

Figure 2. Pinning SSO20 package

Table 1 Pin description

| Name | Type | Function | Alternate Function | Pin-No. | Reset State |
|-----------------|------|------------------------------------|---|---------|-------------|
| V _{DD} | | Supply voltage | — | 1 | NA |
| V _{SS} | | Circuit ground | — | 20 | NA |
| BP10 | I/O | Bidirectional I/O line of Port 1.0 | — | 10 | Input |
| BP13 | I/O | Bidirectional I/O line of Port 1.3 | — | 11 | Input |
| BP20 | I/O | Bidirectional I/O line of Port 2.0 | NTE—test mode enable | 13 | Input |
| BP21 | I/O | Bidirectional I/O line of Port 2.1 | — | 14 | Input |
| BP22 | I/O | Bidirectional I/O line of Port 2.2 | — | 15 | Input |
| BP23 | I/O | Bidirectional I/O line of Port 2.3 | — | 16 | Input |
| BP40 | I/O | Bidirectional I/O line of Port 4.0 | SC—serial clock or INT3 external interrupt input | 2 | Input |
| BP41 | I/O | Bidirectional I/O line of Port 4.1 | VMI voltage monitor input or T2I external clock input Timer 2 | 17 | Input |
| BP42 | I/O | Bidirectional I/O line of Port 4.2 | T2O Timer 2 output | 18 | Input |
| BP43 | I/O | Bidirectional I/O line of Port 4.3 | SD serial data I/O or INT3—external interrupt input | 19 | Input |
| BP50 | I/O | Bidirectional I/O line of Port 5.0 | INT6 external interrupt input | 6 | Input |
| BP51 | I/O | Bidirectional I/O line of Port 5.1 | INT6 external interrupt input | 5 | Input |
| BP52 | I/O | Bidirectional I/O line of Port 5.2 | INT1 external interrupt input | 4 | Input |
| BP53 | I/O | Bidirectional I/O line of Port 5.3 | INT1 external interrupt input | 3 | Input |
| BP60 | I/O | Bidirectional I/O line of Port 6.0 | T3O Timer 3 output | 9 | Input |
| BP63 | I/O | Bidirectional I/O line of Port 6.3 | T3I Timer 3 input | 12 | Input |
| OSC1 | I | Oscillator input | 4-MHz crystal input or 32-kHz crystal input or external clock input or external trimming resistor input | 7 | Input |
| OSC2 | O | Oscillator output | 4-MHz crystal output or 32-kHz crystal output | 8 | NA |

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1 Introduction

The M44C092-H / M44C892-H are members of Atmel's family of 4-bit single-chip microcontrollers. They contain ROM, RAM, parallel I/O ports, two 8-bit

programmable multifunction timer/counters, voltage supervisor, interval timer with watchdog function and a sophisticated on-chip clock generation with integrated RC-, 32-kHz crystal- and 4-MHz crystal-oscillators.

Table 2 provides an overview of the available variants.

Table 2 Available variants of M4xCx9x

| Version | Type | ROM | E2PROM peripheral | Packages |
|--------------|-----------|------------------|-------------------|----------|
| Flash device | T48C893 | 4 Kbyte EEPROM | 64 byte | SSO20 |
| Production | M44C092-H | 4 Kbyte mask ROM | — | SSO20 |
| Production | M44C892-H | 4 Kbyte mask ROM | 64 byte | SSO20 |

2 MARC4 Architecture

2.1 General Description

The MARC4 microcontroller consists of an advanced stack-based, 4-bit CPU core and on-chip peripherals. The CPU is based on the HARVARD architecture with physically separate program memory (ROM) and data memory (RAM). Three independent buses, the instruction bus, the memory bus and the I/O bus, are used for parallel communication between ROM, RAM and

peripherals. This enhances program execution speed by allowing both instruction prefetching, and a simultaneous communication to the on-chip peripheral circuitry. The extremely powerful integrated interrupt controller with associated eight prioritized interrupt levels supports fast and efficient processing of hardware events. The MARC4 is designed for the high-level programming language qFORTH. The core includes both, an expression and a return stack. This architecture enables high-level language programming without any loss of efficiency or code density.

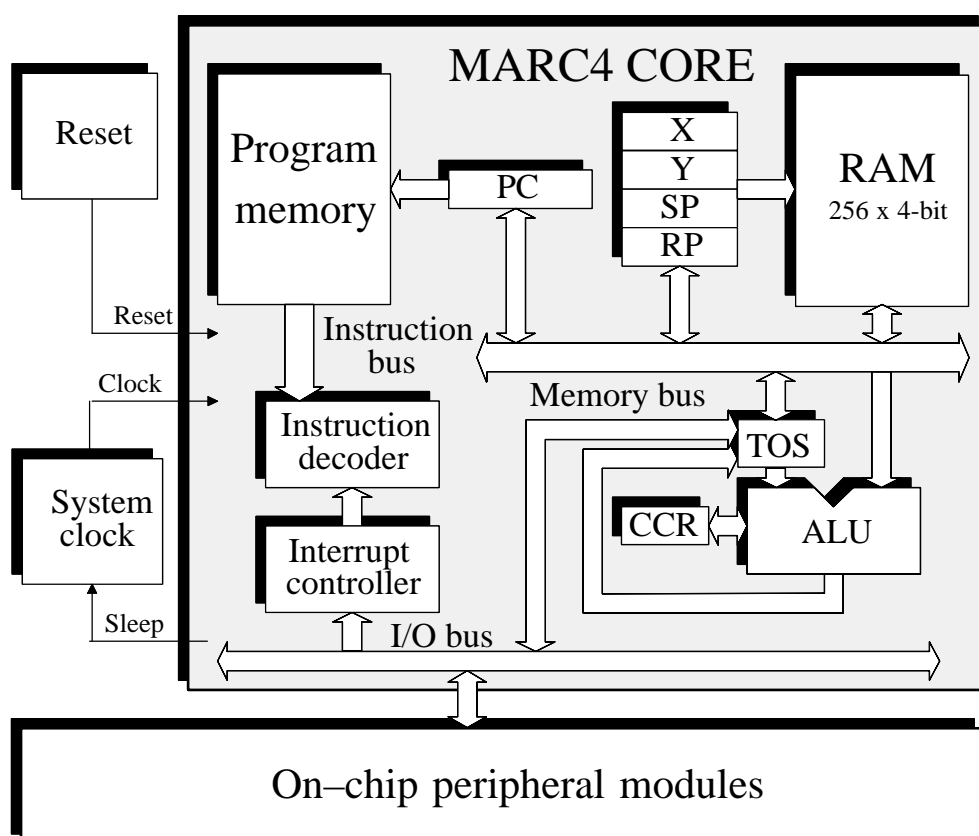
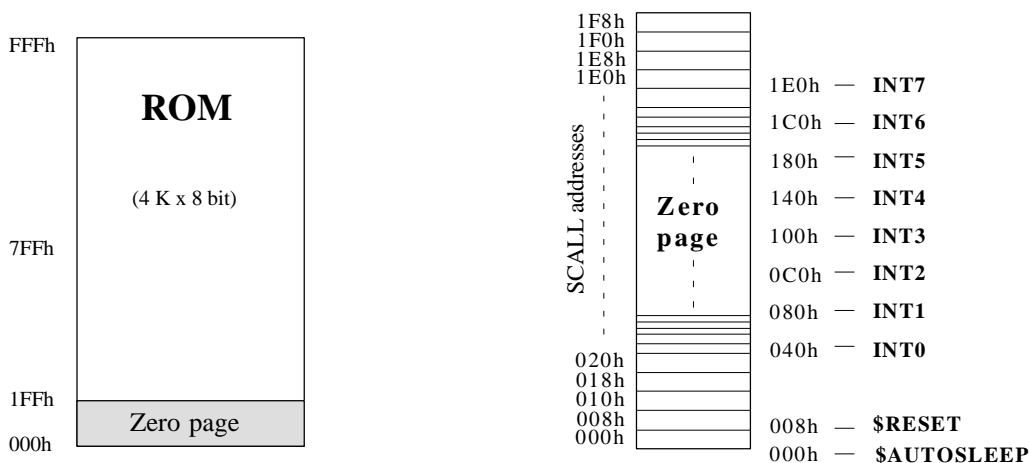


Figure 3. MARC4 core

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2.2 Components of MARC4 Core



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Figure 4. ROM map of M44C092–H

The core contains ROM, RAM, ALU, program counter, RAM address registers, instruction decoder and interrupt controller. The following sections describe each functional block in more detail:

2.2.1 ROM

The program memory (ROM) is mask programmed with the customer application program during the fabrication of the microcontroller. The ROM is addressed by a 12-bit wide program counter, thus predefining a maximum program bank size of 4 Kbytes. An additional 1 Kbyte of ROM exists which is reserved for quality control self-test software. The lowest user ROM address segment is taken up by a 512 byte zero page which contains predefined start addresses for interrupt service routines and special subroutines accessible with single byte instructions (SCALL).

The corresponding memory map is shown in figure 4. Look-up tables of constants can also be held in ROM and are accessed via the MARC4's built-in TABLE instruction.

2.2.2 RAM

The M44C092–H and M44C892–H contain 256 x 4-bit wide static random access memory (RAM). It is used for

the expression stack, the return stack and data memory for variables and arrays. The RAM is addressed by any of the four 8-bit wide RAM address registers SP, RP, X and Y.

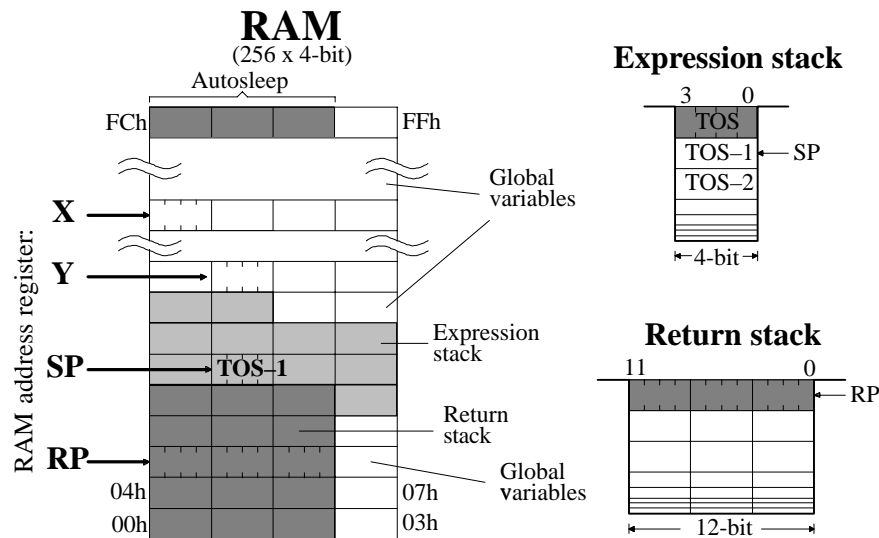
Expression Stack

The 4-bit wide expression stack is addressed with the expression stack pointer (SP). All arithmetic, I/O and memory reference operations take their operands from, and return their results to the expression stack. The MARC4 performs the operations with the top of stack items (TOS and TOS–1). The TOS register contains the top element of the expression stack and works in the same way as an accumulator. This stack is also used for passing parameters between subroutines and as a scratch pad area for temporary storage of data.

Return Stack

The 12-bit wide return stack is addressed by the return stack pointer (RP). It is used for storing return addresses of subroutines, interrupt routines and for keeping loop index counts. The return stack can also be used as a temporary storage area.

The MARC4 instruction set supports the exchange of data between the top elements of the expression stack and the return stack. The two stacks within the RAM have a user definable location and maximum depth.



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Figure 5. RAM map

2.2.3 Registers

The MARC4 controller has seven programmable registers and one condition code register. They are shown in the following programming model.

Program Counter (PC)

The program counter (PC) is a 12-bit register which contains the address of the next instruction to be fetched

from the ROM. Instructions currently being executed are decoded in the instruction decoder to determine the internal micro-operations. For linear code (no calls or branches) the program counter is incremented with every instruction cycle. If a branch-, call-, return-instruction or an interrupt is executed, the program counter is loaded with a new address. The program counter is also used with the TABLE instruction to fetch 8-bit wide ROM constants.

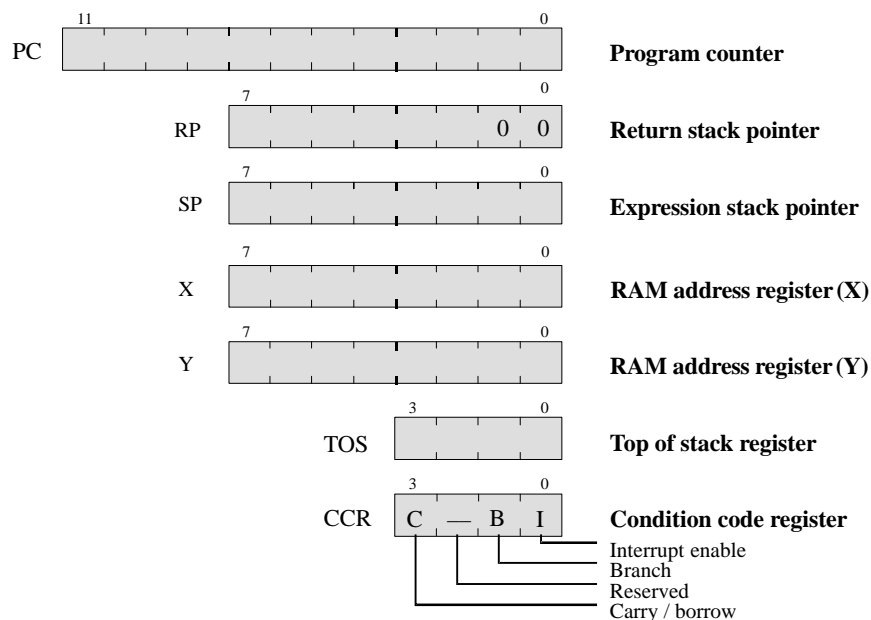


Figure 6. Programming model

RAM Address Registers

The RAM is addressed with the four 8-bit wide RAM address registers: SP, RP, X and Y. These registers allow access to any of the 256 RAM nibbles.

Expression Stack Pointer (SP)

The stack pointer (SP) contains the address of the next-to-top 4-bit item (TOS–1) of the expression stack. The pointer is automatically pre-incremented if a nibble is moved onto the stack or post-decremented if a nibble is removed from the stack. Every post-decrement operation moves the item (TOS–1) to the TOS register before the SP is decremented. After a reset the stack pointer has to be initialized with ">SP S0" to allocate the start address of the expression stack area.

Return Stack Pointer (RP)

The return stack pointer points to the top element of the 12-bit wide return stack. The pointer automatically pre-increments if an element is moved onto the stack, or it post-decrements if an element is removed from the stack. The return stack pointer increments and decrements in steps of 4. This means that every time a 12-bit element is stacked, a 4-bit RAM location is left unwritten. This location is used by the qFORTH compiler to allocate 4-bit variables. After a reset the return stack pointer has to be initialized via ">RP FCh".

