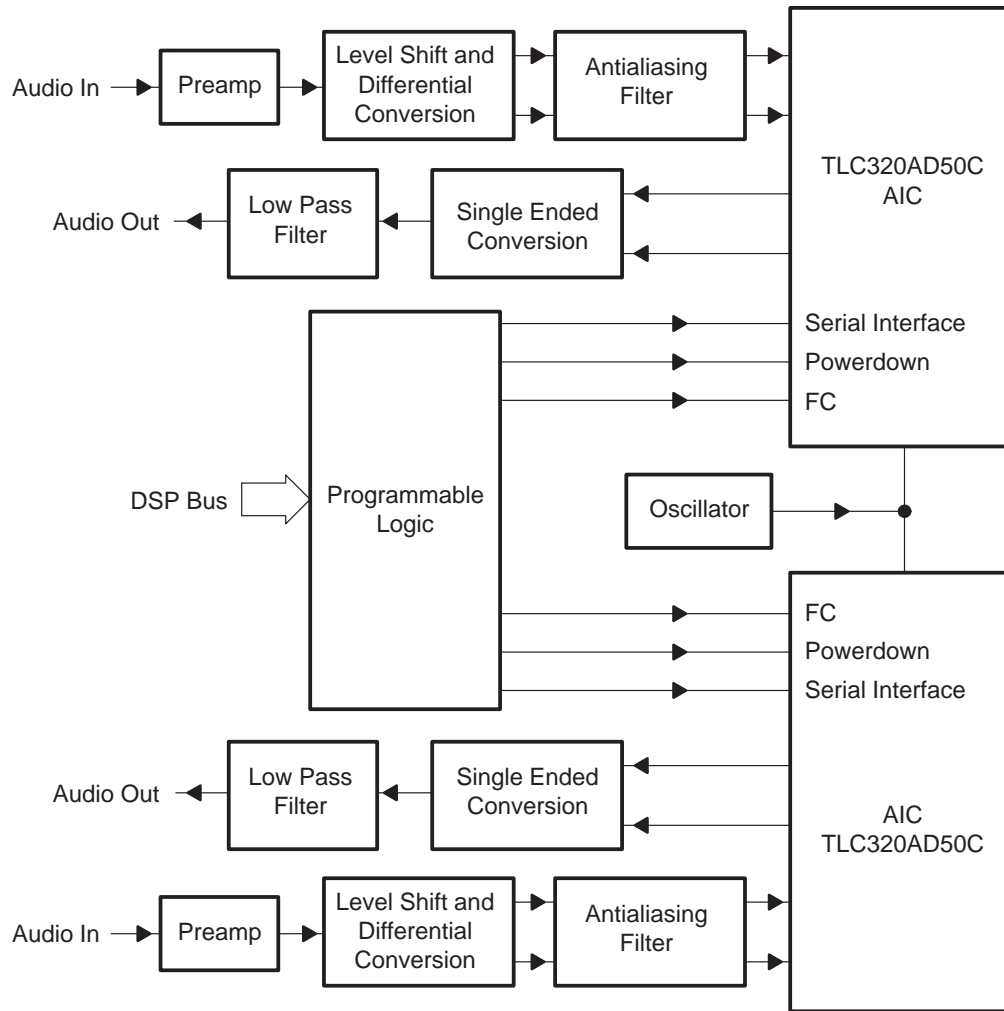
The objective was to design a development board (the AD50-EVM) which would allow prospective users of the AD50 to determine its capabilities with a minimum effort. The board can be directly connected to the low cost TMS320C54x DSP starter kit (DSK+), or to any other system with a compatible synchronous serial interface. Directly compatible DSP devices include TMS320C2x, C2xx, C3x, C5x, C54x and C6xxx.

An example program for the DSK+ development system allows the board to be used as a sine-wave generator, or to output on the DAC samples read in from the ADC. In this echo mode signal processing functions such as filtering can easily be included. The AD50-EVM board was also interfaced to a TMS320C25 development board, which was used to transfer analog data to a personal computer running real-time FFT spectrum analysis software. This system was used to prepare the ADC and DAC FFT spectrograms shown in this user's guide.

1.2 Block Diagram of AD50-EVM

Figure 1–1 shows the block diagram of the AD50-EVM evaluation board.

Figure 1–1. AD50-EVM Block Diagram



The input signal is first buffered and optionally amplified by the preamp stage before being level-shifted and converted to a differential pair of signals. The external antialiasing filter is a simple continuous time filter to remove RF noise. The output from the codec is converted to a single ended signal and is filtered to remove high frequency noise.

A negative power supply is generated on the AD50-EVM to allow for dc coupled input and output signals. The clock signal for the codec can be generated by an oscillator on the board or from an external clock. The sampling frequency is set using the programmable clock divider within the AD50. The AD50 FC, RESET, and POWERDOWN signals can be controlled via a software programmable register.

1.3 Special Considerations When Using Sigma-Delta Converters

Sigma-delta analog-to-digital converters typically consist of an analog modulator (fourth order in the case of the AD50) followed by a digital filter section. The modulator contains a 1-bit ADC (a comparator) which produces a 1-bit wide data stream, which is applied to the input of the digital filter. It is also applied to the input of a 1-bit DAC, the output of which is fed back to the input of the modulator. The 1-bit ADC is clocked much faster than the desired output sampling frequency (F_S), ($64 \times F_S$ for the AD50) and the large amount of quantization noise generated is uniformly spread over a wide range of frequencies. A noise-shaping filter within the analog modulator reduces the noise in the pass-band, increasing it elsewhere. A low-pass digital filter then removes the unwanted high frequency quantization noise and the signal is resampled at the desired output frequency.

The sigma-delta DAC in the AD50 consists of a digital interpolating filter operating at $256 \times F_S$ followed by a 1-bit DAC and a second-order modulator. The digital filter removes most of the image frequencies, that would otherwise be present at the output of a conventional DAC. This avoids the need for a high-order analog low-pass reconstruction filter. Separate $\sin(x)/x$ compensation is not needed, as this is inherent in the DAC architecture.

Sigma-delta converters have several advantages and a few disadvantages relative to other types of ADC which are summarized below.

1.3.1 Advantages

- ☐ Sigma-delta conversion is inherently linear because there is no reference resistor chain as in flash or successive approximation converters. This results in extremely low distortion.
- ☐ Inherent monotonicity
- ☐ No missing codes
- ☐ Antialias filtering is inherent within sigma-delta converters. This greatly simplifies their application, because only a simple external RC filter is required at the input of the ADC to achieve the necessary alias rejection.
- ☐ The digital filters which form an integral part of the ADC and DAC are usually of the finite impulse response type, which gives a linear phase characteristic with high stability.
- ☐ The cutoff frequency of the digital filters automatically tracks the sampling rate.
- ☐ DAC anti-image filtering is greatly simplified because an internal digital filter attenuates image frequencies. A simple analog filter is sufficient to remove HF noise.
- ☐ No need for $\sin(x)/x$ compensation.
- ☐ The sigma-delta architecture is compatible with dense, low-cost, digital IC processes.

1.3.2 Disadvantages

- ☐ The digital filters in sigma-delta converters introduce into the conversion process a time delay that makes them unsuitable for some control applications. The AD50 ADC has a delay of 17 samples and the DAC has an 18 sample delay.
- ☐ It is not practical to multiplex several inputs to a single sigma-delta ADC except at very low rates, because each channel would be corrupted by the earlier samples from other channels still propagating through the digital filter. Although the AD50 has an input multiplexer, this is only for selecting one or another input, not for interleaving two input channels onto one data stream.
- ☐ Audio band converters are optimized for ac signals and a small dc offset may be present.
- ☐ Spurious low level tones can sometimes occur at very low input signal levels, especially if clock signals at $F_s/2$ are allowed to couple into the reference voltage pins. Such tones can be identified because their frequency is affected by small changes in dc offset.

AD50-EVM Design and Construction

This chapter discusses the printed-circuit board design considerations for the AD50-EVM.

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2.1 PCB Construction

The AD50-EVM printed-circuit board is constructed of 4 layers, with ground and power planes sandwiched between the top and bottom signal-carrying layers. This minimizes the coupling of RF noise into the system by providing very low impedance to power and ground, and by shielding signal tracks. The AD50 is available in two surface-mount packages, the DW and PT. The PT has a smaller footprint size and is much thinner (only 1.6 mm high). It has 48 pins at a pitch of 0.5 mm, making it very suitable for PCMCIA and other miniature applications. However it is difficult to hand solder, and for prototype evaluation the use of the DW package is recommended. This package has 28 pins at a pitch of 1.27 mm (0.05 inch) and so is much easier to handle. Surface-mounted components were used except for the connectors, which through-hole connectors offering robustness.

The printed-circuit board is the same size as the DSK+ board. The AD50-EVM and the DSK+ can be stacked one above the other using inter-board links. Either board can be on top, but placing the AD50-EVM on top gives better access to the EVM test points.

A separate connector has been provided for interfacing to systems other than the DSK+. This brings out the SCLK, \overline{FS} , DIN, DOUT, and \overline{RESET} signals. Each signal is interleaved with a ground conductor in the ribbon cable, allowing a longer cable length without crosstalk. Nevertheless, this cable should be as short as reasonably possible, since ringing in the unterminated cable may become excessive for lengths greater than 3 feet. To minimize RF emissions, the cable should be shielded, with the shield connected to the ground plane at the ground point provided adjacent to pins 1 and 2 of the connector. This connector must not be used simultaneously with the connections around the edge of the AD50-EVM board that are specifically intended for the DSK+.

For the best signal-to-noise ratio, it is often recommended that the ground plane be split into separate analog and digital sections, joined together in only one place with the split passing under the AD50 package between its analog and digital ground pins. The split prevents digital noise currents from inducing noise voltages in the analog ground of the device. The DSK+ has ground and power pins along both edges of the board, making the use of split ground planes ineffective. Instead short breaks were inserted on the ground plane to separate the analog and digital grounds and reduce the common impedance

2.2 Power Supply

The AD50-EVM board requires only +5 Vdc; a –5-Vdc supply for some of the op-amps is generated on the board using a CMOS 7660 charge pump phase-locked to the frame sync output of the AD50. Phase-locking minimizes the risk of audible beats between the sampling clock and the 7660 clock. The OSC pin of the 7660 is coupled to the master $\overline{\text{FSD}}$ with a 22 pF capacitor. This method allows the 7660 oscillator to free-run when there is no frame sync, ensuring that the negative power is always maintained, even when the AD50 is held in reset. The 7660 divides the signal on the OSC pin by 2 internally, ensuring that the charge pump operates at a 50% duty cycle, even with a grossly asymmetric input such as that provided by frame sync. Not all negative supply generators contain a divider. Check before using any other type.

The digital part of the AD50 can be operated at 3.3 V. Users wishing to investigate this should remove R16 and R36 and connect 3.3 V power to L_VD and R_VD.

2.3 System Clock

When the AD50-EVM is used in standalone mode, in conjunction with a separate DSP system, a crystal oscillator must be fitted to the socket provided. A frequency of 10.24 MHz is suggested, as this will allow standard sampling frequencies such as 10, 16, and 20 kHz to be achieved by programming the AD50 clock divider and phase-locked loop. Table 2–1 shows the sampling frequencies that can be selected for a number of master clock frequencies. Frequencies shown in brackets are above the maximum sampling frequency specified for the AD50.

In standalone mode the AD50-EVM can be clocked at frequencies up to 22.579 MHz. However, when used in conjunction with the DSK+ the upper frequency is limited by the maximum clock rate of the TMS320C542 DSP device. The 10.24 MHz oscillator supplied with the AD50-EVM will drive the TMS320C542 on the DSK+ at 40.96 MHz, which is only slightly higher than its maximum operating frequency of 40.00 MHz. However, at room temperature this should not cause any problems, because the TMS320C542 is tested by TI at 40.00 MHz over the temperature range 0–70°C. The authors have successfully operated a DSK+ in conjunction with an AD50-EVM at MCLK frequencies from 4 MHz to 13 MHz at room temperature, but this may not work with all DSK+ units. Alternatively, the DSK+ and AD50-EVM clocks can be separated by removing resistor R80 from the AD50-EVM board and a suitable oscillator installed on each board. (If this modification is made, 3 and 4 channel modes will not work correctly because all the AD50s must have an identical MCLK. Removing R80 prevents the clock from being transmitted between the boards.)

It is important to ensure that only one oscillator is used in the system (except as described above), otherwise results will be unpredictable. No damage will be caused, however, as current limiting resistors are provided on the output of each oscillator. Electrically it makes no difference whether the oscillator is fitted to the DSK+, or to the first or second AD50-EVM board.

Table 2–1. Sampling Frequencies for Various MCLK and Register 4 Values

Reg 4, Bit 7	Reg 4, Bits 6–4	8.192 MHz	10.000 MHz	10.240 MHz	11.2896 MHz
bit 7 = 0 PLL on	0 (default)	8 kHz	9.765 kHz	10 kHz	11.025 kHz
	1	(64 kHz)	(78.125 kHz)	(80 kHz)	(88.2 kHz)
	2	(32 kHz)	(39.063kHz)	(40 kHz)	(44.1 kHz)
	3	21.333 kHz	(26.042 kHz)	(26.666 kHz)	(29.4 kHz)
	4	16 kHz	19.531 kHz	20 kHz	22.05 kHz
	5	12.8 kHz	15.625 kHz	16 kHz	17.64 kHz
	6	10.666 kHz	13.021 kHz	13.333 kHz	14.7 kHz
	7	9.1432 kHz	11.161 kHz	11.429 kHz	12.601 kHz
bit 7 = 1 PLL off	0	2 kHz	2.441 kHz	2.500 kHz	2.756 kHz
	1	16 kHz	19.531 kHz	20 kHz	22.05 kHz
	2	8 kHz	9.766 kHz	10 kHz	11.025 kHz
	3	5.333 kHz	6.510 kHz	6.666 kHz	7.35 kHz
	4	4 kHz	4.883 kHz	5 kHz	5.513 kHz
	5	3.2 kHz	3.906 kHz	4 kHz	4.41 kHz
	6	2.666 kHz	3.255 kHz	3.333 kHz	3.675 kHz
	7	0.798 kHz	2.79 kHz	2.857 kHz	3.15 kHz

2.4 Reset

In all operating modes the AD50-EVM is reset when the power is switched on. In DSK+ mode it is also reset when the DSK+ itself is reset. The programmable logic on the AD50-EVM is configured so that in the reset state all the AD50s are powered down and in slave mode and that the AC01 on the DSK+ is powered up as normal. This means that even with one or two AD50-EVMs attached to a DSK+ system, all the DSK+ demonstration software works normally with the AC01 AIC.

To use one or more AD50s it is necessary to program bits 0-3 of I/O address 0. Bit 0 controls the reset state of the AD50s. Bit 0 = 0 resets all AD50s and bit 0 = 1 allows them to run. Bits 1, 2, and 3 are encoded to control the number of AD50s that are powered up (see Table B–1 for details). Note that no provision has been made to operate the AD50s and the AC01 together. This is because their control registers are incompatible and it would be difficult to prevent bus conflicts on the serial interface.

The reset pins are connected to both the SERIAL and CONTROL cable headers for use in standalone mode (see Table A–2 for pinouts).

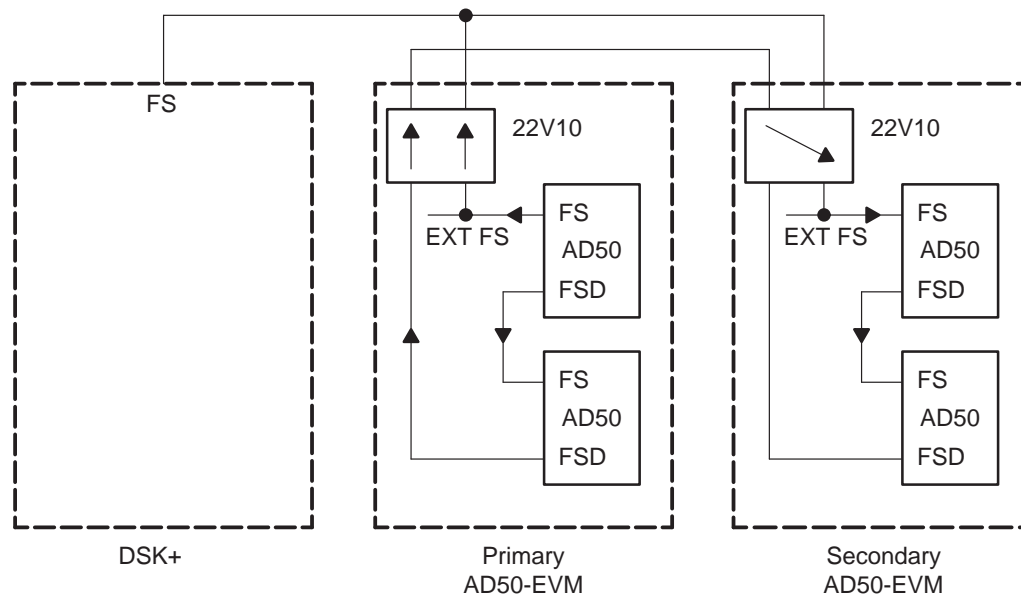
2.5 Serial Port Interfacing

The AD50-EVM has been designed to interface to the same synchronous serial port that is used by the AC01 on the DSK+ system. This is achieved by sharing the serial clock, frame sync and data signals, and by powering down the AC01 when AD50s are in use.

When two AD50-EVMs are connected together the serial interface frame sync signals need to be configured differently on the primary and secondary boards. This is achieved using the 22V10 programmable logic device as shown in Figure 2–1. The logic equations are listed in Appendix B. The serial clock, serial data in, and serial data out signals are connected to all devices. Only one device at a time drives these lines.

When two AD50-EVMs are used in standalone mode together with a separate DSP system, the ribbon cable must be attached to the primary board so that the master device frame sync signal is transmitted to the external system. The two AD50-EVM boards have all the necessary signals linked via the four connectors around the periphery.

Figure 2–1. Configuration of Serial Interface on Primary and Secondary AD50-EVMs



2.6 Analog Input

The AD50 can be operated from a single 5 V supply or from a 5 V analog supply and a 3.3 V digital supply. The AD50-EVM uses 5 V for both analog and digital supplies to be compatible with the DSK+, and generates an additional -5 V supply on the board. This negative supply allows the input and output circuitry to be dc coupled. In an ac coupled system alternative configurations may be used that do not require a negative supply.

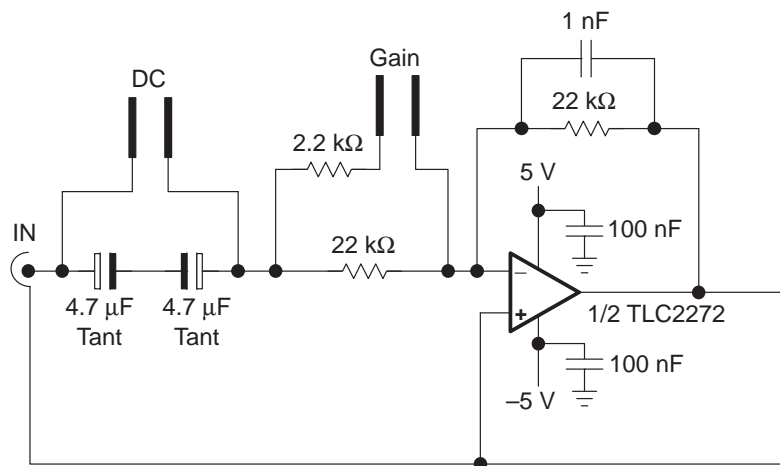
The input circuitry consists of a switchable gain preamp, a single-ended to differential converter with level shifting, and an antialiasing filter.

2.6.1 Preamp Design

This amplifier stage, illustrated in Figure 2–2 allows the use of either a line level input (approximately 1 Vrms full scale) or a low level input (approximately 100mVrms full scale). With the link labeled *GAIN* removed the preamp has unity gain from dc to 20 kHz. With this link inserted the preamp gain is increased by 20 dB.

The signal can be ac coupled, by removing the link labeled *DC*. The use of two tantalum capacitors back-to-back allows ± 10 V dc voltage bias at the input without damaging the capacitors.

Figure 2–2. AD50-EVM Input Preamp

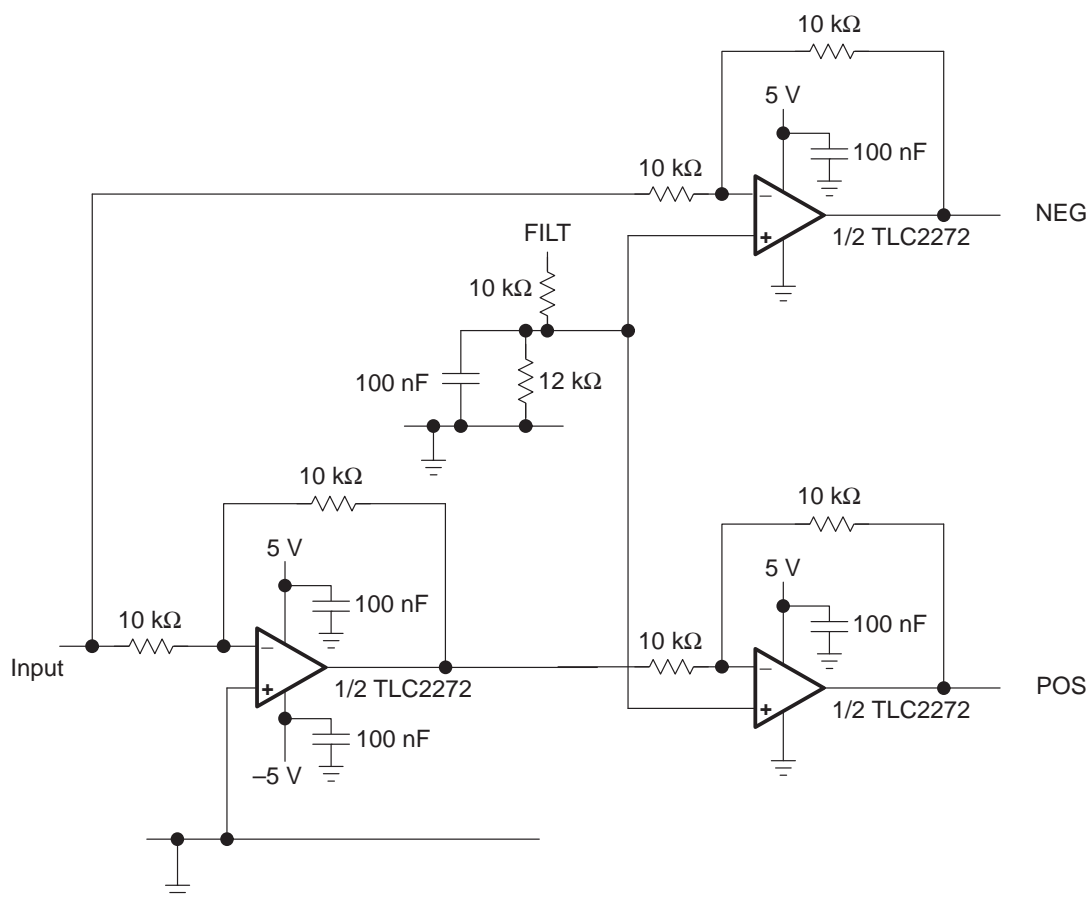


This preamp uses the op amp in the inverting configuration to ensure that the input common mode range of the op amp cannot be exceeded, since both inputs are always at a voltage close to ground. In the non-inverting configuration the op amp inputs can approach the supply rails, causing distortion.

2.6.2 Input Stage Design

The AD50 analog to digital section uses differential inputs biased at 2.5 V. Maximum code is generated with $3 V_{PEAK-PEAK}$ on both differential inputs. The input stage provides the necessary level shifting and single-ended to differential conversion, prior to the inputs of the AD50.

Figure 2–3. AD50-EVM Input Stage



The first op amp inverts the incoming signal, to provide the differential pair while the second two op amps do the necessary level shifting. The design of this stage ensures that any noise on FILT appears equally, and in phase, on both differential inputs. The first op amp needs ± 5 -V supplies, while the second two are powered from just 5-V and ground to ensure that the inputs to the AD50 never go below ground.

It is recognized that this device is likely to be used in predominantly digital designs, powered from a single 5 V supply, where the need for a negative supply could be perceived as a problem. However, since the current drawn from this negative supply will be small, it is possible to use an inverter chip to provide a local negative rail for the op amps. The 7660 device requires no external inductors and just 2 external capacitors. Its internal clock can be synchronized to the codec's sampling clock, ensuring that any interference generated by the 7660 appears on the codec's output as a dc offset or at the Nyquist frequency rather than as noise in the pass-band.

2.6.3 Op Amp Selection

The design of the input stage uses op amps in the inverting configuration. This has the advantage of keeping both inputs at the reference voltage, preventing problems with input common mode range.

For a full-scale signal the outputs of the op amps that drive the codec inputs will reach 4 V. Many op amps will not behave well at this output level with a 5-V supply (in fact the system was tested at 4.75 V to allow for the minimum rated power supply voltage of the AD50).

Miniature systems often require that analog and digital circuits be in close proximity to each other. This can lead to problems where radio frequency interference from clock signals or DSP bus lines is demodulated by op amps, leading to an increased DC offset. If the RF is modulated, then that modulation may appear at the op amp output. CMOS and BiFET op amps can be more resistant to demodulating RF than bipolar devices. Take particular care to keep digital signals away from analog ones, and be generous with power supply decoupling and filtering. Power planes help a great deal in reducing system noise.

The authors chose the TI device TLC2272 which is a low noise CMOS dual op amp designed for single supply operation. It has full rail to rail output swing and low distortion. However the distortion increases when driving loads of about 2 k Ω or less.