RAM Address Registers (X and Y)

The X and Y registers are used to address any 4-bit item in the RAM. A fetch operation moves the addressed nibble onto the TOS. A store operation moves the TOS to the addressed RAM location. By using either the pre-increment or post-decrement addressing mode arrays in the RAM can be compared, filled or moved.

Top Of Stack (TOS)

The top of stack register is the accumulator of the MARC4. All arithmetic/logic, memory reference and I/O operations use this register. The TOS register receives data from the ALU, ROM, RAM or I/O bus.

Condition Code Register (CCR)

The 4-bit wide condition code register contains the branch, the carry and the interrupt enable flag. These bits indicate the current state of the CPU. The CCR flags are set or reset by ALU operations. The instructions SET_BCF, TOG_BF, CCR! and DI allow direct manipulation of the condition code register.

Carry/Borrow (C)

The carry/borrow flag indicates that the borrowing or carrying out of arithmetic logic unit (ALU) occurred during the last arithmetic operation. During shift and rotate operations, this bit is used as a fifth bit. Boolean operations have no effect on the C-flag.

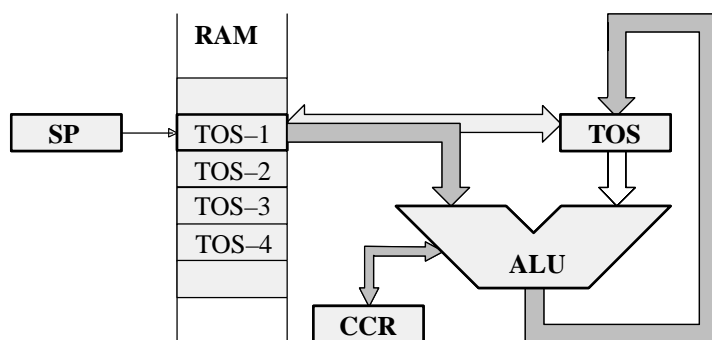
Branch (B)

The branch flag controls the conditional program branching. Should the branch flag have been set by a previous instruction a conditional branch will cause a jump. This flag is affected by arithmetic, logic, shift, and rotate operations.

Interrupt Enable (I)

The interrupt enable flag globally enables or disables the triggering of all interrupt routines with the exception of the non-maskable reset. After a reset or on executing the DI instruction, the interrupt enable flag is reset thus disabling all interrupts. The core will not accept any further interrupt requests until the interrupt enable flag has been set again by either executing an EI, RTI or SLEEP instruction.

2.2.4 ALU



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Figure 7. ALU zero-address operations

The 4-bit ALU performs all the arithmetic, logical, shift and rotate operations with the top two elements of the expression stack (TOS and TOS-1) and returns the result to the TOS. The ALU operations affect the carry/borrow and branch flag in the condition code register (CCR).

2.2.5 I/O Bus

The I/O ports and the registers of the peripheral modules are I/O mapped. All communication between the core and the on-chip peripherals takes place via the I/O bus and the associated I/O control. With the MARC4 IN and OUT instructions the I/O bus allows a direct read or write access to one of the 16 primary I/O addresses. More about the I/O access to the on-chip peripherals is described in the section "Peripheral Modules". The I/O bus is internal and is not accessible by the customer on the final micro-controller device, but it is used as the interface for the MARC4 emulation (see also the section "Emulation").

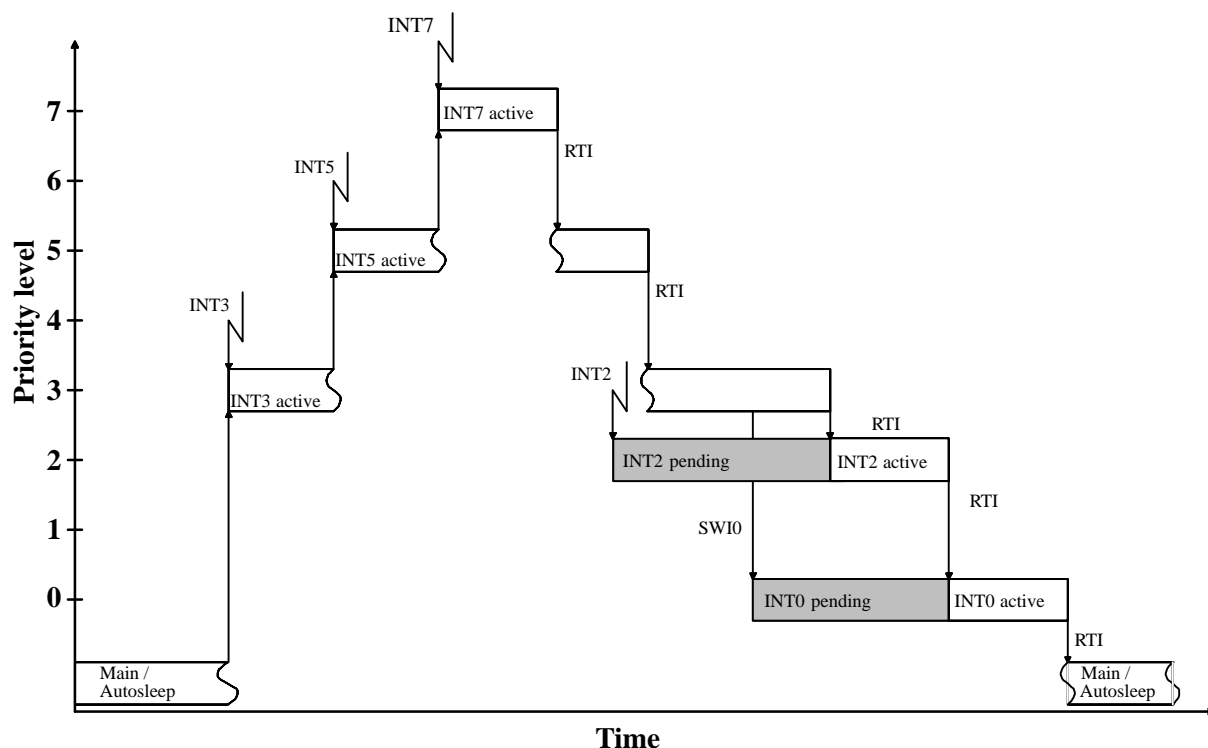
2.2.6 Instruction Set

The MARC4 instruction set is optimized for the high level programming language qFORTH. Many MARC4 instructions are qFORTH words. This enables the compiler to generate a fast and compact program code. The CPU has an instruction pipeline allowing the controller to prefetch an instruction from ROM at the same time as the present instruction is being executed.

The MARC4 is a zero address machine, the instructions containing only the operation to be performed and no source or destination address fields. The operations are implicitly performed on the data placed on the stack. There are one and two byte instructions which are executed within 1 to 4 machine cycles. A MARC4 machine cycle is made up of two system clock cycles (SYSCL). Most of the instructions are only one byte long and are executed in a single machine cycle. For more information refer to the "MARC4 Programmer's Guide".

2.2.7 Interrupt Structure

The MARC4 can handle interrupts with eight different priority levels. They can be generated from the internal and external interrupt sources or by a software interrupt from the CPU itself. Each interrupt level has a hard-wired priority and an associated vector for the service routine in the ROM (see table 2). The programmer can postpone the processing of interrupts by resetting the interrupt enable flag (I) in the CCR. An interrupt occurrence will still be registered, but the interrupt routine only started after the I flag is set. All interrupts can be masked, and the priority individually software configured by programming the appropriate control register of the interrupting module. (see section "Peripheral Modules").



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Figure 8. Interrupt handling

Interrupt Processing

For processing the eight interrupt levels, the MARC4 includes an interrupt controller with two 8-bit wide "interrupt pending" and "interrupt active" registers. The interrupt controller samples all interrupt requests during every non-I/O instruction cycle and latches these in the interrupt pending register. If no higher priority interrupt is present in the interrupt active register, it signals the CPU to interrupt the current program execution. If the interrupt enable bit is set, the processor enters an interrupt acknowledge cycle. During this cycle a short call (SCALL) instruction to the service routine is executed and the current PC is saved on the return stack. An interrupt service routine is completed with the RTI instruction. This instruction sets the interrupt enable flag, resets the corresponding bits in the interrupt pending/active register and fetches the return address from the return stack to the program counter. When the interrupt enable flag is reset (triggering of interrupt routines are disabled), the execution of new interrupt

service routines is inhibited but not the logging of the interrupt requests in the interrupt pending register. The execution of the interrupt is delayed until the interrupt enable flag is set again. Note that interrupts are only lost if an interrupt request occurs while the corresponding bit in the pending register is still set (i.e., the interrupt service routine is not yet finished).

It should also be noted that automatic stacking of the RBR is not carried out by the hardware and so if ROM banking is used, the RBR must be stacked on the expression stack by the application program and restored before the RTI. After a master reset (power-on, brown-out or watchdog reset), the interrupt enable flag and the interrupt pending and interrupt active register are all reset.

Interrupt Latency

The interrupt latency is the time from the occurrence of the interrupt to the interrupt service routine being activated. In MARC4 this is extremely short (taking between 3 to 5 machine cycles depending on the state of the core).

Table 3 Interrupt priority table

| Interrupt | Priority | ROM Address | Interrupt Opcode | Function |
|-----------|----------|-------------|------------------|--|
| INT0 | lowest | 040h | C8h (SCALL 040h) | Software interrupt (SWI0) |
| INT1 | | 080h | D0h (SCALL 080h) | External hardware interrupt, any edge at BP52 or BP53 |
| INT2 | | 0C0h | D8h (SCALL 0C0h) | Timer 1 interrupt |
| INT3 | | 100h | E8h (SCALL 100h) | SSI interrupt or external hardware interrupt at BP40 or BP43 |
| INT4 | | 140h | E8h (SCALL 140h) | Timer 2 interrupt |
| INT5 | | 180h | F0h (SCALL 180h) | Timer 3 interrupt |
| INT6 | ↓ | 1C0h | F8h (SCALL 1C0h) | External hardware interrupt, at any edge at BP50 or BP51 |
| INT7 | highest | 1E0h | FCh (SCALL 1E0h) | Voltage monitor (VM) interrupt |

Table 4 Hardware interrupts

| Interrupt | Interrupt Mask | | Interrupt Source |
|-----------|-----------------------|------------------------------|---|
| | Register | Bit | |
| INT1 | P5CR | P52M1, P52M2 P53M1, P53M2 | Any edge at BP52 any edge at BP53 |
| INT2 | T1M | T1IM | Timer 1 |
| INT3 | SISC | SIM | SSI buffer full / empty or BP40/BP43 interrupt |
| INT4 | T2CM | T2IM | Timer 2 compare match / overflow |
| INT5 | T3CM1 T3CM2 T3C | T3IM1 T3IM2 T3EIM | Timer 3 compare register 1 match Timer 3 compare register 2 match Timer 3 edge event occurs (T3I) |
| INT6 | P5CR | P50M1, P50M2 P51M1, P51M2 | Any edge at BP50, any edge at BP51 |
| INT7 | VCM | VIM | External / internal voltage monitoring |

Software Interrupts

The programmer can generate interrupts by using the software interrupt instruction (SWI) which is supported in qFORTH by predefined macros named SWI0...SWI7. The software triggered interrupt operates exactly like any hardware triggered interrupt. The SWI instruction takes the top two elements from the expression stack and writes the corresponding bits via the I/O bus to the interrupt pending register. Therefore, by using the SWI instruction, interrupts can be re-prioritized or lower priority processes scheduled for later execution.

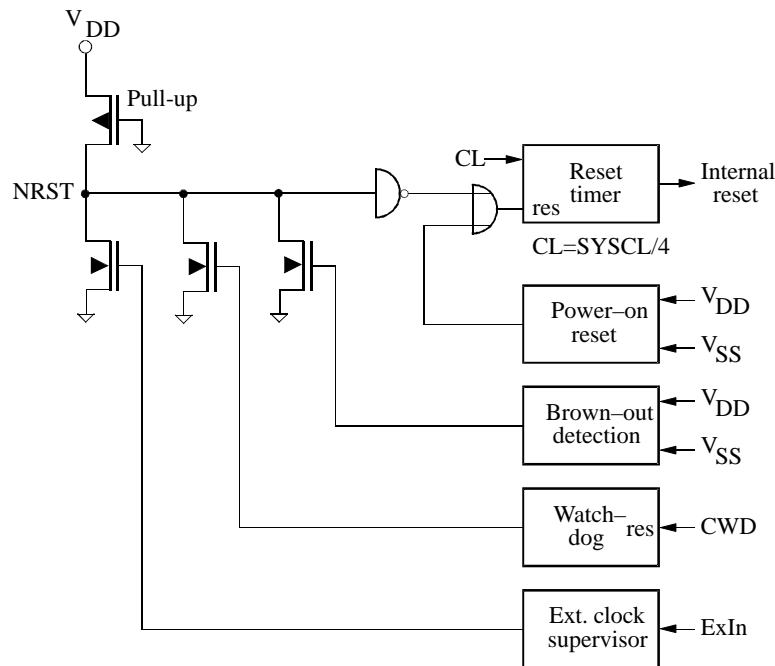
Hardware Interrupts

In the M44C092–H, there are eleven hardware interrupt sources with seven different levels. Each source can be masked individually by mask bits in the corresponding control registers. An overview of the possible hardware configurations is shown in table 4.

2.3 Master Reset

The master reset forces the CPU into a well-defined condition. It is unmaskable and is activated independent of the current program state. It can be triggered by either initial supply power-up, a short collapse of the power supply, brown-out detection circuitry, watchdog time-out, or an external input clock supervisor stage (see figure 9). A master reset activation will reset the interrupt enable flag, the interrupt pending register and the interrupt active register. During the power-on reset phase the I/O bus control signals are set to 'reset mode' thereby initializing all on-chip peripherals. All bidirectional ports are set to input mode. Attention: During any reset phase, the BP20/NTE input is driven towards V_{DD} by a strong pull-up transistor.

Releasing the reset results in a short call instruction (opcode C1h) to the ROM address 008h. This activates the initialization routine \$RESET which in turn has to initialize all necessary RAM variables, stack pointers and peripheral configuration registers (see table 7).



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Figure 9. Reset configuration

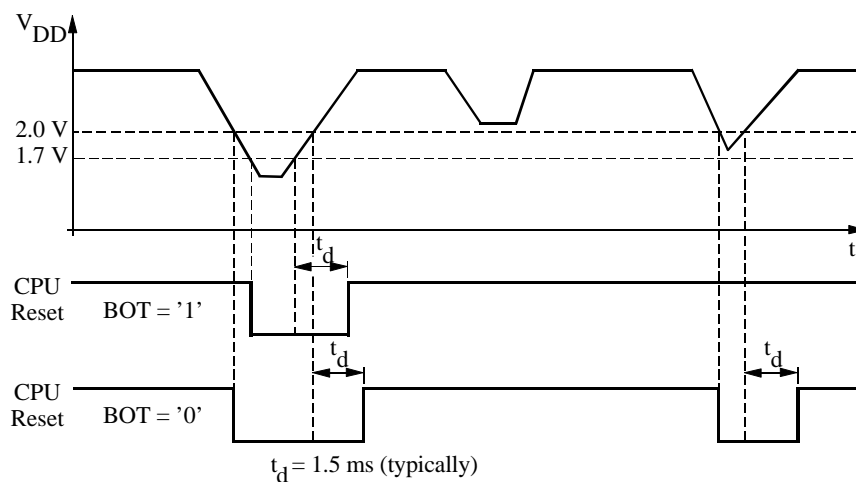
2.3.1 Power-on Reset and Brown-out Detection

The M44C092-H / M44C892-H has a fully integrated power-on reset and brown-out detection circuitry. For reset generation no external components are needed .

These circuits ensure that the core is held in the reset state until the minimum operating supply voltage has been

reached. A reset condition will also be generated should the supply voltage drop momentarily below the minimum operating level except when a power down mode is activated (the core is in SLEEP mode and the peripheral clock is stopped). In this power-down mode the brown-out detection is disabled.

Two values for the brown-out voltage threshold are programmable via the BOT-bit in the SC-register.



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BOT = 1, low brown-out voltage threshold. (1.7 V) is reset value.

BOT = 0, high brown-out voltage threshold (2.0 V).

Figure 10. Brown-out detection

A power-on reset pulse is generated by a V_{DD} rise across the default BOT voltage level (1.7 V). A brown-out reset pulse is generated when V_{DD} falls below the brown-out voltage threshold. Two values for the brown-out voltage threshold are programmable via the BOT-bit in the SC-register. When the controller runs in the upper supply voltage range with a high system clock frequency, the high threshold must be used. When it runs with a lower system clock frequency, the low threshold and a wider supply voltage range may be chosen. For further details, see the electrical specification and the SC-register description for BOT programming.

2.3.2 Watchdog Reset

The watchdog's function can be enabled at the WDC-register and triggers a reset with every watchdog counter overflow. To suppress the watchdog reset, the watchdog counter must be regularly reset by reading the watchdog register address (CWD).

The CPU reacts in exactly the same manner as a reset stimulus from any of the above sources.

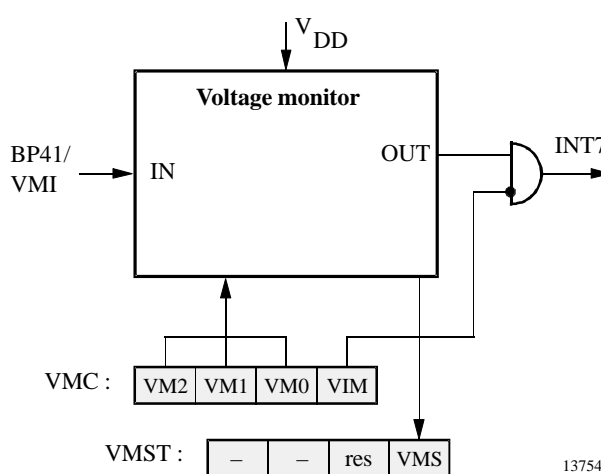
2.3.3 External Clock Supervisor

The external input clock supervisor function can be enabled if the external input clock is selected within the CM- and SC-registers of the clock module.

The CPU reacts in exactly the same manner as a reset stimulus from any of the above sources.

2.4 Voltage Monitor

The voltage monitor consists of a comparator with internal voltage reference. It is used to supervise the supply voltage or an external voltage at the VMI-pin. The comparator for the supply voltage has three internal programmable thresholds one lower threshold (2.2 V), one middle threshold (2.6 V), and one higher threshold (3.0 V). For external voltages at the VMI-pin, the comparator threshold is set to $V_{BG} = 1.3$ V. The VMS-bit indicates if the supervised voltage is below (VMS = 0) or above (VMS = 1) this threshold. An interrupt can be generated when the VMS-bit is set or reset to detect a rising or falling slope. A voltage monitor interrupt (INT7) is enabled when the interrupt mask bit (VIM) is reset in the VMC-register.



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Figure 11. Voltage monitor

2.4.1 Voltage Monitor Control / Status Register

Primary register address: 'F'hex

| | | | | | |
|------------|-------|-------|----------|-------|--------------------|
| VMC: Write | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset value: 1111b |
| | VM2 | VM1 | VM0 | VIM | |
| VMST: Read | — | — | reserved | VMS | Reset value: xx11b |

VM2: Voltage monitor Mode bit 2

VM1: Voltage monitor Mode bit 1

VM0: Voltage monitor Mode bit 0

| VM2 | VM1 | VM0 | Function |
|-----|-----|-----|---|
| 1 | 1 | 1 | Disable voltage monitor |
| 1 | 1 | 0 | External (VIM-input), internal reference threshold (1.3 V), interrupt with negative slope |
| 1 | 0 | 1 | Not allowed |
| 1 | 0 | 0 | External (VMI-input), internal reference threshold (1.3 V), interrupt with positive slope |
| 0 | 1 | 1 | Internal (supply voltage), high threshold (3.0 V), interrupt with negative slope |
| 0 | 1 | 0 | Internal (supply voltage), middle threshold (2.6 V), interrupt with negative slope |
| 0 | 0 | 1 | Internal (supply voltage), low threshold (2.2 V), interrupt with negative slope |
| 0 | 0 | 0 | Not allowed |

VIM Voltage Interrupt Mask bit

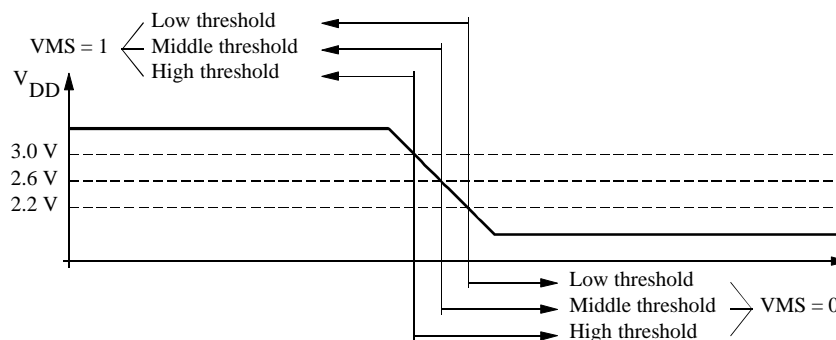
VIM = 0, voltage monitor interrupt is enabled

VIM = 1, voltage monitor interrupt is disabled

VMS Voltage Monitor Status bit

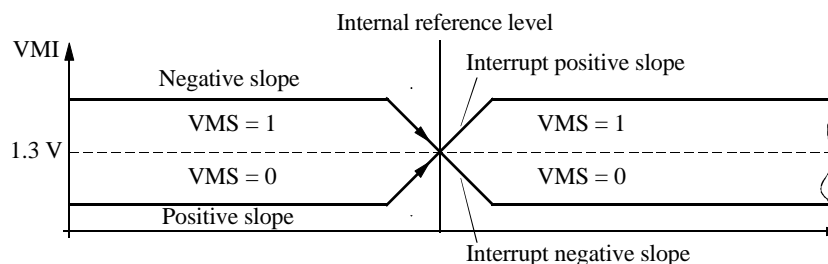
VMS = 0, the voltage at the comparator input is below Vref

VMS = 1, the voltage at the comparator input is above Vref



13755

Figure 12. Internal supply voltage supervisor



13756

Figure 13. External input voltage supervisor

2.5 Clock Generation

2.5.1 Clock Module

The M44C092-H / M44C892-H contains a clock module with 4 different internal oscillator types: two RC-oscillators, one 4-MHz crystal oscillator and one 32-kHz crystal oscillator. The pins OSC1 and OSC2 are the interface to connect a crystal either to the 4-MHz, or to the 32-kHz crystal oscillator. OSC1 can be used as input for external clocks or to connect an external trimming resistor for the RC-oscillator 2. All necessary circuitry except the crystal and the trimming resistor is integrated on-chip. One of these oscillator types or an external input clock can be selected to generate the system clock (SYSCL).

In applications that do not require exact timing, it is possible to use the fully integrated RC-oscillator 1 without any external components. The RC-oscillator 1 center frequency tolerance is better than $\pm 50\%$. The RC-oscillator 2 is a trimmable oscillator whereby the oscillator frequency can be trimmed with an external resistor attached between OSC1 and V_{DD} . In this

configuration, the RC-oscillator 2 frequency can be maintained stable to within a tolerance of $\pm 15\%$ over the full operating temperature and voltage range.

The clock module is programmable via software with the clock management register (CM) and the system configuration register (SC). The required oscillator configuration can be selected with the OS1-bit and the OS0-bit in the SC-register. A programmable 4-bit divider stage allows the adjustment of the system clock speed. A special feature of the clock management is that an external oscillator may be used and switched on and off via a port pin for the power-down mode. Before the external clock is switched off, the internal RC-oscillator 1 must be selected with the CCS-bit and then the SLEEP mode may be activated. In this state an interrupt can wake up the controller with the RC-oscillator, and the external oscillator can be activated and selected by software. A synchronization stage avoids too short clock periods if the clock source or the clock speed is changed. If an external input clock is selected, a supervisor circuit monitors the external input and generates a hardware reset if the external clock source fails or drops below 500 kHz for more than 1 msec.

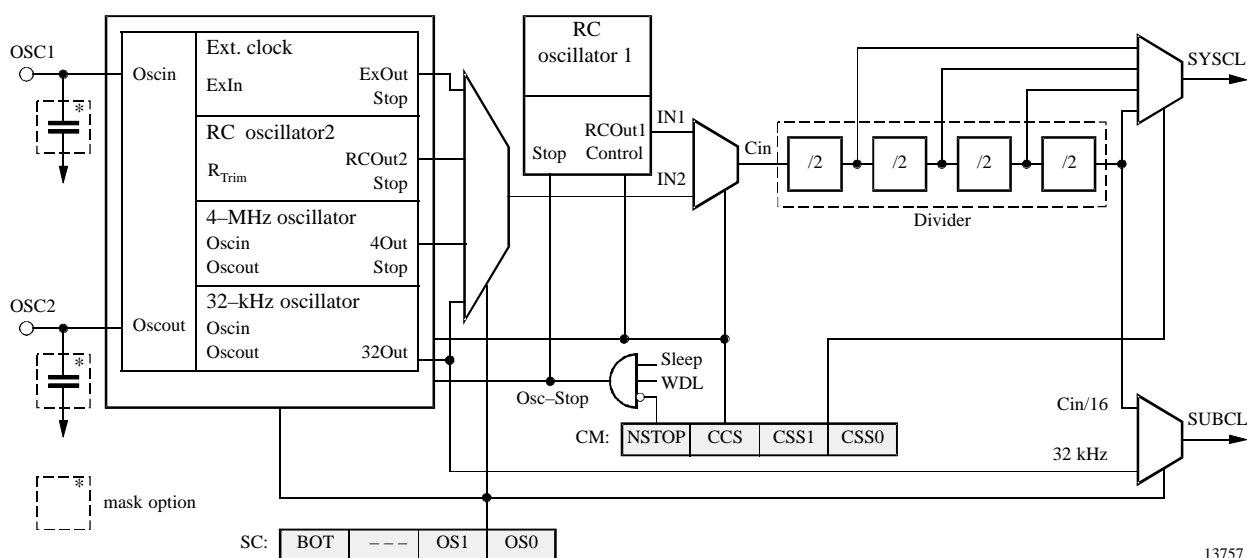


Figure 14. Clock module

Table 5 Clock modes

| Mode | | | Clock Source for SYSCL | | Clock Source for SUBCL |
|------|-----|-----|--------------------------|---|------------------------|
| | OS1 | OS0 | CCS = 1 | CCS = 0 | |
| 1 | 1 | 1 | RC-oscillator 1 (intern) | External input clock | $C_{in} / 16$ |
| 2 | 0 | 1 | RC-oscillator 1 (intern) | RC-oscillator 2 with external trimming resistor | $C_{in} / 16$ |
| 3 | 1 | 0 | RC-oscillator 1 (intern) | 4-MHz oscillator | $C_{in} / 16$ |
| 4 | 0 | 0 | RC-oscillator 1 (intern) | 32-kHz oscillator | 32 kHz |

The clock module generates two output clocks. One is the system clock (SYSCL) and the other the periphery (SUBCL). The SYSCL can supply the core and the peripherals and the SUBCL can supply only the peripherals with clocks. The modes for clock sources are programmable with the OS1-bit and OS0-bit in the SC-register and the CCS-bit in the CM-register.

2.5.2 Oscillator Circuits and External Clock Input Stage

The M44C092-H / M44C892-H series consists of four different internal oscillators: two RC-oscillators, one 4-MHz crystal oscillator, one 32-kHz crystal oscillator and one external clock input stage.

RC-Oscillator 1 Fully Integrated

For timing insensitive applications, it is possible to use the fully integrated RC oscillator 1. It operates without any external components and saves additional costs. The RC-oscillator 1 center frequency tolerance is better than $\pm 50\%$ over the full temperature and voltage range. The basic center frequency of the RC-oscillator 1 is $f_0 \approx 3.8$ MHz. The RC oscillator 1 is selected by default after power-on reset.

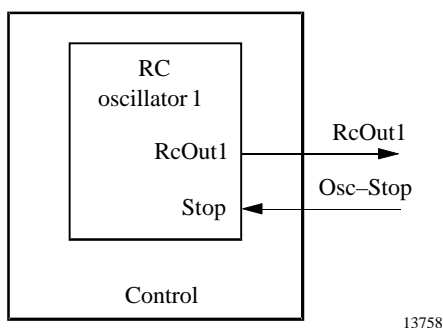


Figure 15. RC-oscillator 1

External Input Clock

The OSC1 can be driven by an external clock source provided it meets the specified duty cycle, rise and fall times and input levels. Additionally the external clock stage contains a supervisory circuit for the input clock. The supervisor function is controlled via the OS1, OS0-bit in the SC-register and the CCS-bit in the CM-register. If the external input clock is missing for more than 1 ms and CCS = 0 is set in the CM-register, the supervisory circuit generates a hardware reset..