2.6.4 Antialiasing Filter

Sigma-delta converters have the advantage of providing anti-aliasing filtering as an integral part of their operation. However this filtering has *holes* in it, at multiples of the oversampling frequency. The AD50 is a 64 times oversampling converter so for a sampling rate of 20 kHz the first *hole* will be centered at 1.28 MHz and will be 20 kHz wide. It is important to ensure that no energy at this frequency is present across the inputs to the codec. A simple single pole RC filter is sufficient. If ceramic capacitors are used for this filter they should be of the COG or NPO dielectric type. Significant distortion can be introduced by the voltage dependent capacitance of other types of dielectric. Two filter configurations are possible, as shown in Figure 2–4.

Figure 2–4. AD50-EVM Antialiasing Filter Options

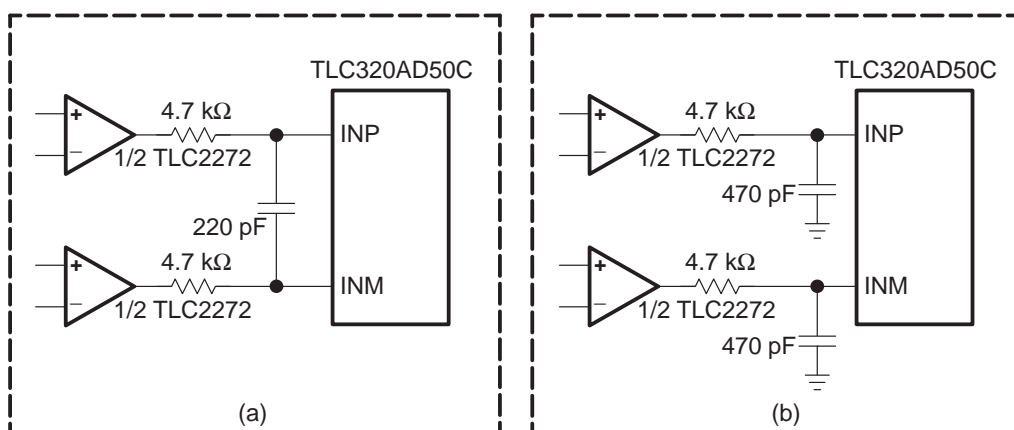


Figure 2–4a gives good rejection of differential noise signals but common mode noise is not removed.

Figure 2–4b gives good rejection of common mode noise signals and, if the capacitors and resistors are well matched, good rejection of differential noise.

Since most noise sources are likely to be coupled equally into both inputs the filter in Figure 2–4b is usually the best choice (the evaluation board has footprints to allow for either or both configurations, with the capacitor in Figure 2–4a not connected).

The AD50 allows higher source impedances at its inputs than many sigma-delta converters, easing the drive requirements.

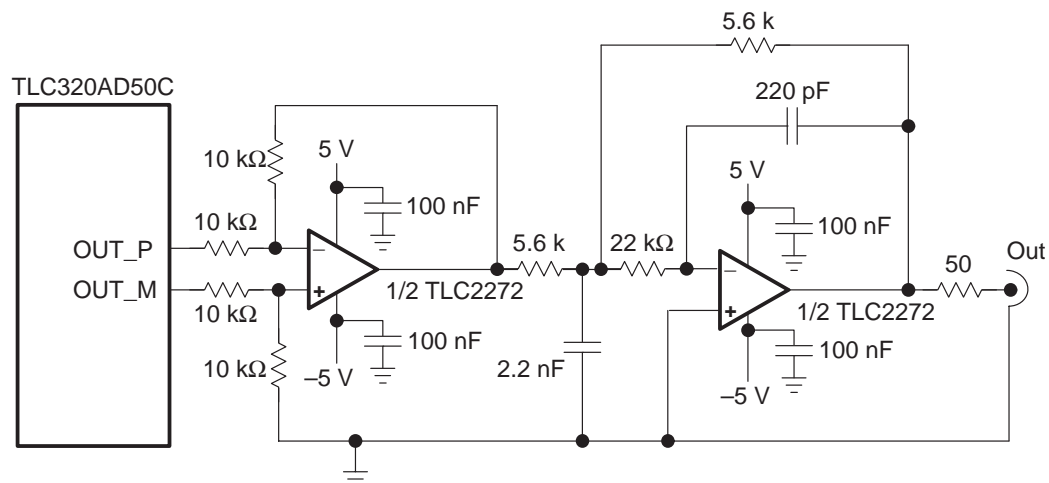
The aux inputs also have antialiasing filters connected, allowing the use of these inputs from external sources without any special filtering requirements.

The antialiasing filter components are placed as close as possible to the inputs to the AD50 to reduce the possibility of noise pickup between the filter and the ADC. The use of 0603 footprint surface-mount components makes this possible.

2.7 Analog Output

The AD50 uses a pair of differential voltage outputs. The output stage needs to convert the differential signals to a single ended output, and to attenuate noise outside the pass-band. This filter is not a conventional reconstruction filter, since the AD50 has an integral low-pass reconstruction filter. The filter is a 12-kHz second order low-pass filter designed for use at 20 ksp/s and will be less effective at lower sampling rates. Since this filter only removes out of band noise it will be unnecessary in many applications.

Figure 2–5. AD50-EVM Differential to Single-Ended Output Converter



Setting Up the AD50-EVM

Setting up the AD50-EVM normally involves coupling it to the synchronous serial port of a digital signal processing system. Two configurations are supported. The first is to couple the AD50-EVM to a DSK+ development system by means of four connectors directly linking the circuit boards together. The DSK+ consists of a TMS320C542 16-bit fixed point digital signal processor combined with a 14-bit analog interface circuit and PC printer port compatible data interface. This allows the downloading and debugging of C or assembler programs from a DOS or Windows environment. Many example programs are supplied with the DSK+, which can readily be adapted to interface to the AD50-EVM. (See Appendix C for detailed information on adapting example programs.) The second configuration is standalone mode where a ribbon cable up to 1 m long connects the AD50-EVM to a separate DSP system. In standalone mode some options such as the number of active AD50s must be set using jumpers or by control signals from the remote processor. In DSK+ mode, almost everything is set from software.

Topic	Page
3.1 Setting Up for the DSK+	3-2
3.2 Setting Up for Stand-Alone Mode	3-2
3.3 Configuring the Serial Interface	3-2
3.4 Programming the AD50 Registers	3-3
3.5 Sine-Wave Generator and Loopback Program	3-4

3.1 Setting Up the AD50-EVM With the DSK+

The first step is to couple the two circuit boards together using four inter-board connectors. See Appendix A for full details. The AD50-EVM takes its power directly from the DSK+. Inputs and outputs are provided on stereo 3.5 mm jacks. Note that in single channel mode, the active device connects to the ring contact on the jack. This is conventionally the left channel on stereo headphones. Attaching the AD50-EVM to the DSK+ does not affect the operation of the AC01 on the DSK+ because the reset default is for both AD50 devices to be powered down and in slave mode. The master clock source can be the 10.00 MHz oscillator supplied with the DSK+, the 10.24 MHz oscillator provided with the AD50-EVM or any other oscillator module within the capabilities of the DSK+ (see section 2.3 for details). It does not matter which socket is used, so long as only one oscillator module is present in the system. Check that the operation of the SELFTEST and example programs has not been affected by adding the AD50-EVM or the oscillator selection.

Either type in and assemble a program such as the sine-wave generator listed below, or modify an appropriate example program such as OSCOPE supplied with the DSK+. Appendix C shows how OSCOPE was modified to work with the AD50-EVM.

3.2 Setting Up the AD50-EVM in Stand-Alone Mode

In stand-alone mode the AD50-EVM needs to be connected to a system with a suitable DSP serial interface. This should include a means of controlling the RESET signal. The connector marked SERIAL has all the necessary signals, with interleaved grounds. The POWERDOWN and other control signals are available on the CONTROL connector. In stand-alone mode all the devices are powered up by default.

A power supply of 5 V at approximately 150 mA needs to be supplied to the positive (+) and 0 pins of the PSU connector. The negative (–) pin is an output from the negative voltage converter.

In stand-alone mode an oscillator of up to 22.579 MHz needs to be inserted in the socket on the AD50-EVM. A 10.24 MHz oscillator is supplied.

Configuration of the registers in the AD50s is software controlled via the serial interface, as described for the DSK+ mode. However the RESET, POWERDOWN and FC signals need to be set via the CONTROL connector.

3.3 Configuring the Serial Interface

First, the four least significant bits of I/O address 0 must be configured with the value 2. This selects one AD50 and disables the AC01, holding the AD50 in reset. Note that the AC01 on the DSK+ is reset by manipulating the XF output of the C542, whereas the AD50-EVM implements the software reset control as bit 0 of I/O address 0. This frees the XF bit for other uses, such as driving

an oscilloscope for debugging purposes. Then the 'C542 serial port is initialized to 16-bit mode with external frame sync pulses. The AD50 is then taken out of reset by writing 3 to I/O address 0. The AD50 starts sending frame sync pulses and a serial clock signal to the 'C542, which responds by transmitting data from the serial port. This data consists of a list of alternating primary and secondary transmissions. The primary transmissions are dummy words with bit 0 set to 1 to request the interleaved secondary transmissions. Each secondary transmission is a 16-bit word containing commands to initialize the four control registers of the AD50.

For each secondary transmission, bits 12-8 encode the binary address of the register to be accessed, bit 13 specifies a read or write access and bits 7-0 contain the data to be written for write operations.

3.4 Programming the AD50 Registers

Register 0 is a dummy register, equivalent in principle to a NOP instruction. Writing to it does nothing. It is needed so that when several AD50s are cascaded it is possible to modify the control registers of one device without disturbing the others.

Register 1 controls reset and power-down status, input selection, monitor amplifier signal source and gain, digital loop back test mode and 15 or 16 bit DAC mode. Once 16 bit mode has been selected, it is not possible to request further secondary communications by setting bit 0 of a serial port data word. However, the AD50-EVM provides a means for driving the FC pin of the AD50 chip by writing to bit 0 of I/O address 1. The value written is applied to the FC pin, and a 1 initiates a hardware secondary communication request. FC is sampled by the AD50 on the rising edge of frame sync.

Register 2 allows control of the flag bit, reading of the decimator overflow flag, selection of 15 or 16 bit ADC mode and control of the analog loop back test mode.

Register 3 is used to select the number of cascaded AD50 devices (up to a maximum of 1 master and 3 slaves) and to control the time between communications from each device. It is important that register 3 is only programmed after the sampling frequency has been set. It is convenient to program everything common to all the devices prior to setting register 3. Thereafter, when secondary communications are requested, each device takes part in turn and must either be programmed individually or a NOP instruction sent to register 0.

Register 4 controls the analog input and output gains and the sampling frequency.

It is only necessary to reprogram those registers that have unsuitable default values. If none of the AD50 registers are programmed, the system will default to a sampling frequency of 10 kHz for a MCLK frequency of 10.24 MHz. If I/O address 0 is not programmed as described above, the system will default to using the AC01 ADC and DAC as normal.

3.5 Sine-Wave Generator and Loopback Program

The following code is for a very simple sine-wave generator and loop back program that demonstrates how the DSK+ can be programmed to drive the AD50-EVM. The sine-wave is generated by means of a lookup table which is cyclically indexed. This program was used to generate the test waveforms for the DAC distortion measurements in section 4 below. The sine table delivers an output waveform at -3dB relative to clipping, to allow direct comparison with the levels used in the AD50 data book.

This program (AD50SIN.ASM) generates a sine-wave when the assembler variable *makesine* is set to 1. Otherwise it reads data from the ADC and copies it to the DAC. Filtering code can be inserted if desired. The variable *mode16bit* specifies whether the least significant bit of data written to the DAC is masked out.