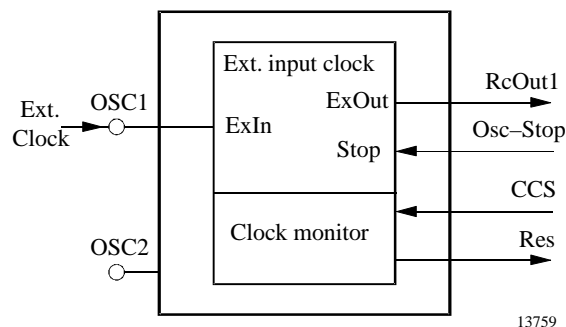


Figure 16. External input clock

| OS1 | OS0 | CCS | Supervisor Reset Output (Res) |
|-----|-----|-----|-------------------------------|
| 1 | 1 | 0 | enable |
| 1 | 1 | 1 | disable |
| x | 0 | x | disable |

RC-Oscillator 2 with External Trimming Resistor

The RC-oscillator 2 is a high resolution trimmable oscillator whereby the oscillator frequency can be trimmed with an external resistor between OSC1 and V_{DD} . In this configuration, the RC-oscillator 2 frequency can be maintained stable to within a tolerance of $\pm 10\%$ over the full operating temperature and a voltage range V_{DD} from 2.5 V to 6.0 V.

For example: An output frequency at the RC-oscillator 2 of 2 MHz, can be obtained by connecting a resistor $R_{ext} = 360$ k Ω (see figures 17).

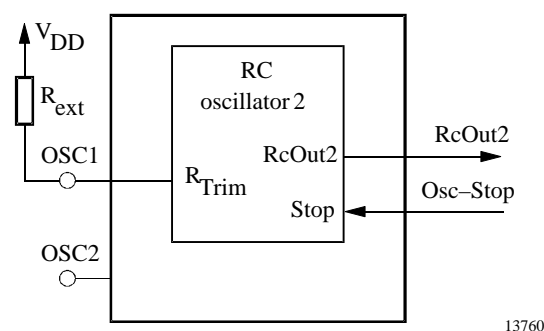


Figure 17. RC-oscillator 2

4-MHz Oscillator

The M44C092-H / M44C892-H 4-MHz oscillator options need a crystal or ceramic resonator connected to the OSC1 and OSC2 pins to establish oscillation. All the necessary oscillator circuitry, with the exception of the actual crystal, resonator, C3 and C4 are integrated on-chip.

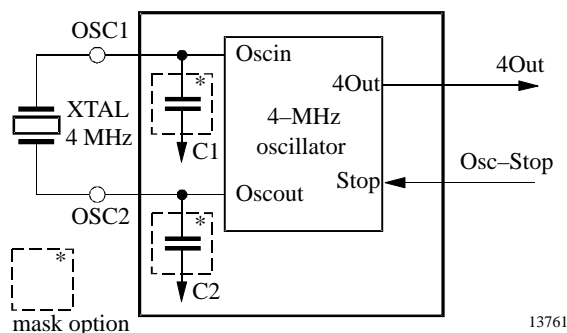


Figure 18. 4-MHz crystal oscillator

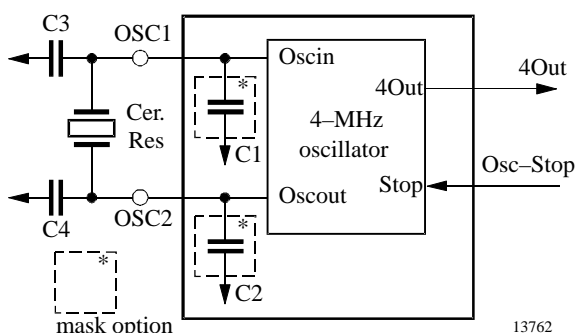


Figure 19. Ceramic resonator

32-kHz Oscillator

Some applications require long-term time keeping or low resolution timing. In this case, an on-chip, low power 32-kHz crystal oscillator can be used to generate both the SUBCL and the SYSCL. In this mode, power consumption is greatly reduced. The 32-kHz crystal oscillator can not be stopped while the power-down mode is in operation.

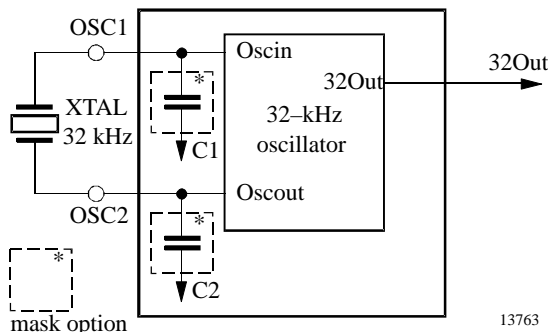


Figure 20. 32-kHz crystal oscillator

2.5.3 Clock Management

The clock management register controls the system clock divider and synchronization stage. Writing to this register triggers the synchronization cycle.

Clock Management Register (CM)

Auxiliary register address: '3'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
|-----|-------|-------|-------|-------|
| CM: | NSTOP | CCS | CSS1 | CSS0 |

Reset value: 1111b

NSTOP Not STOP peripheral clock

NSTOP = 0, stops the peripheral clock while the core is in SLEEP mode

NSTOP = 1, enables the peripheral clock while the core is in SLEEP mode

CCS Core Clock Select

CCS = 1, the internal RC-oscillator 1 generates SYSCL

CCS = 0, the 4-MHz crystal oscillator, the 32-kHz crystal oscillator, an external clock source or the internal RC-oscillator 2 with the external resistor at OSC1 generates SYSCL dependent on the setting of OS0 and OS1 in the system configuration register

CSS1 Core Speed Select 1

CSS0 Core Speed Select 0

| CSS1 | CSS0 | Divider | Note |
|------|------|---------|-------------|
| 0 | 0 | 16 | |
| 1 | 1 | 8 | Reset value |
| 1 | 0 | 4 | |
| 0 | 1 | 2 | |

System Configuration Register (SC)

Primary register address: '3'hex

| | | | | |
|-----------|------------|-------|------------|------------|
| | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| SC: write | BOT | — | OS1 | OS0 |

Reset value: 1x11b

- BOT** **Brown-Out Threshold**
 BOT = 1, low brown-out voltage threshold (1.7 V)
 BOT = 0, high brown-out voltage threshold (2.0 V)
- OS1** **Oscillator Select 1**
- OS0** **Oscillator Select 0**

| Mode | OS1 | OS0 | Input for SUBCL | Selected Oscillators |
|------|-----|-----|----------------------|---|
| 1 | 1 | 1 | C _{in} / 16 | RC-oscillator 1 and external input clock |
| 2 | 0 | 1 | C _{in} / 16 | RC-oscillator 1 and RC-oscillator 2 |
| 3 | 1 | 0 | C _{in} / 16 | RC-oscillator 1 and 4-MHz crystal oscillator |
| 4 | 0 | 0 | 32 kHz | RC-oscillator 1 and 32-kHz crystal oscillator |

If the bit CCS = 0 in the CM-register the RC-oscillator 1 always stops.

2.6 Power-down Modes

The sleep mode is a shut-down condition which is used to reduce the average system power consumption in applications where the µC is not fully utilized. In this mode, the system clock is stopped. The sleep mode is entered via the SLEEP instruction. This instruction sets the interrupt enable bit (I) in the condition code register to enable all interrupts and stops the core. During the sleep mode the peripheral modules remain active and are able to generate interrupts. The µC exits the sleep mode by carrying out any interrupt or a reset.

The sleep mode can only be kept when none of the interrupt pending or active register bits are set. The application of the \$AUTOSLEEP routine ensures the correct function of the sleep mode. For standard applications use the \$AUTOSLEEP routine to enter the power-down mode. Using the SLEEP instruction instead of the \$AUTOSLEEP following an I/O instruction requires to insert 3 non I/O

instruction cycles (for example NOP NOP NOP) between the IN or OUT command and the SLEEP command.

The total power consumption is directly proportional to the active time of the µC. For a rough estimation of the expected average system current consumption, the following formula should be used:

$$I_{\text{total}} (V_{\text{DD}}, f_{\text{syscl}}) = I_{\text{Sleep}} + (I_{\text{DD}} \times t_{\text{active}} / t_{\text{total}})$$

I_{DD} depends on V_{DD} and f_{syscl}.

The M44C092–H / M44C892–H has various power-down modes. During the sleep mode the clock for the MARC4 core is stopped. With the NSTOP-bit in the clock management register (CM) it is programmable if the clock for the on-chip peripherals is active or stopped during the sleep mode. If the clock for the core and the peripherals is stopped the selected oscillator is switched off. An exception is the 32-kHz oscillator, if it is selected it runs continuously independent of the NSTOP-bit. If the oscillator is stopped or the 32 kHz oscillator is selected, power consumption is extremely low.

Table 6 Power-down modes

| Mode | CPU Core | Osc-Stop* | Brown-out Function | RC-Oscillator 1 RC-Oscillator 2 4-MHz Oscillator | 32-kHz Oscillator | External Input Clock |
|------------|----------|-----------|--------------------|--|-------------------|----------------------|
| Active | RUN | NO | Active | RUN | RUN | YES |
| Power-down | SLEEP | NO | Active | RUN | RUN | YES |
| SLEEP | SLEEP | YES | STOP | STOP | RUN | STOP |

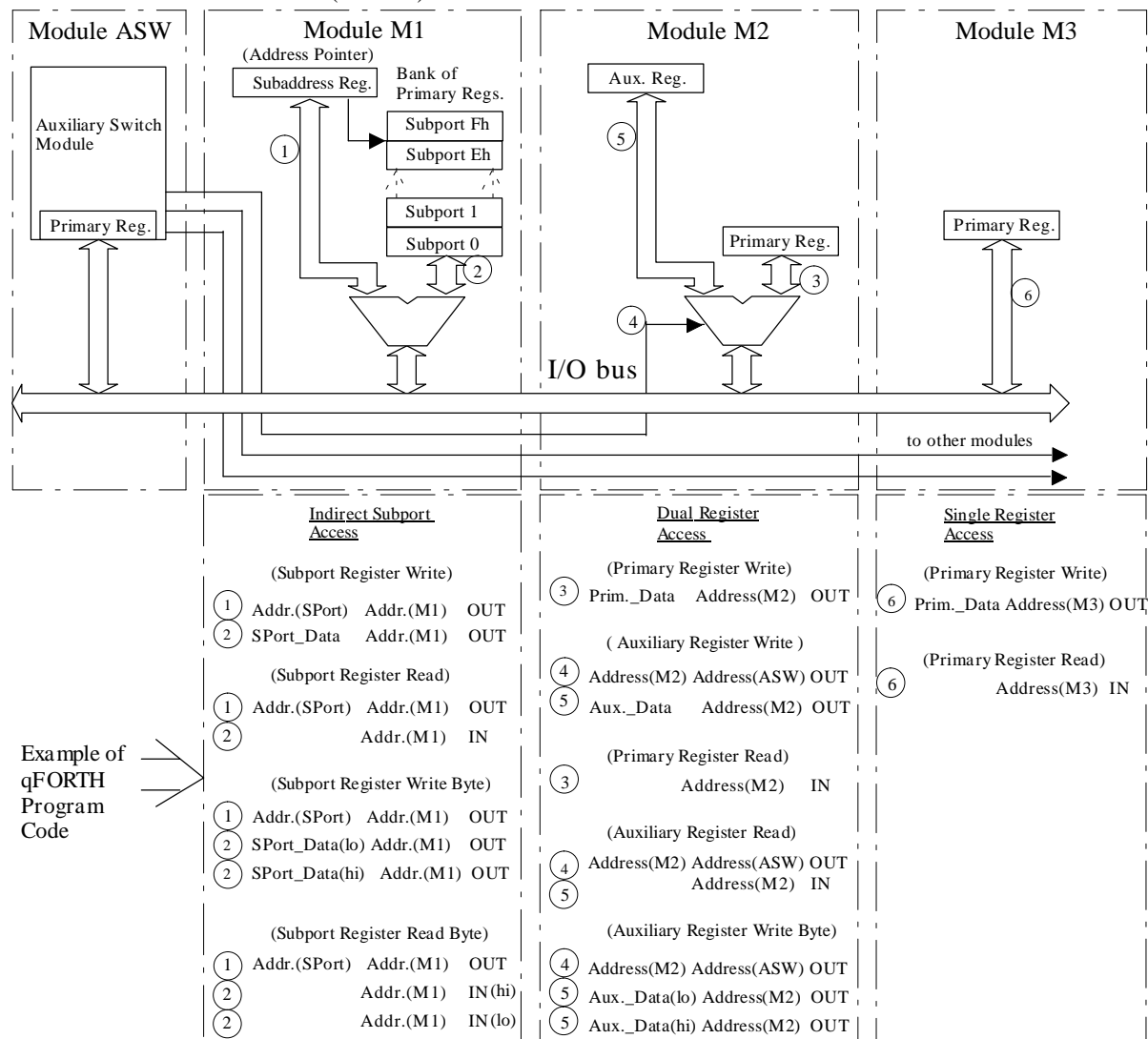
* Osc-Stop = SLEEP & NSTOP & WDL

3 Peripheral Modules

3.1 Addressing Peripherals

Accessing the peripheral modules takes place via the I/O bus (see figure 21). The IN or OUT instructions allow direct addressing of up to 16 I/O modules. A dual register addressing scheme has been adopted to enable direct addressing of the "primary register". To address the "auxiliary register", the access must be switched with an "auxiliary switching module". Thus a single IN (or OUT) to the module address will read (or write) into the module

primary register. Accessing the auxiliary register is performed with the same instruction preceded by writing the module address into the auxiliary switching module. Byte wide registers are accessed by multiple IN- (or OUT-) instructions. For more complex peripheral modules, with a larger number of registers, extended addressing is used. In this case a bank of up to 16 subport registers are indirectly addressed with the subport address. The first OUT-instruction writes the subport address to the sub-address register, the second IN- or OUT-instruction reads data from or writes data to the addressed subport.



Addr.(ASW) = Auxiliary Switch Module Address

Addr.(Mx) = Module Mx Address

Addr.(SPort) = Subport Address

Prim._Data = data to be written into Primary Register.

Aux._Data = data to be written into Auxiliary Register

Aux._Data (lo) = data to be written into Auxiliary Register (low nibble)

Aux._Data (hi) = data to be written into Auxiliary Register (high nibble)

SPort_Data(lo) = data to be written into SubPort (low nibble)

SPort_Data(hi) = data to be written into Subport (high nibble)

(lo) = SPort_Data (low nibble)

(hi) = SPort_Data (high nibble)

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Figure 21. Example of I/O addressing

Table 7 Peripheral addresses

| Port Address | | Name | Write /Read | Reset Value | Register Function | Module Type | See Page |
|-----------------|-------|--------|-------------|-------------------------------------|--|-------------|----------|
| 1 | | P1DAT | W/R | 1xx1b | Port 1 – data register / input data | M3 | 22 |
| 2 | | P2DAT | W/R | 1111b | Port 2 – data register / pin data | M2 | 23 |
| Aux. | P2CR | W | 1111b | Port 2 – control register | 23 | | |
| 3 | | SC | W | 1x11b | Port 3 – system configuration register | M3 | 19 |
| | | CWD | R | xxxxb | Watchdog reset | M3 | 29 |
| Aux. | CM | W/R | 1111b | Port 3 – clock management register | M2 | 18 | |
| 4 | | P4DAT | W/R | 1111b | Port 4 – data register / pin data | M2 | 26 |
| Aux. | P4CR | W | 1111 1111b | Port 4 – control register (byte) | 26 | | |
| 5 | | P5DAT | W/R | 1111b | Port 5 – data register / pin data | M2 | 25 |
| Aux. | P5CR | W | 1111 1111b | Port 5 – control register (byte) | 25 | | |
| 6 | | P6DAT | W/R | 1xx1b | Port 6 – data register / pin data | M2 | 27 |
| Aux | P6CR | W | 1111b | Port 6 – control register (byte) | 27 | | |
| 7 | | T12SUB | W | —— | Data to Timer 1/2 subport | M1 | 20 |
| Subport address | | | | | | | |
| 0 | T2C | W | 0000b | Timer 2 control register | M1 | 37 | |
| 1 | T2M1 | W | 1111b | Timer 2 mode register 1 | M1 | 38 | |
| 2 | T2M2 | W | 1111b | Timer 2 mode register 2 | M1 | 39 | |
| 3 | T2CM | W | 0000b | Timer 2 compare mode register | M1 | 40 | |
| 4 | T2CO1 | W | 1111b | Timer 2 compare register 1 | M1 | 40 | |
| 5 | T2CO2 | W | 1111 1111b | Timer 2 compare register 2 (byte) | M1 | 40 | |
| 6 | —— | —— | —— | Reserved | | | |
| 7 | —— | —— | —— | Reserved | | | |
| 8 | T1C1 | W | 1111b | Timer 1 control register 1 | M1 | 30 | |
| 9 | T1C2 | W | x111b | Timer 1 control register 2 | M1 | 30 | |
| A | WDC | W | 1111b | Watchdog control register | M1 | 31 | |
| B-F | | | | Reserved | | | |
| 8 | | ASW | W | 1111b | Auxiliary / switch register | ASW | 20 |
| 9 | | STB | W | xxxx xxxxb | Serial transmit buffer (byte) | M2 | 60 |
| | | SRB | R | xxxx xxxxb | Serial receive buffer (byte) | | 61 |
| Aux. | SIC1 | W | 1111b | Serial interface control register 1 | | 59 | |
| A | | SISC | W/R | 1x11b | Serial interface status / control register | M2 | 60 |
| Aux. | SIC2 | W | 1111b | Serial interface control register 2 | 59 | | |
| B | | T3SUB | W/R | —— | Data to / from Timer 3 subport | M1 | 20 |
| Subport address | | | | | | | |
| 0 | T3M | W | 1111b | Timer 3 mode register | M1 | 48 | |
| 1 | T3CS | W | 1111b | Timer 3 clock select register | M1 | 49 | |
| 2 | T3CM1 | W | 0000b | Timer 3 compare mode register 1 | M1 | 50 | |
| 3 | T3CM2 | W | 0000b | Timer 3 compare mode register 2 | M1 | 50 | |
| 4 | T3CO1 | W | 1111 1111b | Timer 3 compare register 1 (byte) | M1 | 51 | |
| 4 | T3CP | R | xxxx xxxxb | Timer 3 capture register (byte) | M1 | 51 | |
| 5 | T3CO2 | W | 1111 1111b | Timer 3 compare register 2 (byte) | M1 | 51 | |
| 6-F | | | —— | Reserved | | | |
| C | | T3C | W | 0000b | Timer 3 control register | M3 | 48 |
| | | T3ST | R | x000b | Timer 3 status register | M3 | 49 |
| D | | —— | | —— | Reserved | | |
| E | | —— | | —— | Reserved | | |
| F | | VMC | W | 1111b | Voltage monitor control register | M3 | 15 |
| | | VMST | R | xx11b | Voltage monitor status register | M3 | 15 |

3.2 Bidirectional Ports

With the exception of Port 1 and Port 6, all other ports (2, 4 and 5) are 4 bits wide. Port 1 and Port 6 have a data width of 2 bits (bit 0 and bit 3). All ports may be used for data input or output. All ports are equipped with Schmitt trigger inputs and a variety of mask options for open drain, open source, full complementary outputs, pull up and pull down transistors. All Port Data Registers (PxDAT) are I/O mapped to the primary address register of the respective port address and the Port Control Register (PxCR), to the corresponding auxiliary register.

There are five different directional ports available:

- Port 1 2-bit wide bidirectional ports with automatic full bus width direction switching.
- Port 2 4-bit wide bitwise-programmable I/O port.
- Port 5 4-bit wide bitwise-programmable bidirectional port with optional strong pull-ups and programmable interrupt logic.
- Port 4 4-bit wide bitwise-programmable bidirectional port also provides the I/O interface to Timer 2, SSI, voltage monitor input and external interrupt input.
- Port 6 2-bit wide bitwise-programmable bidirectional port also provides the I/O interface to Timer 3 and external interrupt input.

3.2.1 Bidirectional Port 1

In Port 1 the data direction register is not independently software programmable, the direction of the complete port being switched automatically when an I/O instruction occurs (see figure 22). The port is switched to output mode via an OUT instruction and to input via an IN instruction. The data written to a port will be stored into the output data latches and appears immediately at the port pin following the OUT instruction. After RESET all output latches are set to '1' and the port is switched to input mode. An IN instruction reads the condition of the associated pins.

Note:

Care must be taken when switching the bidirectional port from output to input. The capacitive pin loading at this port in conjunction with the high resistance pull-ups may cause the CPU to read the contents of the output data register rather than the external input state. To avoid this, one should use either of the following programming techniques:

Use two IN-instructions and DROP the first data nibble. The first IN switches the port from output to input and the DROP removes the first invalid nibble. The second IN reads the valid pin state.

Use an OUT-instruction followed by an IN-instruction. Via the OUT-instruction, the capacitive load is charged or discharged depending on the optional pull-up / pull-down configuration. Write a "1" for pins with pull-up resistors and a "0" for pins with pull-down resistors.

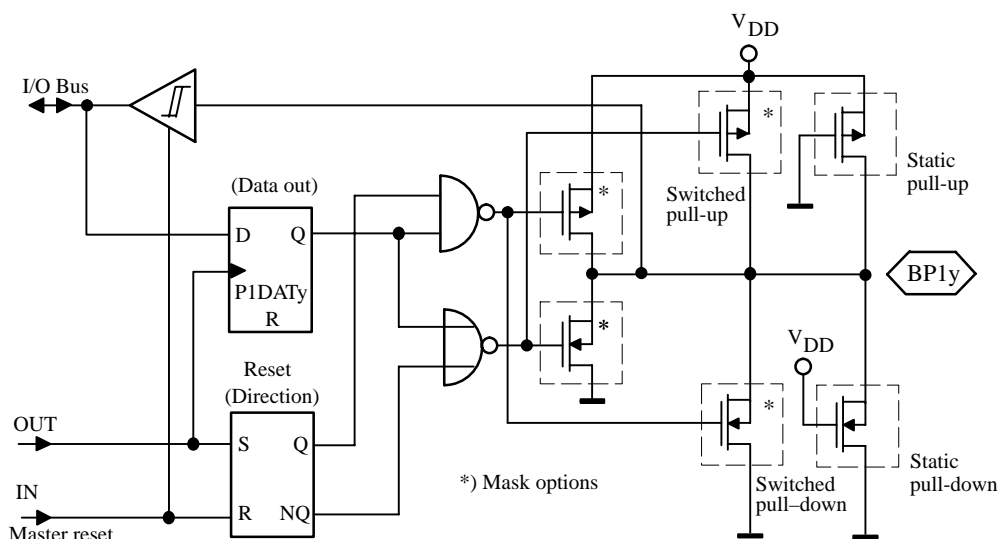


Figure 22. Bidirectional Port 1

3.2.2 Bidirectional Port 2

This, and all other bidirectional ports include a bitwise programmable Control Register (P2CR), which enables the individual programming of each port bit as input or output. It also opens up the possibility of reading the pin

condition when in output mode. This is a useful feature for self testing and for serial bus applications.

Port 2 however, has an increased drive capability and an additional low resistance pull-up/-down transistor mask option.

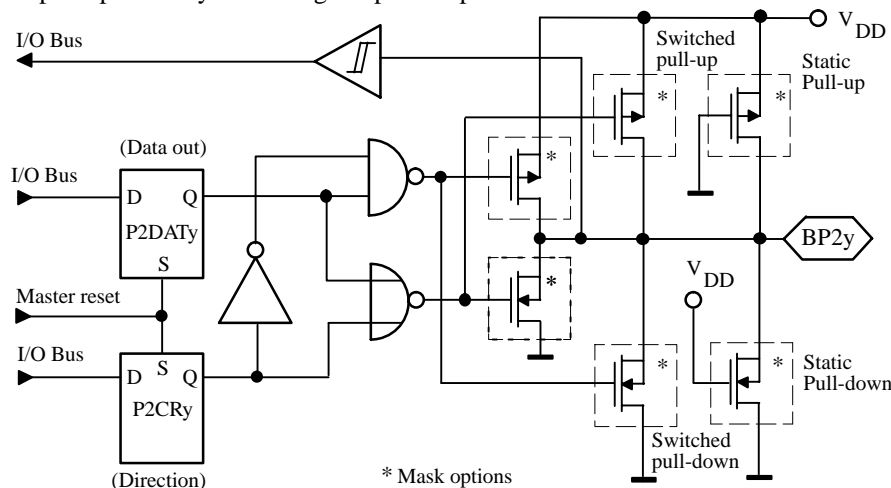


Figure 23. Bidirectional Port 2

Port 2 Data Register (P2DAT)

Primary register address: '2'hex

| | Bit 3 * | Bit 2 | Bit 1 | Bit 0 |
|-------|---------|--------|--------|--------|
| P2DAT | P2DAT3 | P2DAT2 | P2DAT1 | P2DAT0 |

Reset value: 1111b

* Bit 3 → MSB, Bit 0 → LSB

Port 2 Control Register (P2CR)

Auxiliary register address: '2'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
|------|-------|-------|-------|-------|
| P2CR | P2CR3 | P2CR2 | P2CR1 | P2CR0 |

Reset value: 1111b

Value: 1111b means all pins in input mode

| Code 3 2 1 0 | Function |
|-----------------|---------------------|
| x x x 1 | BP20 in input mode |
| x x x 0 | BP20 in output mode |
| x x 1 x | BP21 in input mode |
| x x 0 x | BP21 in output mode |
| x 1 x x | BP22 in input mode |
| x 0 x x | BP22 in output mode |
| 1 x x x | BP23 in input mode |
| 0 x x x | BP23 in output mode |

3.2.3 Bidirectional Port 5

This, and all other bidirectional ports include a bitwise programmable Control Register (P5CR), which allows the individual programming of each port bit as input or output. It also opens up the possibility of reading the pin condition when in output mode. This is a useful feature for self testing and for serial bus applications.

The port pins can also be used as external interrupt inputs (see figures 24 & 25). The interrupts (INT1 and INT6) can be masked or independently configured to trigger on ei-

ther edge. The interrupt configuration and port direction is controlled by the Port 5 Control Register (P5CR). An additional low resistance pull-up/-down transistor mask option provides an internal bus pull-up for serial bus applications.

The Port 5 Data Register (P5DAT) is I/O mapped to the primary address register of address '5'h and the Port 5 Control Register (P5CR) to the corresponding auxiliary register. The P5CR is a byte-wide register and is configured by writing first the low nibble then the high nibble (see section 2.1 "Addressing peripherals").

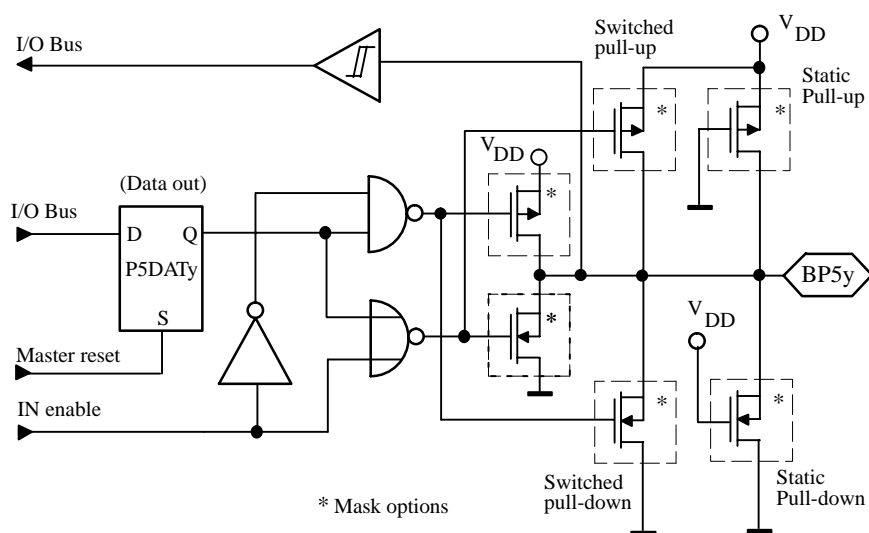


Figure 24. Bidirectional Port 5

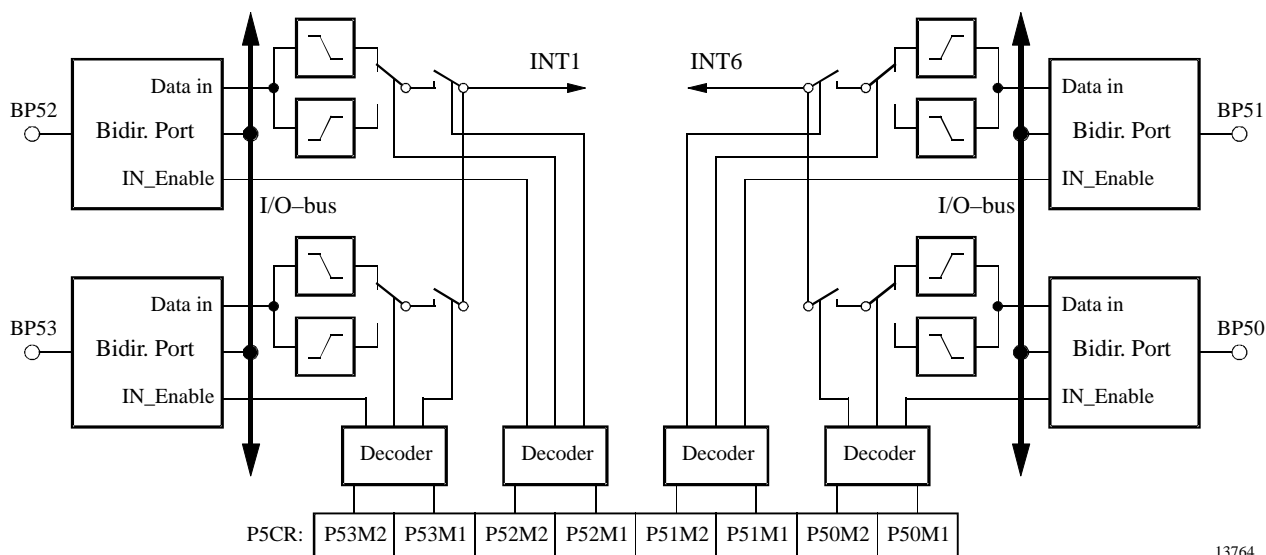


Figure 25. Port 5 external interrupts

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Port 5 Data Register (P5DAT)