The program should be edited using a text editor (such as MSDOS EDIT) and assembled as follows:

DSKPLASM AD50SIN

This generates a binary object file called AD50SIN.OBJ which is loaded into the DSK+ as follows:

LOADAPP -A AD50SIN.OBJ

```
.title "Sine-wave generator for AD50-EVM and C54x DSK+"
.width 80
.length 55
; Adapted from DSK+ example programs by John Walliker and Julian Daley
;                                     June 1997
.mmregs
.setsect ".text",0x1800,0      ; these assembler directives specify
.setsect ".data",0x0200,1     ; the absolute addresses of different
.setsect "vectors",0x0180,0   ; sections of the assembly code
makesine      .set 1
model6bit     .set 0
sinestepsize  .set 1
sinetablesize .set 20          ; for 1kHz at 20 ksps

; The vectors in this table can be configured for processing external and
; internal software interrupts. The DSKplus debugger uses four interrupt
; vectors. These are RESET, TRAP2, INT2, and HPIINT.
; * DO NOT MODIFY THESE FOUR VECTORS IF YOU PLAN TO USE THE DEBUGGER *
;
; All other vector locations are free to use. When programming always be sure
; the HPIINT bit is unmasked (IMR=200h) to allow the communications kernel and
; host PC interact. INT2 should normally be masked (IMR(bit 2) = 0) so that the
; DSP will not interrupt itself during a HINT. HINT is tied to INT2 externally.
.sect "vectors"
reset    goto #80h      ;00; RESET * DO NOT MODIFY IF USING DEBUGGER *
        nop
        nop
```

```

nmi    return_enable ;04; non-maskable external interrupt
        nop
        nop
        nop
trap2   goto #88h      ;08; trap2 * DO NOT MODIFY IF USING DEBUGGER *
        nop
        nop
        .space 52*16    ;0C-3F: vectors for software interrupts 18-30
int0    return_enable ;40; external interrupt int0
        nop
        nop
        nop
int1    return_enable ;44; external interrupt int1
        nop
        nop
        nop
int2    return_enable ;48; external interrupt int2
        nop
        nop
        nop
tint    return_enable ;4C; internal timer interrupt
        nop
        nop
        nop
brint   return_enable ;50; BSP receive interrupt
        nop
        nop
        nop
bxint   return_enable ;54; BSP transmit interrupt
        nop
        nop
        nop
trint   dgoto receive  ;58; TDM receive interrupt
        nop
        nop
txint   return_enable ;5C; TDM transmit interrupt
        nop
        nop
        nop
int3    return_enable ;60; external interrupt int3
        nop
        nop
        nop
hpiint  goto #0e4h     ;64; HPIint * DO NOT MODIFY IF USING DEBUGGER *
        nop
        nop
        .space 24*16    ;68-7F; reserved area
        .data
sineindex .word 0      ; index to sine-wave table
sinevalue .word 0      ; workspace for sine-wave generation

```

```
initwork .word 0      ; workspace for initialization
        .text
start:   intm = 1      ; disable all interrupts
        pmst = #01a0h ; Interrupt pointer maps vectors to address 180h
        sp = #0ffah   ; stack pointer located in Communications Kernel
        imr = #240h   ; unmask TDM RINT and HPIINT(host port interface)
        OVM = 0       ; OVM = 0 (do not clip overflow)
        ASM = #0      ; ASM = 0 (shift=0)
        DP = #0
AC01INIT:

; initialize serial port and ad50-evm on-board logic
        arl = #initwork ; workspace for storing value to be output to ad50-evm
        *arl = #0010b   ; switch on one AD50 and hold it in reset
        port(0) = *arl  ; write the data to i/o port 0
        repeat(#23)    ; hold reset low for at least 6 mclk cycles
        nop            ; (4 dsp clock cycles == 1 mclk cycle)
        tcr = #10h     ; stop timer
        imr = #240h    ; unmask RXINT and HPIINT
        tspc = #0008h  ; stop TDM serial port
        tdxr = #0h     ; send 0 as first xmit word
        tspc = #00c8h  ; reset and start TDM serial port
        *arl = #0011b  ; bring ad50 out of reset with one channel powered up
        port(0) = *arl

; initialize ad50 registers
        a = #0000010010010000b ; register 4
        call ad50init           ; change fsamp to 20kHz
a = #000000010110000000b ; register 4
        call ad50init           ; light led
;   a = #000000001000000001b ; select 16 bit dac mode
;   call ad50init             ; this must be the last command

        intm = 0              ; enable all interrupts

WAIT    idle(1)              ; idle waiting for interrupts
        nop
        goto WAIT
;-----
; ad50 init routines here
ad50init
        ifr = #080h          ; clear flag from IFR
        tdxr = #01h          ; request secondary when AD50 starts
        call waitfortdx      ; wait for primary to xmit
        tdxr = a              ; send register value to serial port
        ifr = #080h          ; clear flag from IFR
        call waitfortdx      ; wait for secondary to xmit
        tdxr = #0h           ; send neutral state in case last init
        ifr = #080h          ; clear flag from IFR
        call waitfortdx      ; wait for neutral state to xmit
```



```

    return                ; return from subroutine
waitfortdx               ; a robust way of waiting for a tdm port
    b = mmr(ifr)          ; transmission to complete
    b &= #10000000b
nop
nop
    if(BEQ) goto waitfortdx
    return
;-----
receive
                                ; save context if necessary
    B = mmr(TRCV)             ; always read serial port to avoid buffer overflow
                                ; which would stop interrupts from being generated
.if makesine = 0               ; don't make a sinewave - just echo data from adc to dac
                                ; put code for filtering etc here
.else                           ; make a sinewave on LEFT channel
    AR1 = #sineindex
    AR2 = #sinevalue
    A = *AR1
    A += #sinestepsize
    B = A
    A -= #sinetablesize
    nop                        ; two nops to allow for pipeline delay
    nop
    if (ALT) execute(1) ; check whether index points outside table
    A = B

    *AR1 = A

    A += #sinetable20_3dB
    *AR2 = prog(A)             ; get sinewave value from table in program memory
    B = *AR2                   ; (could use sine table in rom for this)
.endif
.if model6bit = 0
    B &= #111111111111110b ; mask least significant bit if in 15 bit mode to
.endif                        ; prevent inadvertent secondary communication requests

    mmr(TDXR) = B

                                ; restore context if necessary
    return_enable             ; return to waiting loop
;-----
transmit:                  ; Transmit and receive interrupts are inherently
                            ; synchronized so there is no point in using both

    return_enable
;-----
sinetable20_3dB ; -3dB re clipping, 1kHz @ 20 ksps, 20 samples/cycle
.word 0, 7160, 13619, 18745, 22036, 23170, 22036, 18745, 13619, 7160
.word 0, -7160, -13619, -18745, -22036, -23170, -22036, -18745, -13619, -7160
sinetable8_3dB ; -3dB re clipping, 1kHz @ 8 ksps , 8 samples/cycle
.word 0, 16384, 23170, 16384, 0, -16384, -23170, -16384
.end

```

Results Obtained With AD50 EVM

This chapter describes the results possible with the AD50 EVM. Actual measurements are presented.

Topic	Page
4.1 ADC Results	4-2
4.2 DAC Results	4-4

4.1 ADC Results

To evaluate the performance of the ADC channel of the AD50-EVM, FFT measurements were performed.

Figure 4–1. AD50-EVM ADC Distortion Measurement at 8 kps

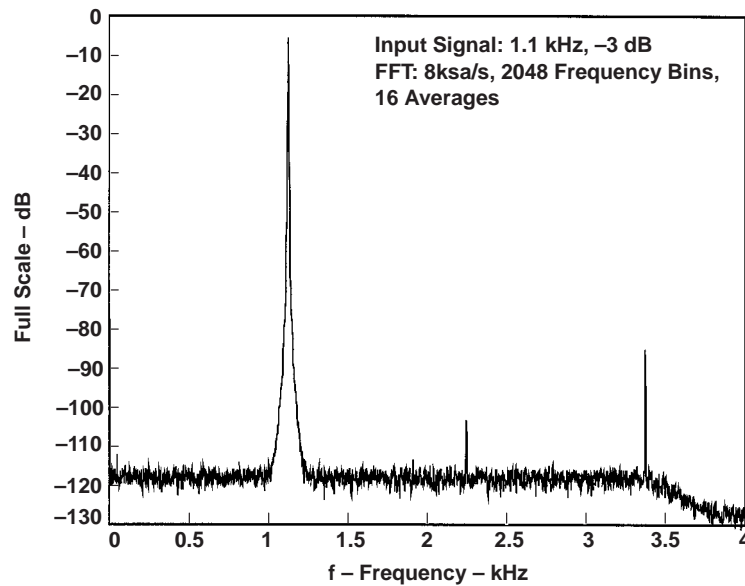


Figure 4–1 shows an FFT plot obtained from the ADC of the AD50-EVM. The input signal was a 1.2-kHz sine wave at -3 dB relative to maximum input. A digital oscillator using a 20 bit DAC was used to produce the test signal. By adding up the energy in each frequency bin within (a) the signal, (b) the harmonics and (c) the rest of the noise floor, figures for SNR, SNR and SINAD can be calculated. The figures were 82.9 dB SNR, 79.4 dB SDR and 77.8 dB SINAD. These figures have not been compensated for noise and distortion components in the oscillator output.

Figure 4–2. AD50-EVM ADC Distortion Measurement at 20 kps

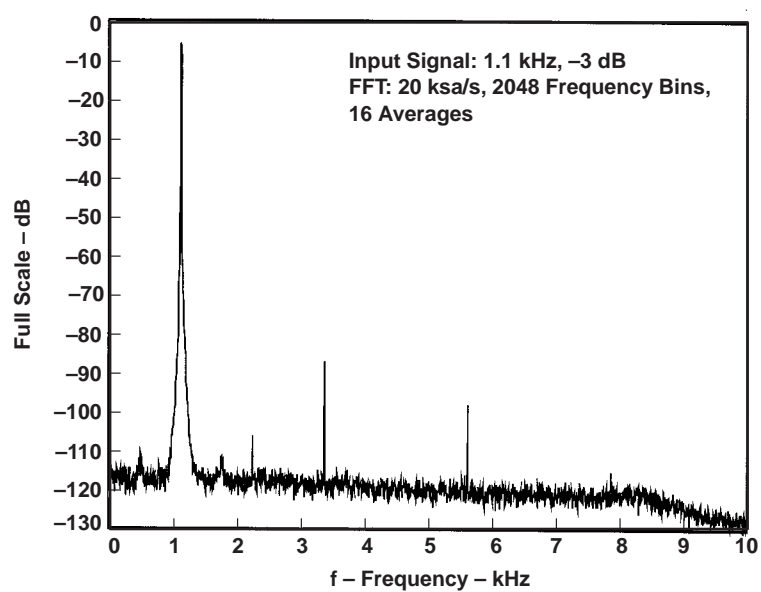


Figure 4–2 shows an FFT plot obtained from the ADC while sampling at 20 kps. The performance figures were 82.7 dB SNR, 81.0 dB SDR and 78.7 dB SINAD.

4.2 DAC Results

The DAC was characterized using a sine-wave table lookup program running on a TMS320C5x DSK+ coupled to an AD50-EVM.

Figure 4–3. AD50-EVM DAC Distortion Measurement at 8 kps

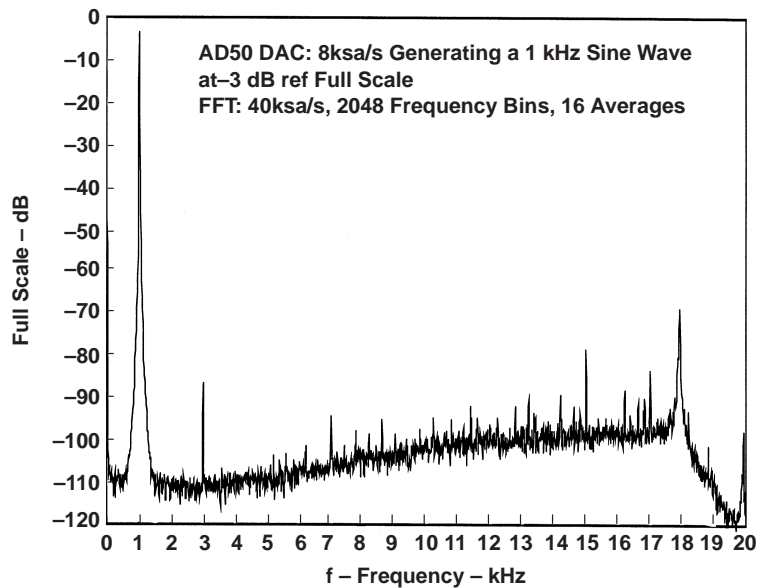
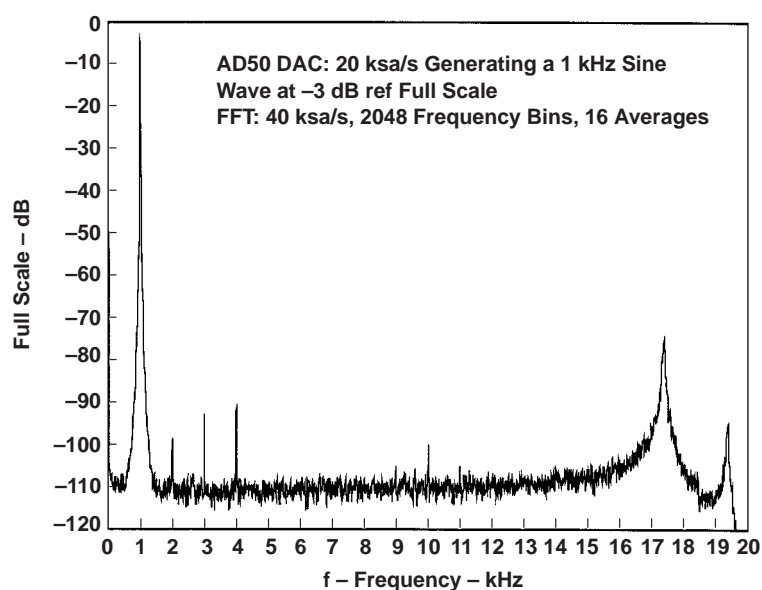


Figure 4–3 shows an FFT plot of the output from the AD50-EVM. This was measured using a Texas Instruments AD55-EVM acquisition system running at 40 kps coupled to a DSP development system in a PC.