Primary register address: '5'hex

| | | | | | |
|-------|--------|--------|--------|--------|--------------------|
| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
| P5DAT | P5DAT3 | P5DAT2 | P5DAT1 | P5DAT0 | Reset value: 1111b |

Port 5 Control Register (P5CR) Byte Write

Auxiliary register address: '5'hex

| | | | | | | |
|------|--------------------|-------|-------|-------|-------|--------------------|
| | | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
| P5CR | First write cycle | P51M2 | P51M1 | P50M2 | P50M1 | Reset value: 1111b |
| | | Bit 7 | Bit 6 | Bit 5 | Bit 4 | |
| | Second write cycle | P53M2 | P53M1 | P52M2 | P52M1 | Reset value: 1111b |

P5xM2, P5xM1 – Port 5x Interrupt mode/direction code

| Auxiliary Address: '5'hex | | First Write Cycle | Second Write Cycle | |
|---------------------------|---|-------------------|--------------------|---|
| Code 3 2 1 0 | Function | | Code 3 2 1 0 | Function |
| x x 1 1 | BP50 in input mode – interrupt disabled | | x x 1 1 | BP52 in input mode – interrupt disabled |
| x x 0 1 | BP50 in input mode – rising edge interrupt | | x x 0 1 | BP52 in input mode – rising edge interrupt |
| x x 1 0 | BP50 in input mode – falling edge interrupt | | x x 1 0 | BP52 in input mode – falling edge interrupt |
| x x 0 0 | BP50 in output mode – interrupt disabled | | x x 0 0 | BP52 in output mode – interrupt disabled |
| 1 1 x x | BP51 in input mode – interrupt disabled | | 1 1 x x | BP53 in input mode – interrupt disabled |
| 0 1 x x | BP51 in input mode – rising edge interrupt | | 0 1 x x | BP53 in input mode – rising edge interrupt |
| 1 0 x x | BP51 in input mode – falling edge interrupt | | 1 0 x x | BP53 in input mode – falling edge interrupt |
| 0 0 x x | BP51 in output mode – interrupt disabled | | 0 0 x x | BP53 in output mode – interrupt disabled |

3.2.4 Bidirectional Port 4

The bidirectional Port 4 is both a bitwise configurable I/O port and provides the external pins for the Timer 2, SSI and the voltage monitor input (VMI). As a normal port, it performs in exactly the same way as bidirectional Port 2 (see figure 26). Two additional multiplexes allow data and port direction control to be passed over to other internal modules (Timer 2, VM or SSI). The I/O-pins for SC

and SD line have an additional mode to generate an SSI-interrupt.

All four Port 4 pins can be individually switched by the P4CR-register. Figure 26 shows the internal interfaces to bidirectional Port 4.

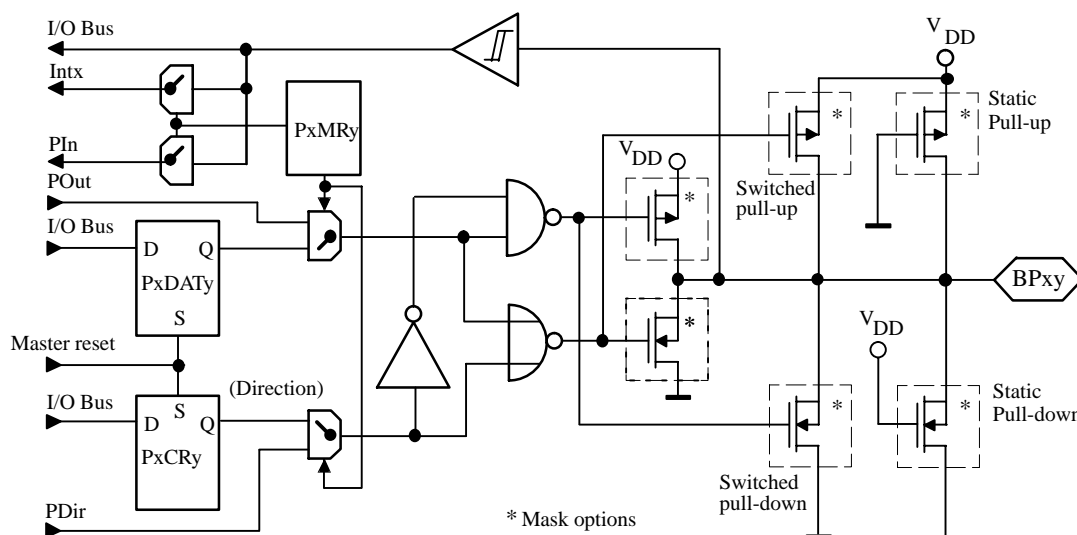


Figure 26. Bidirectional Port 4 and Port 6

Port 4 Data Register (P4DAT)

Primary register address: '4'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
|--------------|---------------|---------------|---------------|---------------|
| P4DAT | P4DAT3 | P4DAT2 | P4DAT1 | P4DAT0 |

Reset value: 1111b

Port 4 Control Register (P4CR) Byte Write

Auxiliary register address: '4'hex

| | | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
|-------------|---------------------------|--------------|--------------|--------------|--------------|
| P4CR | First write cycle | P41M2 | P41M1 | P40M2 | P40M1 |
| | Second write cycle | P43M2 | P43M1 | P42M2 | P42M1 |

Reset value: 1111b

Reset value: 1111b

P4xM2, P4xM1 – Port 4x Interrupt mode/direction code

| Auxiliary Address: '4'hex | | First Write Cycle | Second Write Cycle | |
|---------------------------|--|-------------------|--------------------|--|
| Code 3 2 1 0 | Function | | Code 3 2 1 0 | Function |
| x x 1 1 | BP40 in input mode | | x x 1 1 | BP42 in input mode |
| x x 1 0 | BP40 in output mode | | x x 1 0 | BP42 in output mode |
| x x 0 1 | BP40 enable alternate function (SC for SSI) | | x x 0 x | BP42 enable alternate function (T2O for Timer 2) |
| x x 0 0 | BP40 enable alternate function (falling edge interrupt input for INT3) | | 1 1 x x | BP43 in input mode |
| 1 1 x x | BP41 in input mode | | 1 0 x x | BP43 in output mode |
| 1 0 x x | BP41 in output mode | | 0 1 x x | BP43 enable alternate function (SD for SSI) |
| 0 1 x x | BP41 enable alternate function (VMI for voltage monitor input) | | 0 0 x x | BP43 enable alternate function (falling edge interrupt input for INT3) |
| 0 0 x x | BP41 enable alternate function (T2I external clock input for Timer 2) | | — | — |

3.2.5 Bidirectional Port 6

The bidirectional Port 6 is both a bitwise configurable I/O port and provides the external pins for the Timer 3. As a normal port, it performs in exactly the same way as bidirectional Port 6 (see figure 26). Two additional multiplexes allow data and port direction control to be passed

over to other internal module (Timer 3). The I/O-pin for T3I line has an additional mode to generate a Timer 3–interrupt.

All two Port 6 pins can be individually switched by the P6CR-register . Figure 26 shows the internal interfaces to bidirectional Port 6.

Port 6 Data Register (P6DAT)

Primary register address: '6'hex

| | | | | | |
|--------------|---------------|-------|-------|---------------|---------------------------|
| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
| P6DAT | P6DAT3 | --- | --- | P6DAT0 | Reset value: 1xx1b |

Port 6 Control Register (P6CR)

Auxiliary register address: '6'hex

| | | | | | |
|-------------|--------------|--------------|--------------|--------------|---------------------------|
| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
| P6CR | P63M2 | P63M1 | P60M2 | P60M0 | Reset value: 1111b |

P6xM2, P6xM1 – Port 6x Interrupt mode/direction code

| Auxiliary Address: '6'hex | | Write Cycle | |
|---------------------------|---|-----------------|---|
| Code 3 2 1 0 | Function | Code 3 2 1 0 | Function |
| x x 1 1 | BP60 in input mode | 1 1 x x | BP63 in input mode |
| x x 1 0 | BP60 in output mode | 1 0 x x | BP63 in output mode |
| x x 0 x | BP60 enable alternate port function (T3O for Timer 3) | 0 x x x | BP63 enable alternate port function (T3I for Timer 3) |

3.3 Universal Timer/Counter / Communication Module (UTCM)

The Universal Timer/counter/Communication Module (UTCM) consists of three timers (Timer 1 ,Timer 2, Timer 3) and a Synchronous Serial Interface (SSI).

- Timer 1 is an interval timer that can be used to generate periodical interrupts and as prescaler for Timer 2, Timer 3, the serial interface and the watchdog function.
- Timer 2 is an 8/12-bit timer with an external clock input (T2I) and an output (T2O).
- Timer 3 is an 8-bit timer/counter with its own input (T3I) and output (T3O).
- The SSI operates as two wire serial interface or as shift register for modulation and demodulation. The modulator and demodulator units work together with the timers and shift the data bits into or out of the shift register.

There is a multitude of modes in which the timers and the serial interface can work together.

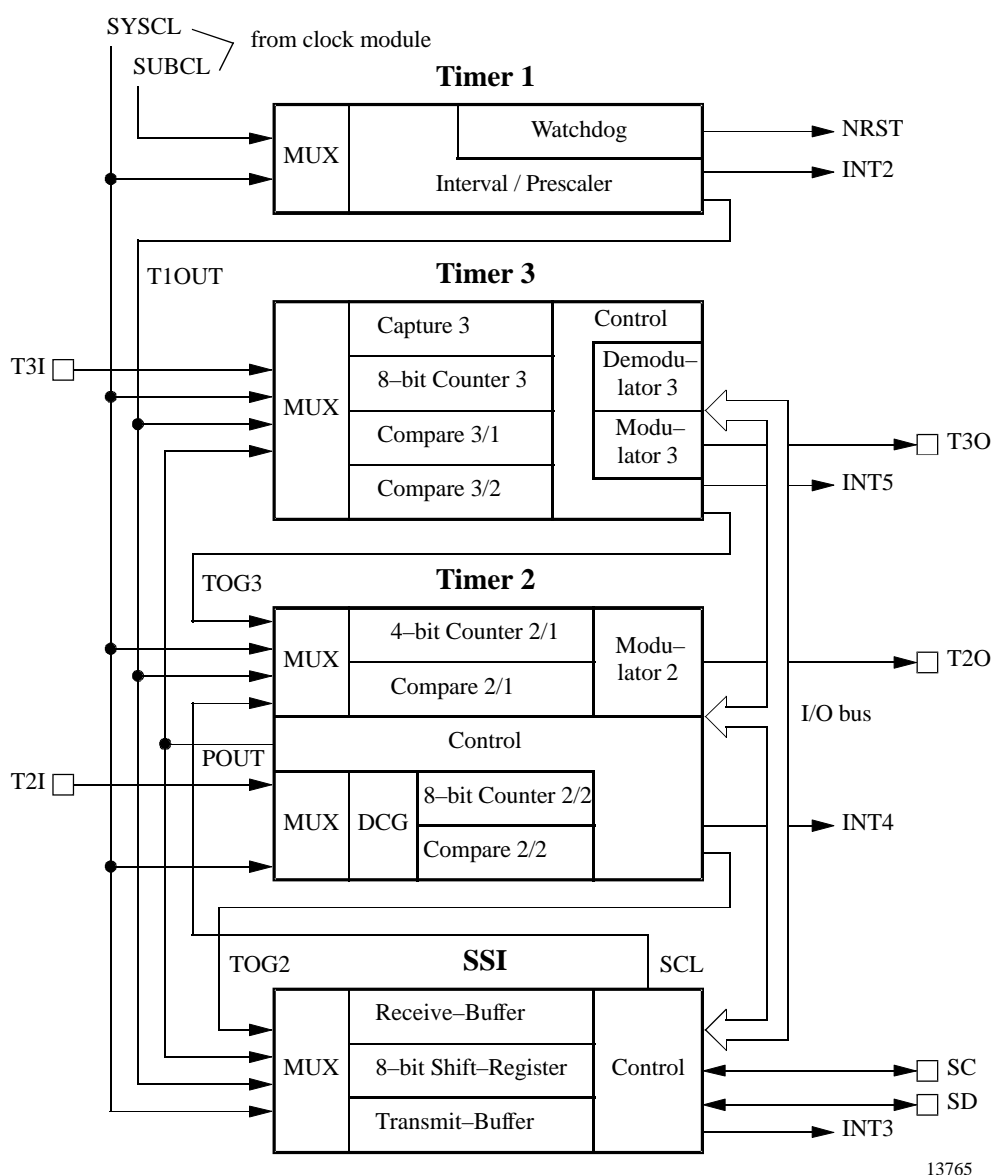


Figure 27. UTCM block diagram

Address: '7'hex – Subaddress: '8'hex

* Bit 3 \rightarrow MSB, Bit 0 \rightarrow LSB

The three bits T1C[2:0] select the divider for timer 1. The resulting time interval depends on this divider and the timer 1 input clock source. The timer input can be supplied by the system clock, the 32kHz oscillator or via the

clock management. If the clock management generates the SUBCL, the selected input clock from the RC oscillator, 4MHz oscillator or an external clock is divided by 16.

| T1C2 | T1C1 | T1C0 | Divider | Time Interval with SUBCL | Time Interval with SUBCL = 32 kHz | Time Interval with SYSCL = 2/1 MHz |
|------|------|------|---------|--------------------------|-----------------------------------|------------------------------------|
| 0 | 0 | 0 | 2 | SUBCL / 2 | 61 μ s | 1 μ s / 2 μ s |
| 0 | 0 | 1 | 4 | SUBCL / 4 | 122 μ s | 2 μ s / 4 μ s |
| 0 | 1 | 0 | 8 | SUBCL / 8 | 244 μ s | 4 μ s / 8 μ s |
| 0 | 1 | 1 | 16 | SUBCL / 16 | 488 μ s | 8 μ s / 16 μ s |
| 1 | 0 | 0 | 32 | SUBCL / 32 | 0.977 ms | 16 μ s / 32 μ s |
| 1 | 0 | 1 | 256 | SUBCL / 256 | 7.812 ms | 128 μ s / 256 μ s |
| 1 | 1 | 0 | 2048 | SUBCL / 2048 | 62.5 ms | 1024 μ s / 2048 μ s |
| 1 | 1 | 1 | 16384 | SUBCL / 16384 | 500 ms | 8192 μ s / 16384 μ s |

Address: '7'hex – Subaddress: '9'hex

* Bit 3 → MSB, Bit 0 → LSB

| | |
|-------------|---|
| T1BP | Timer 1 SUBCL ByPassed T1BP = 1, T1OUT = T1MUX T1BP = 0, T1OUT = SUBCL |
| T1CS | Timer 1 input Clock Select T1CS = 1, CL1 = SUBCL (see figure 28) T1CS = 0, CL1 = SYSCL (see figure 28) |
| T1IM | Timer 1 Interrupt Mask T1IM = 1, disables Timer 1 interrupt T1IM = 0, enables Timer 1 interrupt |

Watchdog Control Register (WDC)

Address: '7'hex – Subaddress: 'A'hex

| | Bit 3 * | Bit 2 | Bit 1 | Bit 0 | |
|-----|---------|-------|-------|-------|--------------------|
| WDC | WDL | WDR | WDT1 | WDT0 | Reset value: 1111b |

* Bit 3 → MSB, Bit 0 → LSB

- WDL** **WatchDog Lock mode**
WDL = 1, the watchdog can be enabled and disabled by using the WDR-bit
WDL = 0, the watchdog is enabled and locked. In this mode the WDR-bit has no effect. After the WDL-bit is cleared, the watchdog is active until a system reset or power-on reset occurs.
- WDR** **WatchDog Run and stop mode**
WDR = 1, the watchdog is stopped / disabled
WDR = 0, the watchdog is active / enabled
- WDT1** **WatchDog Time 1**
- WDT0** **WatchDog Time 0**

Both these bits control the time interval for the watchdog reset

| WDT1 | WDT0 | Divider | Delay Time to Reset with SUBCL = 32 kHz | Delay Time to Reset with SYSCL = 2 / 1 MHz |
|------|------|---------|--|---|
| 0 | 0 | 512 | 15.625 ms | 0.256 ms / 0.512 ms |
| 0 | 1 | 2048 | 62.5 ms | 1.024 ms / 2.048 ms |
| 1 | 0 | 16384 | 0.5 s | 8.2 ms / 16.4 ms |
| 1 | 1 | 131072 | 4 s | 65.5 ms / 131 ms |

3.3.2 Timer 2

Features: 8/12 bit timer for

- Interrupt, square-wave, pulse and duty cycle generation
- Baud-rate generation for the internal shift register
- Manchester and Biphase modulation together with the SSI
- Carrier frequency generation and modulation together with the SSI

Timer 2 can be used as interval timer for interrupt generation, as signal generator or as baud-rate generator and modulator for the serial interface. It consists of a 4-bit and an 8-bit up counter stage which both have compare registers. The 4-bit counter stages of Timer 2 are cascadable as 12-bit timer or as 8-bit timer with 4-bit prescaler. The timer can also be configured as 8-bit timer and separate 4-bit prescaler.

The Timer 2 input can be supplied via the system clock, the external input clock (T2I), the Timer 1 output clock, the Timer 3 output clock or the shift clock of the serial

interface. The external input clock T2I is not synchronized with SYSCL. Therefore it is possible to use Timer 2 with a higher clock speed than SYSCL. Furthermore with that input clock the Timer 2 operates in the power-down mode SLEEP (CPU core → sleep and OSC–Stop → yes) as well as in the POWER-DOWN (CPU core → sleep and OSC–Stop → no). All other clock sources supplied no clock signal in SLEEP. The 4-bit counter stages of Timer 2 have an additional clock output (POUT).

Its output has a modulator stage that allows the generation of pulses as well as the generation and modulation of carrier frequencies. The Timer 2 output can modulate with the shift register data output to generate Biphase- or Manchester-code.

If the serial interface is used to modulate a bitstream, the 4-bit stage of Timer 2 has a special task. The shift register can only handle bitstream lengths divisible by 8. For other lengths, the 4-bit counter stage can be used to stop the modulator after the right bitcount is shifted out.

If the timer is used for carrier frequency modulation, the 4-bit stage works together with an additional 2-bit duty cycle generator like a 6-bit prescaler to generate carrier frequency and duty cycle. The 8-bit counter is used to en-

able and disable the modulator output for a programmable count of pulses.

For programming the time interval, the timer has a 4-bit and an 8-bit compare register. For programming the timer function, it has four mode and control registers. The comparator output of stage 2 is controlled by a special compare mode register (T2CM). This register contains mask bits for the actions (counter reset, output toggle, timer interrupt) which can be triggered by a compare match event or the counter overflow. This architecture enables the timer function for various modes.

Timer 2 compare data values

The Timer 2 has a 4-bit compare register (T2CO1) and an 8-bit compare register (T2CO2). Both these compare registers are cascaded as a 12-bit compare register, or 8-bit compare register and 4-bit compare register.

For 12-bit compare data value: $m = x + 1 \quad 0 \leq x \leq 4095$

For 8-bit compare data value: $n = y + 1 \quad 0 \leq y \leq 255$

For 4-bit compare data value: $l = z + 1 \quad 0 \leq z \leq 15$

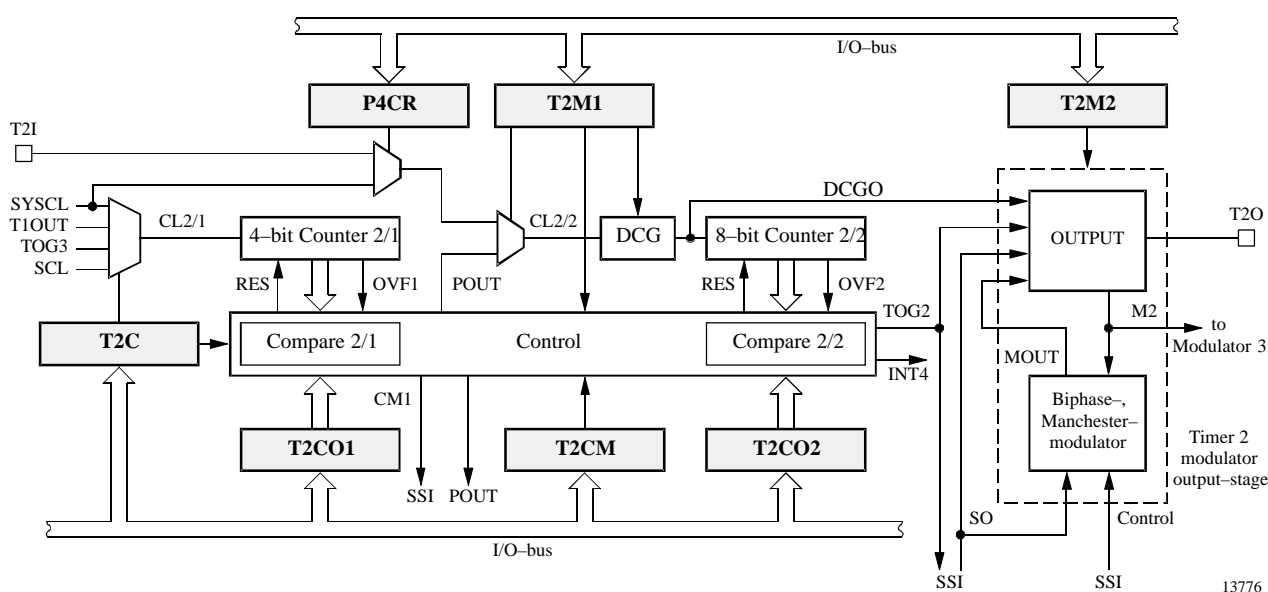


Figure 30. Timer 2

Timer 2 Modes

Mode 1: 12-bit compare counter

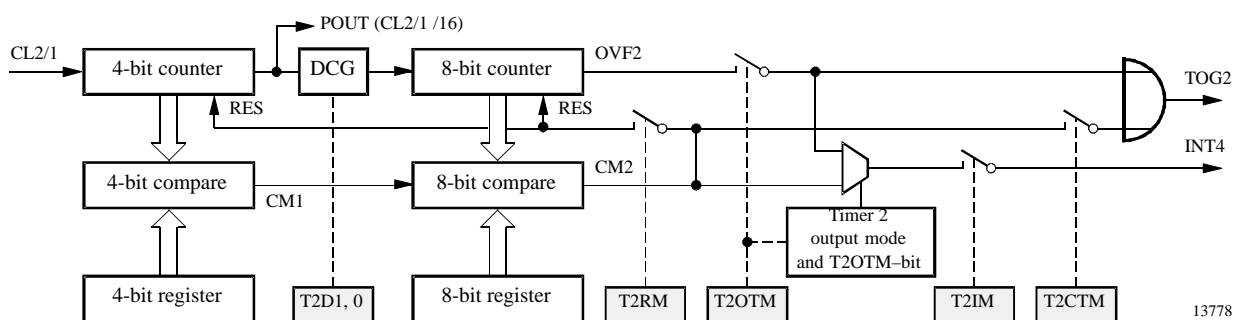


Figure 31. 12-bit compare counter

The 4-bit stage and the 8-bit stage work together as a 12-bit compare counter. A compare match signal of the 4-bit and the 8-bit stage generates the signal for the counter reset, toggle flip-flop or interrupt. The compare action is programmable via the compare mode register (T2CM). The 4-bit counter overflow (OVF1) supplies the clock output (POUT) with clocks. The duty cycle generator (DCG) has to be bypassed in this mode.

Mode 2: 8-bit compare counter with 4-bit programmable prescaler

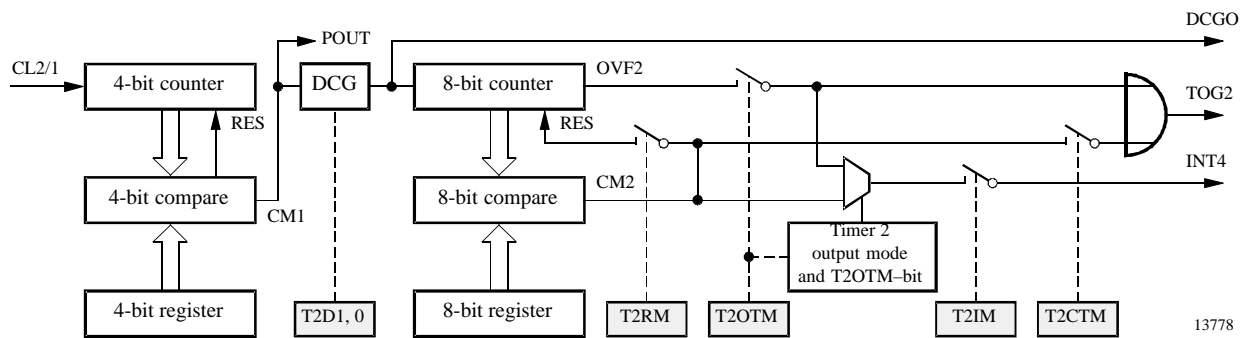


Figure 32. 8-bit compare counter

The 4-bit stage is used as programmable prescaler for the 8-bit counter stage. In this mode, a duty cycle stage is also available. This stage can be used as an additional 2-bit prescaler or for generating duty cycles of 25%, 33% and 50%. The 4-bit compare output (CM1) supplies the clock output (POUT) with clocks.

Mode 3/4: 8-bit compare counter and 4-bit programmable prescaler

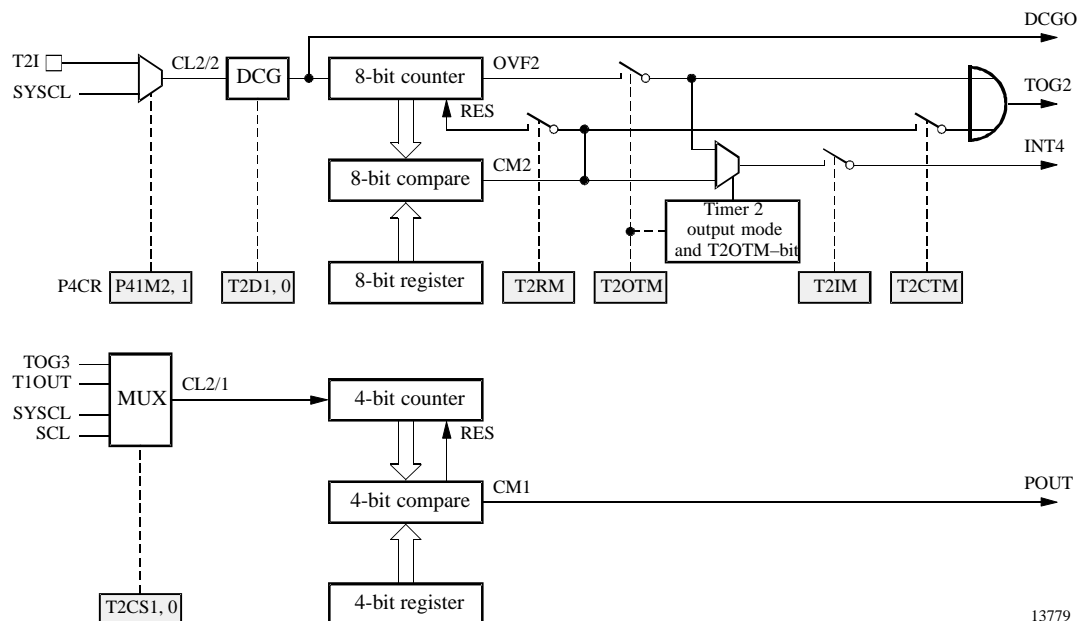


Figure 33. 4-/8-bit compare counter

In these modes the 4-bit and the 8-bit counter stages work independently as a 4-bit prescaler and an 8-bit timer with an 2-bit prescaler or as a duty cycle generator. Only in the mode 3 and mode 4, can the 8-bit counter be supplied via the external clock input (T2I) which is selected via the P4CR register. The 4-bit prescaler is started via activating of mode 3 and stopped and reset in mode 4. Changing mode 3 and 4 has no effect for the 8-bit timer stage. The 4-bit stage can be used as prescaler for Timer 3, the SSI

or to generate the stop signal for modulator 2 and modulator 3.

Timer 2 Output Modes

The signal at the timer output is generated via modulator 2. In the toggle mode, the compare match event toggles the output T2O. For high resolution duty cycle modulation 8 bits or 12 bits can be used to toggle the output. In the duty cycle burst modulator modes the DCG

output is connected to T2O and switched on and off either by the toggle flipflop output or the serial data line of the SSI. Modulator 2 also has 2 modes to output the content of the serial interface as Biphase or Manchester code.

The modulator output stage can be configured by the output control bits in the T2M2 register. The modulator

is started with the start of the shift register (SIR = 0) and stopped either by carrying out a shift register stop (SIR = 1) or compare match event of stage 1 (CM1) of Timer 2. For this task, Timer 2 mode 3 must be used and the prescaler has to be supplied with the internal shift clock (SCL).