The signal to noise plus distortion ratio (SINAD) measured in the pass-band (excluding dc) was 80.5 dB. The SNR was 83.6 dB and the SDR was 83.4 dB. These measurements have not been adjusted for the noise and distortion of the acquisition system.

The rise in the noise floor above 4 kHz is due to the digital noise-shaping filter in the AD50 DAC, which decreases pass-band noise at the expense of increased high frequency noise. This is partially attenuated by the output filter on the AD50-EVM but continues to rise to a plateau at approximately -100 dB. This filter is optimized for use at 20 kps and therefore the out of band noise is worse than at higher sampling frequencies.

Figure 4–4. AD50-EVM DAC Distortion Measurement at 20 kps



Similar measurements were made at 20 kps. The results were 78.6 dB SNR, 84.8 dB SDR and 77.6 dB SINAD. Again this has not been compensated for the characteristics of the acquisition system.

The spike at 10 kHz is caused by breakthrough from the negative supply generator. The larger peaks above 16 kHz are images that have not been completely removed by the reconstruction filters. These are well outside the pass-band of the DAC.

All the above measurements were made using a well-regulated low noise 5 V power supply because the power supply shipped with the DSK+ was found to introduce extra noise.

Installing the AD50-EVM

This chapter discusses the configuration jumper settings for the AD50 EVM, along with guidelines for connecting it to the DSK+. The parts list, schematics, gerber plots, and board outline drawings are included.

Topic	Page
A.1 Board Outline Drawing With Jumper Locations on the AD50 EVM .	A-2
A.2 Connecting the AD50-EVM to the DSK+	A-4
A.3 Parts List	A-4
A.4 Circuit Diagrams	A-7
A.5 PCB Diagrams	A-??

A.1 Board Outline Drawing With Jumper Locations on the AD50 EVM

Figure A–1 shows a diagram of the AD50-EVM board showing locations of connectors and jumpers etc.

Figure A–1. Diagram of the AD50-EVM

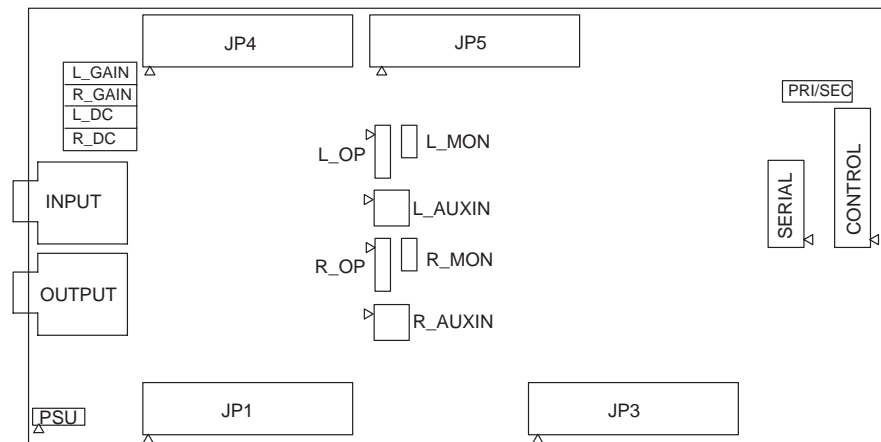


Table A–1. Jumper Positions

	OPEN	CLOSED	DEFAULT
L_DC	Left channel ac input coupling	Left channel dc input coupling	OPEN
R_DC	Right channel ac input coupling	Right channel dc input coupling	OPEN
L_GAIN	Left channel 0 dB input gain	Left channel 20 dB input gain	OPEN
R_GAIN	Right channel 0 dB input gain	Right channel 20 dB input gain	OPEN
PRI/SEC	Primary board	Secondary board	OPEN

The connectors marked SERIAL and CONTROL are for use in interfacing the AD50-EVM to systems other than the DSK+. The pinouts are shown in Table A–2.

If two AD50-EVMs are used the serial connection should be made to the primary board.

Table A–2. Control and Serial Connections

Pin	Control	Serial
1	GROUND	GROUND
2	PRI/SEC	SERIAL DATA IN
3	GROUND	GROUND
4	RIGHT ALTDATA	SERIAL DATA OUT
5	GROUND	GROUND
6	RIGHT $\overline{\text{PWRDN}}$	LEFT FRAME $\overline{\text{SYNC}}$
7	RESET	GROUND
8	LEFT PWRDN	SERIAL CLOCK
9	GROUND	GROUND
10	FC	$\overline{\text{RESET}}$
11	GROUND	
12	RIGHT FLAG	
13	GROUND	
14	LEFT FLAG	
14	GROUND	
16	LEFT ALTDATA	

There are 4 sets of connectors that provide direct access to the analog I/O pins of the AD50. The pinouts for these are shown in Table A–3.

Table A–3. Analog Input/Output Connectors

Label	Aux ip	Diff-op	Mon out
Left	L_AUXIN	L_OP	L_MON
Right	R_AUXIN	R_OP	R_MON

Pin	Aux ip	Diff-op	Mon out
1	AUXP	OUTP	MONOUT
2	1.25 V	GROUND	GROUND
3	AUXM	OUTM	
4	GROUND		

There is a set of pins labeled PSU provided for connecting power supplies. The locally generated –5 V supply can be accessed via this connector.

A.2 Connecting the AD50-EVM to the DSK+

The two boards are stacked together using plug and socket connectors. Four pairs of connectors are needed, in JP2, JP3, JP4, and JP5.

The boards can be stacked either way round, although a slightly wider spacing is needed if the AD50-EVM is on top of the DSK+, due to the height of the audio jack connectors used on the DSK+. Suitable connectors that will allow the boards to be stacked either way with a spacing of approximately 13 mm, can be obtained from the following supplier.

Manufacturer	Plug	Socket
Samtec Tel: 1-800-SAMTEC-9 (USA) +44 1236 739292 (UK)	TSW-112-14-T-T	SSW-112-21-G-T

In order to ensure that the connectors are properly aligned it is suggested that the plugs and sockets be mated before being soldered to the boards. To allow development work with the AD50-EVM it is suggested that the sockets be soldered to the bottom of the AD50-EVM and the pins to the top of the DSK+. This puts the AD50-EVM on top of the DSK+ and provides good access to the signals on the AD50-EVM

If two AD50-EVM boards are being used in conjunction with a DSK+, then a different socket, with pass through pins, should be used on the middle board. The Samtec part number is SSQ-112-24-G-T.

A.3 Parts List

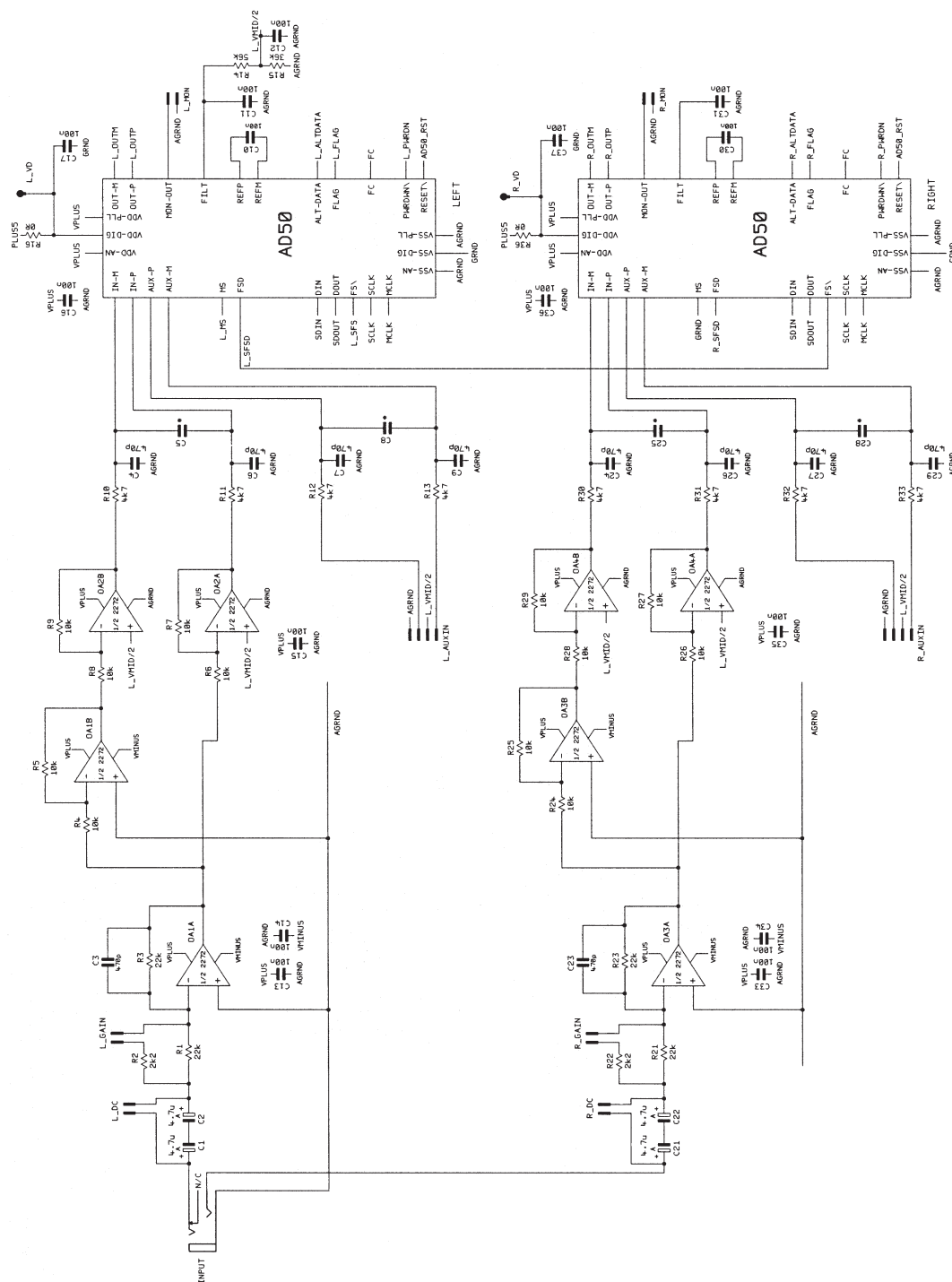
Silk Screen Name	Component	Silk Screen Name	Component
LEFT	TLC320AD50CDW	OA6	TLC2272ACD
RIGHT	TLC320AD50CDW	Q1	BC848 SOT23
GAL1	CMOS 22V10 25 ns, PLCC	Q2	BC848 SOT23
INV	ICL7660 8 pin SO	10–24 MHZ	10.24 MHz 8-pin format TTL or CMOS oscillator module
ON	SM LED 2 mm × 1.25 mm green	R1	22K (all resistors 1% 0603)
L	SM LED 2 mm × 1.25 mm red	R2	(2.2 k) 2K2 [†]
R	SM LED 2 mm × 1.25 mm red	R3	22K
OA1	TLC2272ACD	R4	10K
OA2	TLC2272ACD	R5	10K
OA3	TLC2272ACD	R6	10K
OA4	TLC2272ACD	R7	10K
OA5	TLC2272ACD	R8	10K

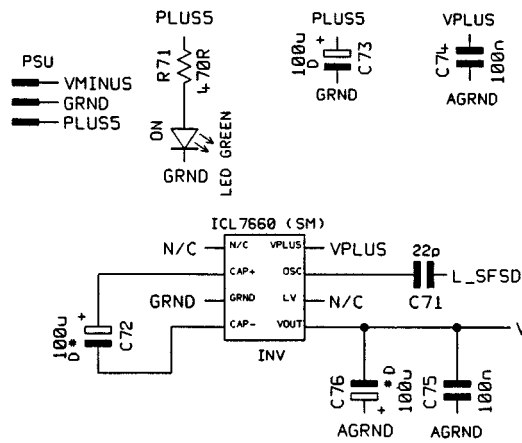
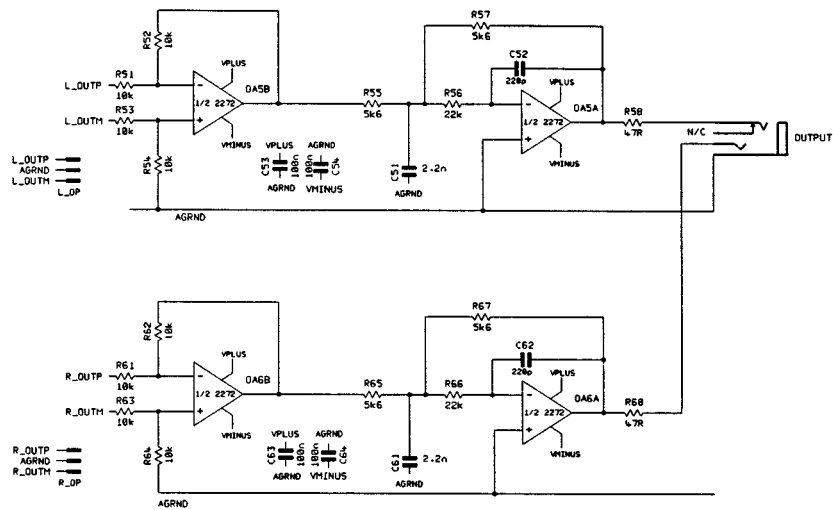
[†] British notation used in schematics A–4.

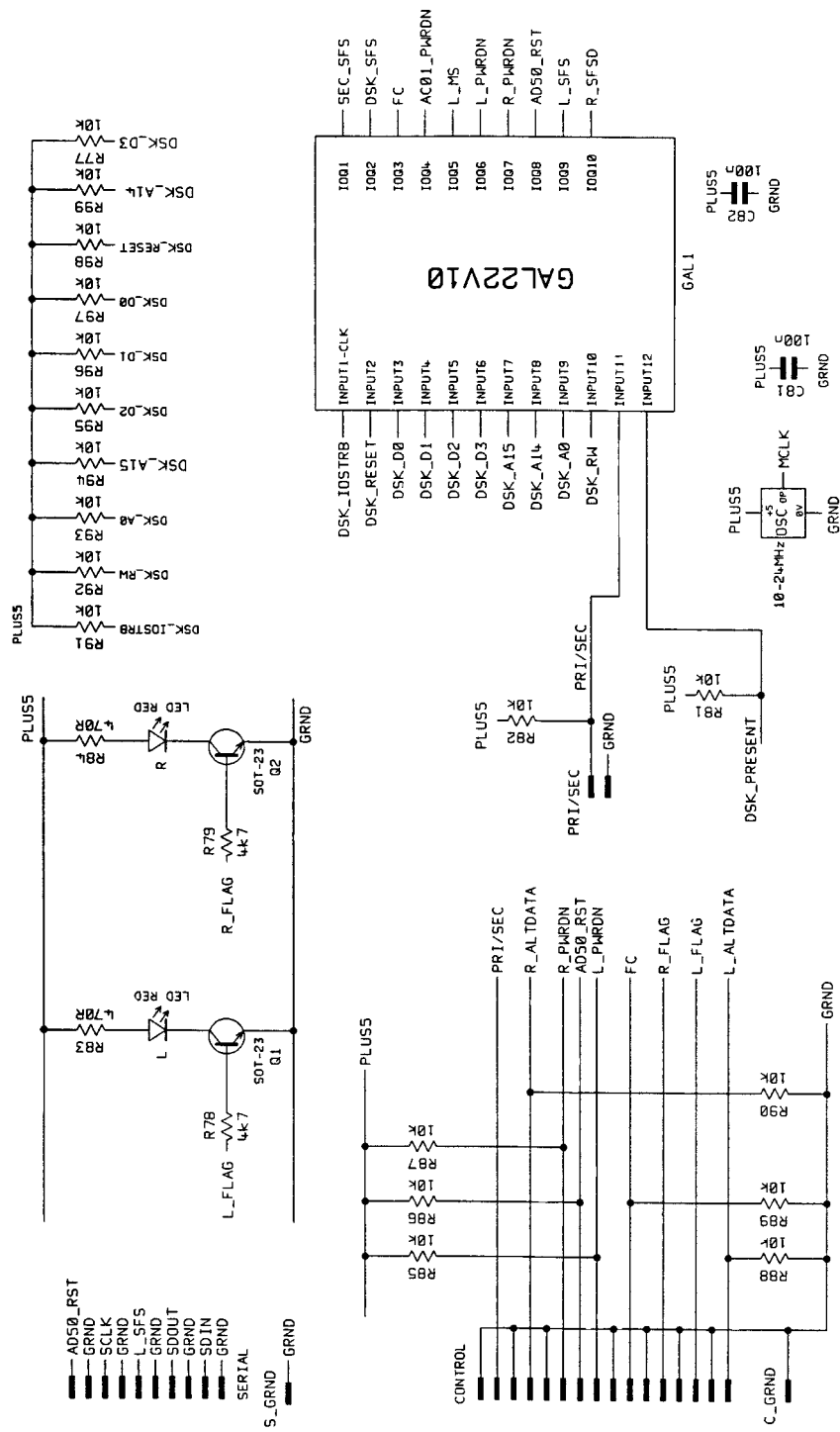
Silk Screen Name	Component	Silk Screen Name	Component
R9	10K	R58	(47) 47R [†]
R10	(4.7 k) 4K7 [†]	R61	10K
R11	(4.7 k) 4K7 [†]	R62	10K
R12	(4.7 k) 4K7 [†]	R63	10K
R13	(4.7 k) 4K7 [†]	R64	10K
R14	56K	R65	12K
R15	39K	R66	18K
R16	0 Ω link	R67	12K
R21	22K	R68	47R
R22	(2.2 k) 2K2 [†]	R71	(1.8 k) 1K8 [†]
R23	22K	R77	10k
R24	10K	R78	(4.7 k) 4K7 [†]
R25	10K	R79	(4.7 k) 4K7 [†]
R26	10K	R80	(4.7 k) 4K7 [†]
R27	10K	R81	10k
R28	10K	R82	10k
R29	10K	R83	(470) 470R [†]
R30	(4.7 k) 4K7 [†]	R84	(470) 470R [†]
R31	(4.7 k) 4K7 [†]	R85	10k
R32	(4.7 k) 4K7 [†]	R86	10k
R33	(4.7 k) 4K7 [†]	R87	10k
R36	0 Ω link	R88	10k
R51	10K	R89	10k
R52	10K	R90	10k
R53	10K	R91	10k
R54	10K	R92	10k
R55	12K	R93	10k
R56	18K	R94	10k
R57	12K	R95	10k
R96	10k	C30	100 nF X7R 0805
R97	10k	C31	100 nF X7R 0805
R98	10k	C33	100 nF X7R 0805
R99	10k	C34	100 nF X7R 0805

[†] British notation used in schematics A–4.

Silk Screen Name	Component	Silk Screen Name	Component
C1	4.7 μ F 10 V size A tantalum	C35	100 nF X7R 0805
C2	4.7 μ F 10 V size A tantalum	C36	100 nF X7R 0805
C3	470 pF COG/NPO 0603	C37	100 nF X7R 0805
C4	470 pF COG/NPO 0603	C51	2.2 nF COG/NPO 0805
C5	470 pF COG/NPO 0603	C52	220 pF COG/NPO 0603
C6	470 pF COG/NPO 0603	C53	100 nF X7R 0805
C7	470 pF COG/NPO 0603	C54	100 nF X7R 0805
C8	470 pF COG/NPO 0603	C61	2.2 nF COG/NPO 0805
C9	470 pF COG/NPO 0603	C62	220 pF COG/NPO 0603
C10	100 nF X7R 0805	C63	100 nF X7R 0805
C11	100 nF X7R 0805	C64	100 nF X7R 0805
C12	100 nF X7R 0805	C71	22 pF COG/NPO 0603
C13	100 nF X7R 0805	C72	100 μ F 10 V size D tantalum
C14	100 nF X7R 0805	C73	100 μ F 10 V size D tantalum
C15	100 nF X7R 0805	C74	100 nF X7R 0805
C16	100 nF X7R 0805	C75	100 nF X7R 0805
C17	100 nF X7R 0805	C76	100 μ F 10V size D tantalum
C21	4.7 μ F 10 V size A tantalum	C81	100 nF X7R 0805
C22	4.7 μ F 10 V size A tantalum	C82	100 nF X7R 0805
C23	470 pF COG/NPO 0603	INPUT	Marushin MJ156L 3.5 mm stereo pcb jack socket
C24	470 pF COG/NPO 0603	OUTPUT	Marushin MJ156L 3.5 mm stereo pcb jack socket
C25	470 pF COG/NPO 0603	PRI/SEC	Right angle header
C26	470 pF COG/NPO 0603	L_DC	Right angle header
C27	470 pF COG/NPO 0603	L_GAIN	Right angle header
C28	470 pF COG/NPO 0603	R_DC	Right angle header
C29	470 pF COG/NPO 0603	R_GAIN	Right angle header





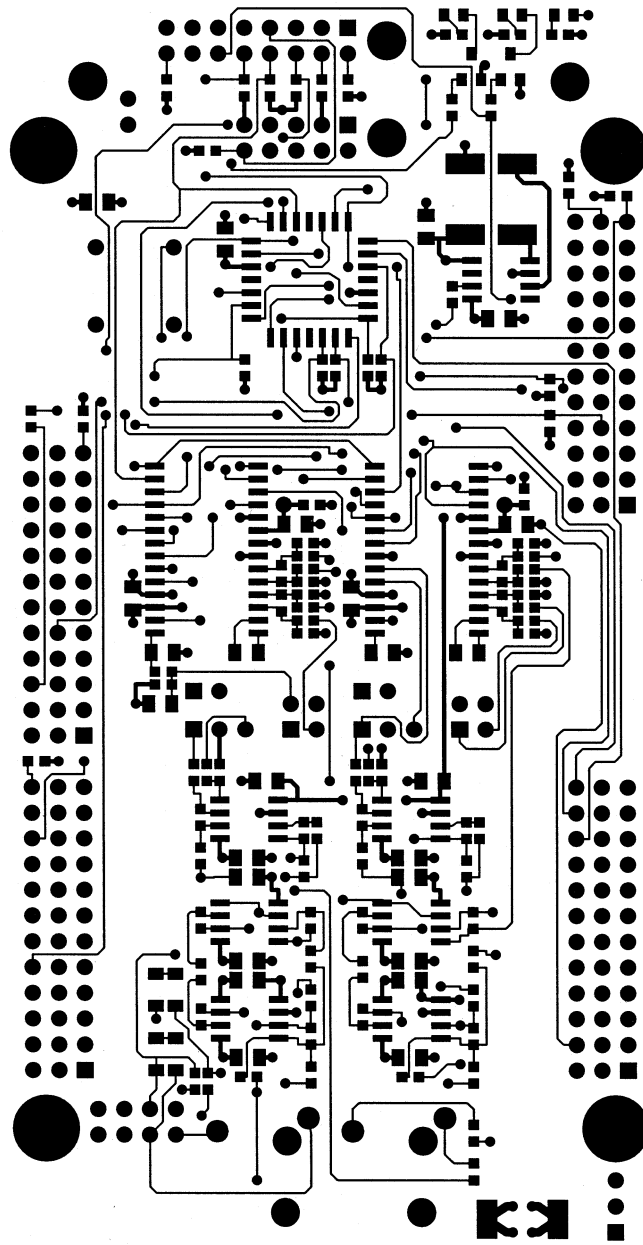




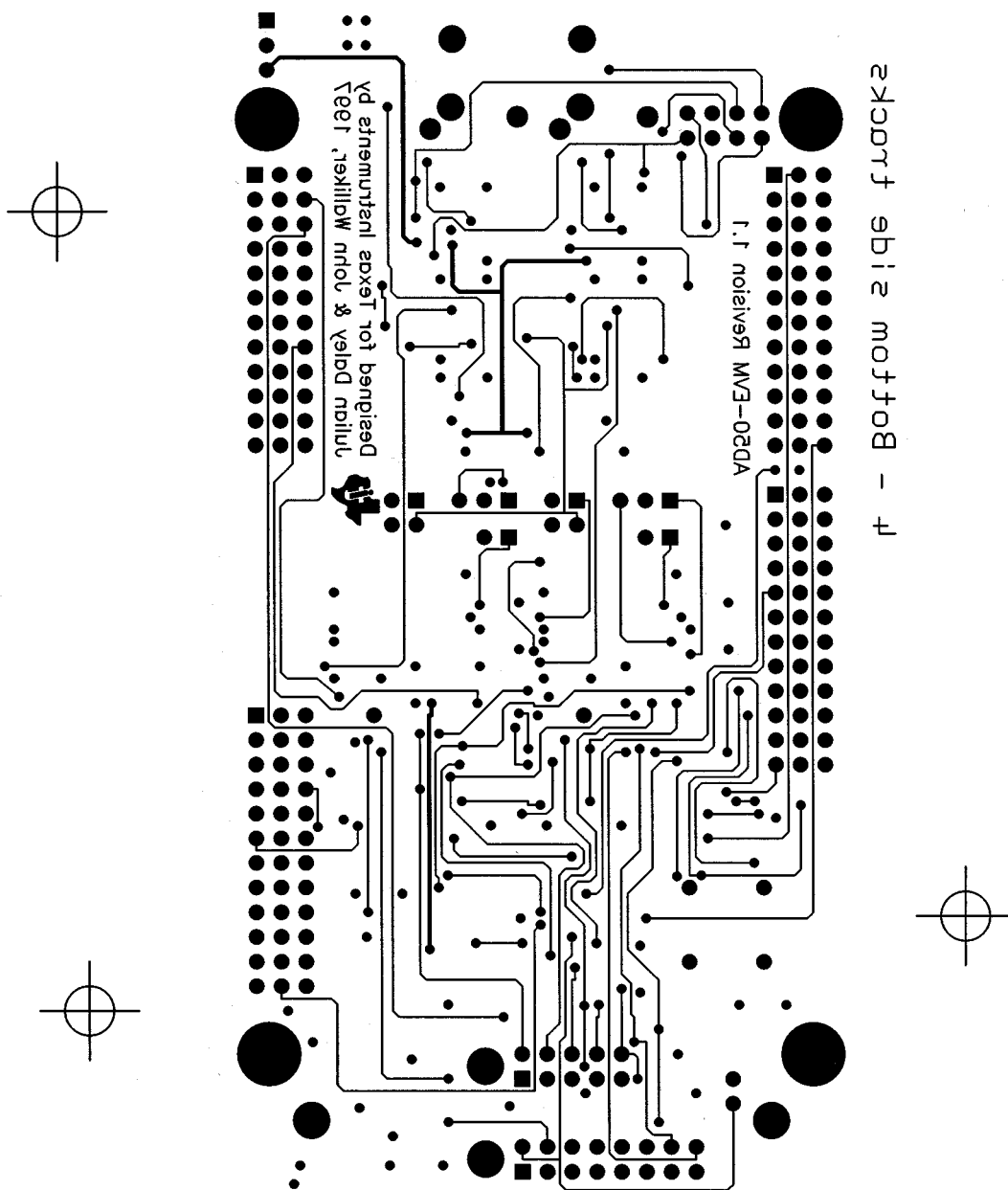


Top Side Tracks

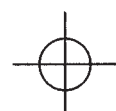
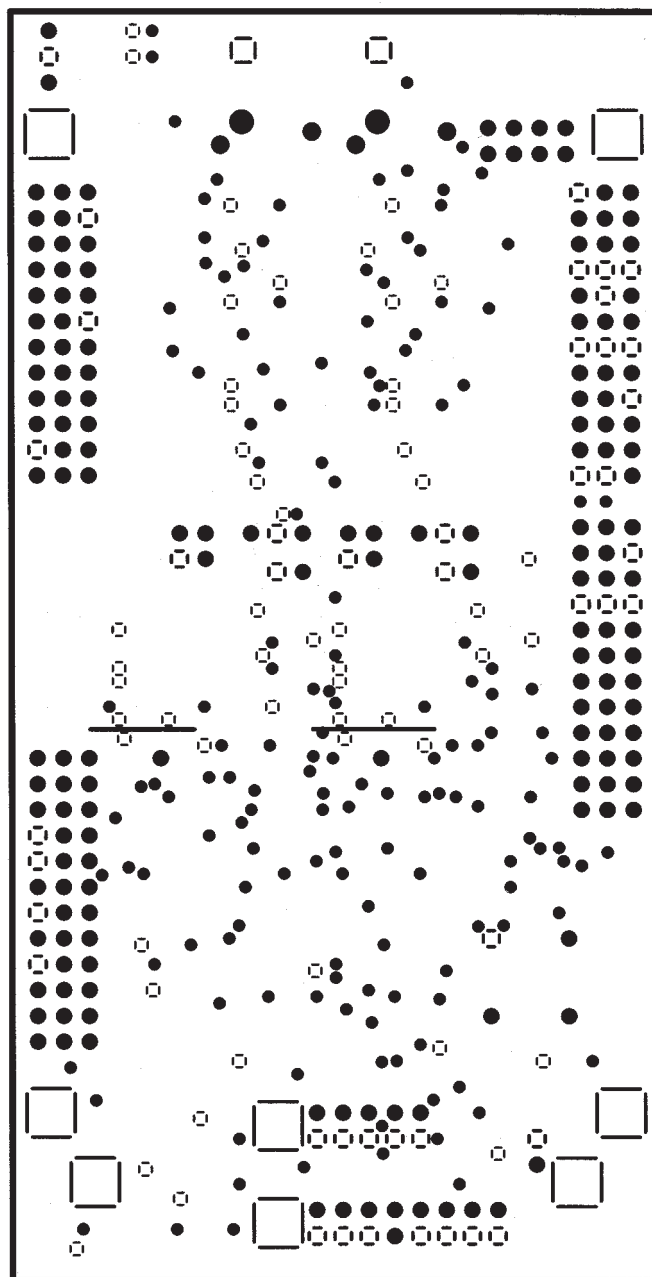
1 – Top Side Tracks



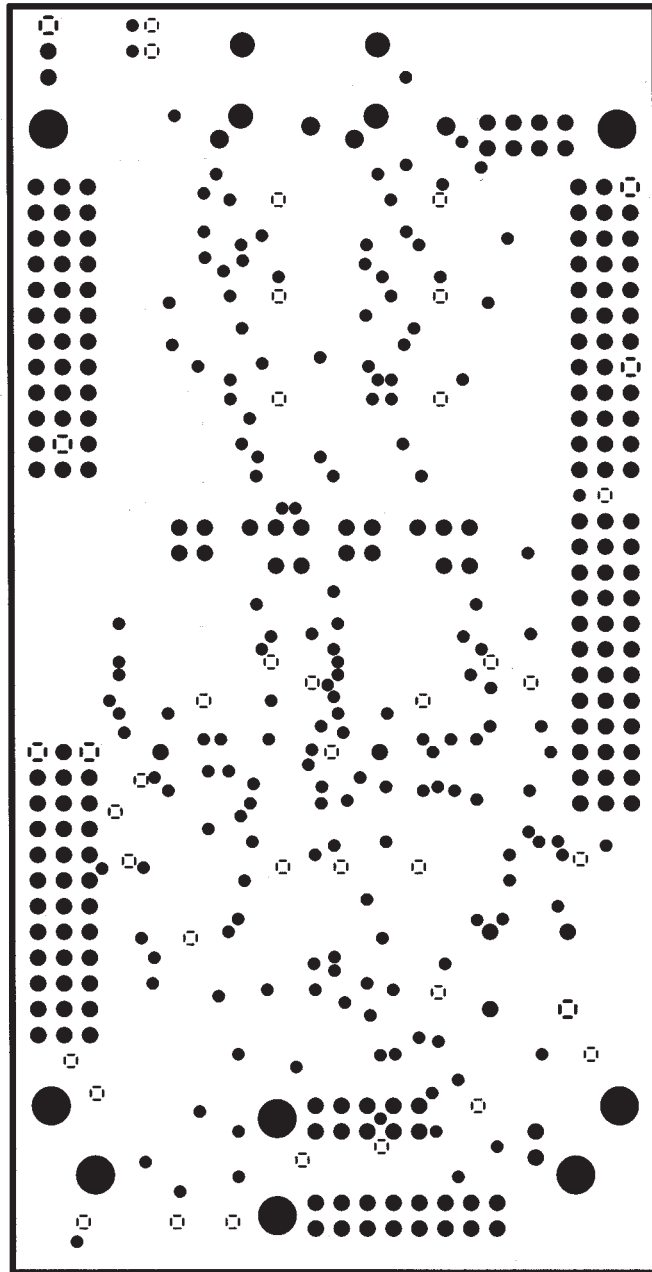
Bottom Side Tracks



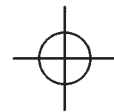
Ground plane



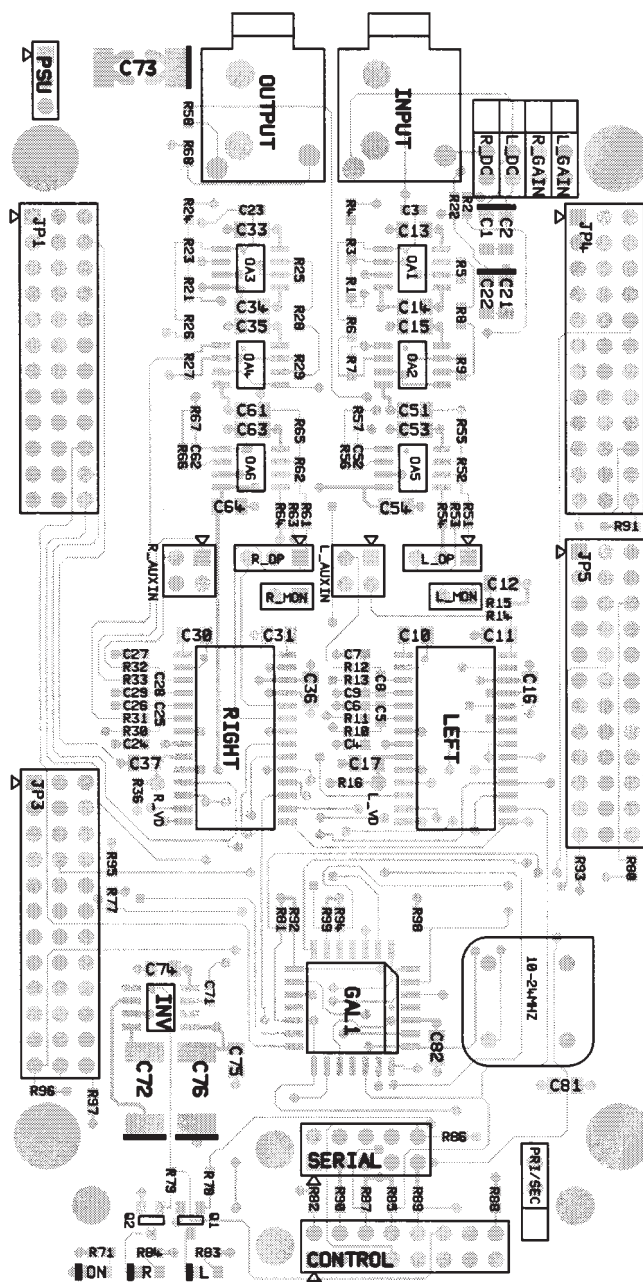
Power plane



3 - power plane



Component positions



Programmable Logic

The 22V10 programmable logic device incorporates all the logic needed by the AD50-EVM when used in stand-alone mode or in conjunction with the DSK+. Note that in many applications little or no logic is required for interfacing the AD50 to a DSP system.

When used in stand-alone mode, most of the 22V10 outputs are high impedance to allow an external device complete freedom of control. This is controlled by the state of the DSK_PRESENT signal, which is pulled low when the AD50-EVM is connected to a DSK+. Otherwise a pull-up resistor makes it default to stand-alone mode.

One or two AD50-EVM boards can provide 1, 2, 3, or 4 simultaneously sampled channels. When two boards are coupled together the primary/secondary jumper controls the daisy-chaining of serial frame sync pulses and controls the way that data from the DSK+ is interpreted.

Designers familiar with the AC01 should note that the power-down behavior of the AD50 serial interface is different. The AC01 clock and frame sync outputs are always high impedance in power-down mode. The AD50 clock and frame sync outputs are driven high in power-down master mode, but are high impedance in power-down slave mode.

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B.2 Stand-Alone Mode	B-2
B.3 Logic Compiler Listing	B-2

B.1 DSK+ Mode

Address bits 0, 14, and 15 are decoded in conjunction with the I/O space select and the write enable from the C542 on the DSK+. Signals on the data bus are latched on the rising edge of the *iostrobe* signal if a valid write address has been decoded. Bits 14 and 15 are decoded so as to leave 48 k words of contiguous I/O address space free for other hardware that the user might wish to interface to the system. Bit 0 selects one of two control registers implemented with D latches within the 22V10 device. The first register is used to control the device reset and *powerdown* inputs for both codecs on each board and to control the powerdown signal to the TLC320AC01 codec on the DSK+ board. The second register just controls the FC signal. This is to simplify programming of interleaved secondary communications when the DAC is in 16 bit mode. When two AD50-EVM boards are used a jumper must be fitted to the primary/secondary header of one of the boards. The board with the jumper becomes the secondary one and has channels 3 and 4.

The 22V10 must have a propagation delay of 25 ns or less, assuming that one or more I/O wait states are programmed in the TMS320C542. Zero wait state operation would have negligible benefit and would need a much faster and more costly device.

The AD50s are reset either when the DSK+ is reset or when a 0 is programmed into bit 0 of address 0. Reset must be held low for at least 6 master clock cycles. The default configuration at reset is for all the AD50s to be powered down and for the AC01 on the DSK+ to be operating normally. This means that DSK+ will continue to operate normally until the 22V10 is programmed for AD50 operation.

B.2 Stand-Alone Mode

In stand-alone mode all the logic except that used to link the serial interface is 3-stated. This allows maximum flexibility to control the AD50s via the CONTROL and SERIAL cable headers.

Table B–1. AD50-EVM Registers

I/O Address	Bits 15–4	Bit 3	Bit 2	Bit 1	Bit 0	Function
0	X	X	X	X	0	Hold AD50 in reset
	X	X	X	X	1	Release AD50 from reset
	X	0	0	0	X	AC01 on DSK+ active
	X	0	0	1	X	1 AD50 active
	X	0	1	0	X	2 AD50s active
	X	0	1	1	X	3 AD50s active
	X	1	0	0	0	4 AD50s active
1	X	X	X	X	0	FC pin low
	X	X	X	X	1	FC pin high

B.3 Logic Compiler Listing

```

TITLE    AD50 programmable address decoder & control port
AUTHOR   John Walliker and Julian Daley
COMPANY  For Texas Instruments
DATE     14 March 1997, 6 April 1997
OPTIONS

EXPAND = ON

INVERSION = OFF           ; to ensure correct power-up reset state
MINIMIZATION = ON

CHIP     AD50LOGIC        EP22V10FN

;PIN     1      NC0
PIN      2      iostrb      ; rising edge of DSK+ iostrobe clocks the D latches
PIN      3      DSK_reset   ; global reset from DSK+
PIN      4      d0
PIN      5      d1          ; data bus
PIN      6      d2
PIN      7      d3
;PIN     8      NC1
PIN      9      a15
PIN     10      a14          ; address bus
PIN     11      a0
PIN     12      rw          ; read/write control signal
PIN     13      primary     ; pulled low with jumper when secondary board
;PIN     14      GRND
;PIN     15      NC2
PIN     16      DSK_present ; pulled low when DSK+ is connected
PIN     17      sec_sfs     ; secondary serial frame sync
PIN     18      dsk_sfs     ; serial frame sync to DSK+
PIN     19      fc          ; control signal for initiating secondary comms
PIN     20      AC01_pwrdsn ; power down AC01 on DSK+
PIN     21      L_ms        ; left master/slave select
;PIN     22      NC3
PIN     23      L_pwrdsn    ; left power down
PIN     24      R_pwrdsn    ; right power down
PIN     25      AD50_rst    ; reset all AD50s
PIN     26      L_sfs       ; left serial frame sync
PIN     27      R_sfsD      ; right serial frame sync delayed
;PIN     28      VCC

STRING select0 '( /a0 * /a14 * /a15 * /rw)' ; address decode
STRING select1 '( a0 * /a14 * /a15 * /rw)'

EQUATIONS
AD50_rst :=      select0 * d0                ; primary and secondary boards are driven
            + /select0 * AD50_rst
AD50_rst.clkf = iostrb
AD50_rst.rstf = /DSK_reset
AD50_rst.trst = /DSK_present                ; 3-state if in standalone mode
fc          :=      select1 * d0
            + /select1 * fc
fc.clkf     = iostrb
fc.rstf     = /DSK_reset

```

```

fc.trst    = /DSK_present
/AC01_pwrtn := select0 * (d3 + d2 + d1) ; NB The logic assembler MUST NOT in-
vert
           + /select0 * /AC01_pwrtn      ; equations, otherwise power-up reset
                                           ; state will be incorrect!

AC01_pwrtn.clkf = iostrb
AC01_pwrtn.rstf = /DSK_reset
L_pwrtn :=      select0 * /d3 * /d2 * d1 * primary
               + select0 * /d3 * d2 * /d1 * primary
               + select0 * /d3 * d2 * d1
               + select0 * d3 * /d2 * /d1
               + /select0 * L_pwrtn
L_pwrtn.clkf    = iostrb
L_pwrtn.rstf    = /DSK_reset
L_pwrtn.trst    = /DSK_present
L_ms           = primary * L_pwrtn * /DSK_present
               + primary * DSK_present
R_pwrtn :=      select0 * /d3 * d2 * /d1 * primary
               + select0 * /d3 * d2 * d1 * primary
               + select0 * d3 * /d2 * /d1
               + /select0 * R_pwrtn
R_pwrtn.clkf    = iostrb
R_pwrtn.rstf    = /DSK_reset
R_pwrtn.trst    = /DSK_present
DSK_sfs         = L_sfs
DSK_sfs.trst    = primary * /AC01_pwrtn
L_sfs           = sec_sfs
L_sfs.trst      = /primary
sec_sfs         = R_sfsd
sec_sfs.trst    = primary

```

Converting DSK+ Software for the AD50-EVM

The AD50-EVM has been designed to make conversion of DSK+ programs originally written for the AC01 analog interface circuit as easy as possible. The serial port used by the AC01 is shared with two or four AD50s fitted to one or two AD50-EVM boards, but logic ensures that only one type of AIC is used at a time.

For single channel operation the only essential changes to existing code relate to initialization. First, one AD50 must be switched on and held in reset for at least 6 master clock cycles. While it is held in reset the DSP serial port is initialized. The AD50 is then released from reset and the registers of the AD50 are programmed with the required clock division ratio, analog gains etc.

It is not essential to change the main program code for single channel operation, as the AD50 uses the same serial port as the AC01. However, the AC01 is a 14-bit device whereas the AD50 supports 15- or 16-bit operations. For best performance, therefore, masking operations that strip the two least significant bits from data sent to or from the AC01 should be modified to remove only bit 0 or removed completely, depending on whether 15- or 16-bit mode is to be used.

The following code shows how the oscilloscope program (OSCOPE) supplied with the DSK+ was modified to use one AD50 device instead of the AC01. OSCOPE provides an oscilloscope style display of the input waveform on the PC screen. This program is split up into several modules, with all the ADC initialization code in OSC_AC01.ASM.

The C54x algebraic assembly language is used. Code that has been added is shown in **bold**, while code that has been deleted is shown in *italics*. No changes were made to the C code running on the PC for transferring and displaying the acquired data, so this code is not listed below.

```

; *****
; File: OSC_MAIN.ASM -> DSKplus Oscilloscope ASM file for the 'C54x DSKplus
; Adapted to operate with AD50-EVM by John Walliker and Julian Daley
; *****

        .width    80
        .length   55
        .title    "DSKplus Oscilloscope program"
        .mmregs
        .setsect  ".text", 0x500
        .setsect  "vectors", 0x180
;=====
;
;  VECTORS
;
;=====
        .sect  "vectors"
        .copy  "osc_vecs.asm"
        .text
start:
        call  AC01INIT
        pmst = #01a0h          ; set up iptr
        sp = #0ffah            ; init stack pointer.
        ar2 = #1200h           ; pointer to receive buffer.
        *ar2+ = data(#0bh)     ; store to rcv buffer
        imr = #40h
        intm = 0               ; ready to rcv int's
wait    nop
        goto  wait
;      ----- Receive Interrupt Routine -----
RINT:
        b = trcv               ; load acc b with input
        b = #0FFFEh & b        ; only strip out lsb for AD50
        *ar2+ = data(#0bh)     ; store to rcv buffer
        tdxr = b               ; transmit the data.
        TC = (@ar2 == #01300h)
        if (TC) goto restrt     ; stop if rcv buffer is at 1300h
        return_enable
restrt
        ar2 = #1200h           ; set intm bit ...no int's
        hpic = #0ah            ; flag host task completed
        return_enable
;      ----- end ISR -----
        .copy  "osc_ac01.asm"
        .end
;*****
; File: Osc_AC01.ASM -> AC01 Init Routine for the Oscilloscope application
; Adapted to operate with AD50-EVM by John Walliker and Julian Daley
;*****

        .width    80
        .length   55

```

```

        .title "AC01 Initialization Program"
        .mmregs
*****
* Certain AC01 registers can be initialized using a conditional assembly
* constant. By setting the constant REGISTER to the appropriate value, the
* assembler will either include initialization for certain registers or
* ignore register initialization.
*
* The constant REGISTER should be set to include the following AC01 register:
*
* REGISTER (binary) =
*
*      0000 0000 0000 0001  -> initialize Register 1  (A Register)
*      0000 0000 0000 0010  -> initialize Register 2  (B Register)
*      0000 0000 0000 0100  -> initialize Register 3  (A' Register)
*      0000 0000 0000 1000  -> initialize Register 4  (Amplifier Gain-Select)
*      0000 0000 0001 0000  -> initialize Register 5  (Analog Configuration)
*      0000 0000 0010 0000  -> initialize Register 6  (Digital Configuration)
*      0000 0000 0100 0000  -> initialize Register 7  (Frame-Sync Delay)
*      0000 0000 1000 0000  -> initialize Register 8  (Fram-Sync number)
*
* For AD50-EVM
*      0000 0000 0000 0001  -> initialize Register 1
*      0000 0000 0000 0010  -> initialize Register 2
*      0000 0000 0000 0100  -> initialize Register 3
*      0000 0000 0000 1000  -> initialize Register 4
*
* Any combination of registers can be initialized by adding the binary
* number to the REGISTER constant. For example to initialize Registers 4
* and 5, REGISTER = 18h. Upon assembly, only code for register 4 & 5
* initialization is included in the AC01INIT module. When called the
* module will load the REG4 and REG5 values into internal AC01 registers.
*
;;REGISTER .set 09h      ; Powerup default values:
;;REG1 .set 120h        ; *A=32      112h
;;REG2 .set 212h        ;           212h
;;REG3 .set 300h        ;           300h
;;REG4 .set 409h        ; *         405h
;;REG5 .set 501h        ;           501h
;;REG6 .set 600h        ;           600h
;;REG7 .set 700h        ;           700h
;;REG8 .set 801h        ;           801h
;For AD50-EVM:
REGISTER .set 1000b      ; just program reg 4
REG1 .set 0000000100000000b ; default
REG2 .set 0000001000010000b ; select 16-bit ADC mode
REG3 .set 0000001100000000b ; default (no slaves)
REG4 .set 0000010010010000b ; 20ksa/s with 10.24MHz MCLK
; Note that once 16-bit DAC mode has been selected in reg 1, bit 0, other
; registers can only be configured by using FC to request secondary comms. On the

```

```

; AD50-EVM this can be achieved by writing 1 to i/o port 1, bit 0 before a primary
; communication and clearing port 1 afterwards.
; Once slave devices have been specified using reg 3, then a separate secondary
; communication (interleaved with normal communications) takes place for each
; device in turn.

```

```
AC01INIT:
```

```

ar2 = #1302h      ; workspace for storing value to be output to AD50-EVM
*ar2 = #0010b     ; switch on one AD50 and hold it in reset
port(0) = *ar2    ; write the data to i/o port 0
repeat(#50)      ; hold reset low for at least 6 mclk cycles
nop              ; ( 4 DSP clock cycles == 1 MCLK cycle )

```

```

;; xf = 0          ; reset ac01
intm = 1          ; disable all int service routines
tcr = #10h        ; stop timer
imr = #280h       ; wakeup from idle when TDM Xmt int
tspc = #0008h     ; stop TDM serial port
tdxr = #0h        ; send 0 as first xmit word
tspc = #00c8h     ; reset and start TDM serial port

```

```
;; xf = 1          ; release ac01 from reset
```

```
*ar2 = #0011b; bring ad50 out of reset with 1 channel selected
```

```
port(0) = *ar2
```

```
; ----- Register init's -----
```

```

.eval REGISTER & 1h, SELECT ; if REG1 then include this source
.if SELECT = 1h            ;
a = #REG1                  ; load Acc A with REG1 value
call REQ2                  ; Call REQ2 subroutine
.endif

.eval REGISTER & 2h, SELECT ; if REG2 then include this source
.if SELECT = 2h            ;
a = #REG2
call REQ2
.endif

.eval REGISTER & 8h, SELECT ; if REG4 then include this source
.if SELECT = 8h            ;
a = #REG4
call REQ2
.endif

```

```
; Note that AD50 register 3 must be programmed last!
```

```

.eval REGISTER & 4h, SELECT ; if REG3 then include this source
.if SELECT = 4h            ;
a = #REG3
call REQ2
.endif

```

```
; Code for registers not used by AD50 deleted
```

```
return
```

```
REQ2
```

```

ifr = #080h          ; clear flag from IFR
tdxr = #03h          ; request secondary when AC01 starts
call waitfortdx      ; wait for primary to xmit
tdxr = a              ; send register value to serial port

```

```

    ifr = #080h          ; clear flag from IFR
    call waitfortdx      ; wait for secondary to xmit
    tdxr = #0h          ; send neutral state in case last init
    ifr = #080h          ; clear flag from IFR
    call waitfortdx      ; wait for neutral state to xmit
    return               ; return from subroutine
waitfortdx               ; a robust way of waiting for a tdm port
    b = mmr(ifr)          ; transmission to complete
    b &= #10000000b
nop
nop
    if(BEQ) goto waitfortdx
    return
.end
; *****
; File: osc_vecs.ASM -> Vector Table for the 'C54x DSKplus Oscilloscope
;
; *****
        .width    80
        .length   55
        .title "Vector Table"
        .mmregs
reset   dgoto start      ;00; RESET
        IMR = #200h
nmi     return_enable    ;04; non-maskable external interrupt
        nop
        nop
        nop
trap2   return_enable    ;08; trap2
        nop
        nop
        nop
        .space 52*16      ;0C-3F: vectors for software interrupts 18-30
int0     return_enable    ;40; external interrupt int0
        nop
        nop
        nop
int1     return_enable    ;44; external interrupt int1
        nop
        nop
        nop
int2     return_enable    ;48; external interrupt int2
        nop
        nop
        nop
tint     return_enable    ;4C; internal timer interrupt
        nop
        nop
        nop
brint    return_enable    ;50; BSP receive interrupt

```

```

        nop
        nop
        nop
bxint   return_enable   ;54; BSP transmit interrupt
        nop
        nop
        nop
trint   goto   RINT      ;58; TDM receive interrupt
        nop
        nop
txint   return_enable   ;5C; TDM transmit interrupt
        nop
        nop
        nop
int3    return_enable   ;60; external interrupt int3
        nop
        nop
        nop
hpiint  return_enable   ;64; HPIint
        nop
        nop
        nop
.space  24*16           ;68-7F; reserved area

```