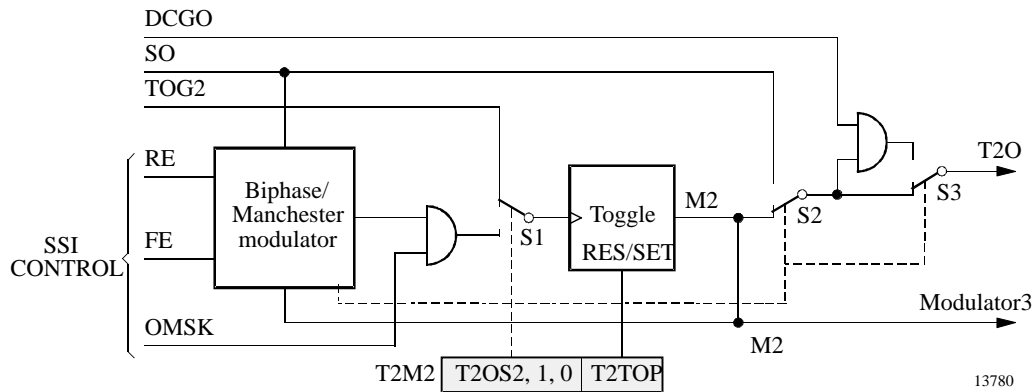


Figure 34. Timer 2 modulator output stage

Timer 2 Output Signals

Timer 2 output mode 1:

Toggle mode A: a Timer 2 compare match toggles the output flip-flop (M2) → T2O

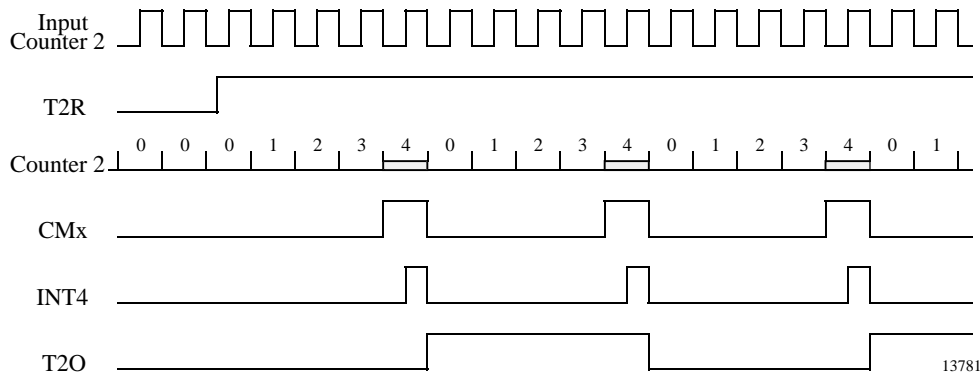


Figure 35. Interrupt timer / square wave generator – the output toggles with each edge compare match event

Timer 2 output mode 1:

Toggle mode B: a Timer 2 compare match toggles the output flip-flop (M2) → T2O

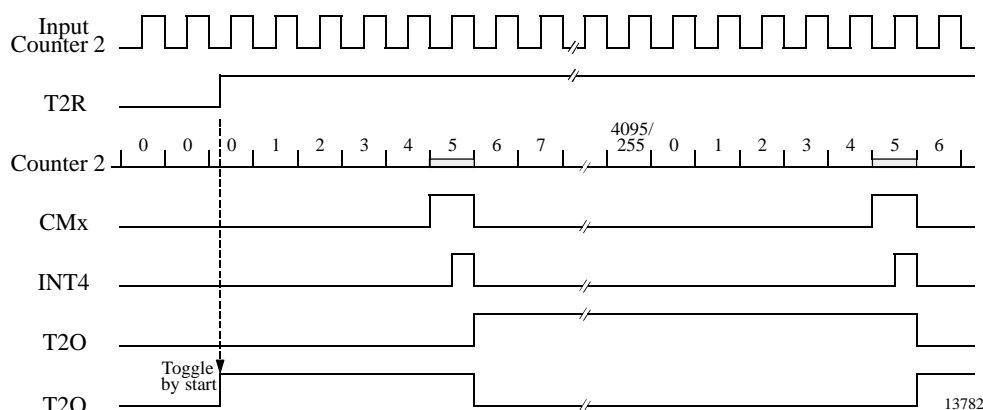


Figure 36. Pulse generator – the timer output toggles with the timer start if the T2TS-bit is set

Timer 2 output mode 1:

Toggle mode C: a Timer 2 compare match toggles the output flip-flop (M2) → T2O

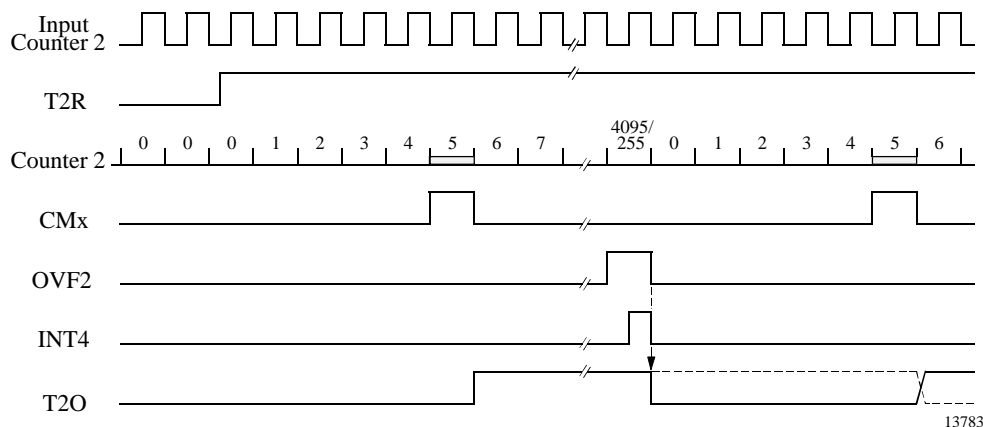


Figure 37. Pulse generator – the timer toggles with timer overflow and compare match

Timer 2 output mode 2:

Duty cycle burst generator 1: the DCG output signal (DCGO) is given to the output, and gated by the output flip-flop (M2)

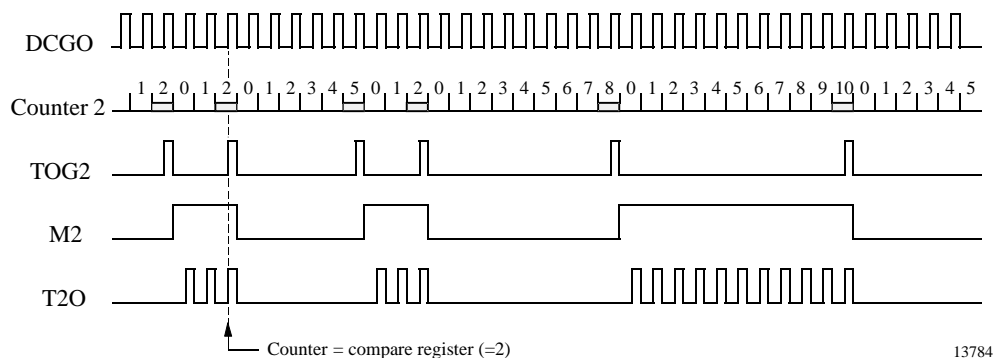


Figure 38. Carrier frequency burst modulation with Timer 2 toggle flip-flop output

Timer 2 output mode 7: PWM mode: Pulse-width modulation output on Timer 2 output pin (T2O)

In this mode the timer overflow defines the period and the compare register defines the duty cycle. During one period only the first compare match occurrence is used to toggle the timer output flip-flop, until the overflow all further compare match are ignored. This avoids the situation that changing the compare register causes the occurrence of several compare match during one period. The resolution at the pulse-width modulation Timer 2 mode 1 is 12-bit and all other Timer 2 modes are 8-bit.

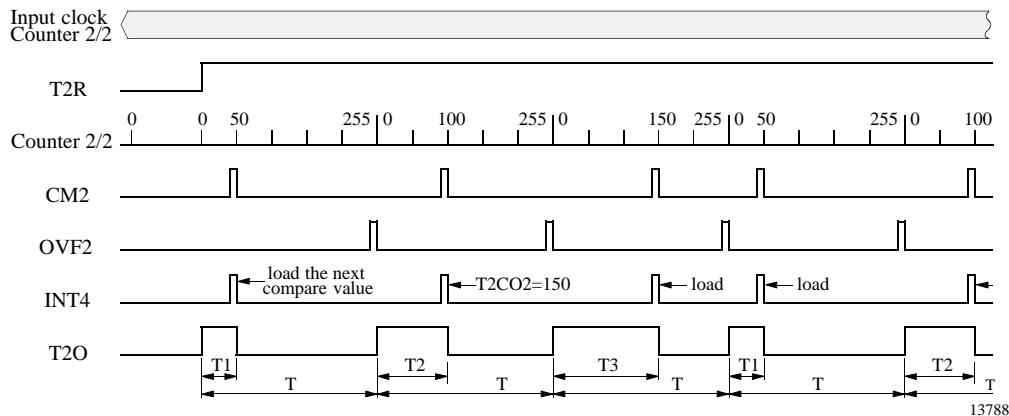


Figure 42. PWM modulation

Timer 2 Registers

Timer 2 has 6 control registers to configure the timer mode, the time interval, the input clock and its output function. All registers are indirectly addressed using extended addressing as described in section "Addressing peripherals". The alternate functions of the Ports BP41 or BP42 must be selected with the Port 4 control register P4CR, if one of the Timer 2 modes require an input at T2I/BP41 or an output at T2O/BP42.

Timer 2 Control Register (T2C)

Address: '7'hex – Subaddress: '0'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|--------------|---|--------------|-------------|------------|---------------------------|
| T2C | T2CS1 | T2CS0 | T2TS | T2R | Reset value: 0000b |
| T2CS1 | Timer 2 Clock Select bit 1 | | | | |
| T2CS0 | Timer 2 Clock Select bit 0 | | | | |
| T2TS | Timer 2 Toggle with Start T2TS = 0, the output flip-flop of Timer 2 is not toggled with the timer start T2TS = 1, the output flip-flop of Timer 2 is toggled when the timer is started with T2R | | | | |
| T2R | Timer 2 Run T2R = 0, Timer 2 stop and reset T2R = 1, Timer 2 run | | | | |

| T2CS1 | T2CS0 | Input Clock (CL 2/1) of Counter Stage 2/1 |
|-------|-------|---|
| 0 | 0 | System clock (SYSCL) |
| 0 | 1 | Output signal of Timer 1 (T1OUT) |
| 1 | 0 | Internal shift clock of SSI (SCL) |
| 1 | 1 | Output signal of Timer 3 (TOG3) |

Timer 2 Mode Register 1 (T2M1)

Address: '7'hex – Subaddress: '1'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|-------------|-------------|-------------|--------------|--------------|---------------------------|
| T2M1 | T2D1 | T2D0 | T2MS1 | T2MS0 | Reset value: 1111b |

T2D1 Timer 2 Duty cycle bit 1

T2D0 Timer 2 Duty cycle bit 0

| T2D1 | T2D0 | Function of Duty Cycle Generator (DCG) | Additional Divider Effect |
|------|------|--|---------------------------|
| 1 | 1 | Bypassed (DCGO0) | / 1 |
| 1 | 0 | Duty cycle 1/1 (DCGO1) | / 2 |
| 0 | 1 | Duty cycle 1/2 (DCGO2) | / 3 |
| 0 | 0 | Duty cycle 1/3 (DCGO3) | / 4 |

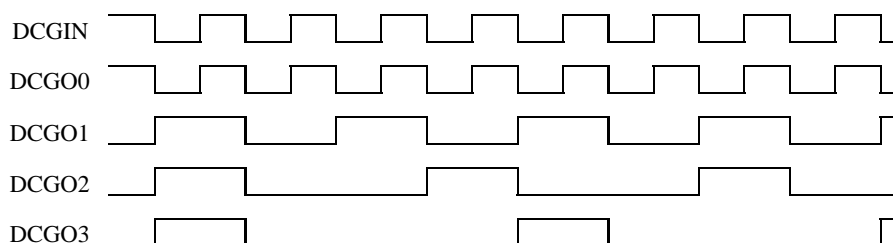
T2MS1 Timer 2 Mode Select bit 1

T2MS0 Timer 2 Mode Select bit 0

| Mode | T2MS1 | T2MS0 | Clock Output (POUT) | Timer 2 Modes |
|------|-------|-------|-------------------------------|--|
| 1 | 1 | 1 | 4-bit counter overflow (OVF1) | 12-bit compare counter; the DCG has to be bypassed in this mode |
| 2 | 1 | 0 | 4-bit compare output (CM1) | 8-bit compare counter with 4-bit programmable prescaler and duty cycle generator |
| 3 | 0 | 1 | 4-bit compare output (CM1) | 8-bit compare counter clocked by SYSCL or the external clock input T2I, 4-bit prescaler run, the counter 2/1 starts after writing mode 3 |
| 4 | 0 | 0 | 4-bit compare output (CM1) | 8-bit compare counter clocked by SYSCL or the external clock input T2I, 4-bit prescaler stop and resets |

Duty Cycle Generator

The duty cycle generator generates duty cycles from 25%, 33% or 50%. The frequency at the duty cycle generator output depends on the duty cycle and the Timer 2 prescaler setting. The DCG-stage can also be used as additional programmable prescaler for Timer 2.



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Figure 43. DCG output signals

Timer 2 Mode Register 2 (T2M2)

Address: '7'hex – Subaddress: '2'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|-------------|--------------|--------------|--------------|--------------|---------------------------|
| T2M2 | T2TOP | T2OS2 | T2OS1 | T2OS0 | Reset value: 1111b |

T2TOP **Timer 2 Toggle Output Preset**
 This bit allows the programmer to preset the Timer 2 output T2O.
 T2TOP = 0, resets the toggle outputs with the write cycle (M2 = 0)
 T2TOP = 1, sets toggle outputs with the write cycle (M2 = 1)
Note: If T2R = 1, no output preset is possible

T2OS2 **Timer 2 Output Select bit 2**

T2OS1 **Timer 2 Output Select bit 1**

T2OS0 **Timer 2 Output Select bit 0**

| Output Mode | T2OS2 | T2OS1 | T2OS0 | Clock Output (POUT) |
|-------------|-------|-------|-------|--|
| 1 | 1 | 1 | 1 | Toggle mode: a Timer 2 compare match toggles the output flip-flop (M2) → T2O |
| 2 | 1 | 1 | 0 | Duty cycle burst generator 1: the DCG output signal (DCG0) is given to the output and gated by the output flip-flop (M2) |
| 3 | 1 | 0 | 1 | Duty cycle burst generator 2: the DCG output signal (DCG0) is given to the output and gated by the SSI internal data output (SO) |
| 4 | 1 | 0 | 0 | Biphase modulator: Timer 2 modulates the SSI internal data output (SO) to Biphase code |
| 5 | 0 | 1 | 1 | Manchester modulator: Timer 2 modulates the SSI internal data output (SO) to Manchester code |
| 6 | 0 | 1 | 0 | SSI output: T2O is used directly as SSI internal data output (SO) |
| 7 | 0 | 0 | 1 | PWM mode: an 8/12-bit PWM mode |
| 8 | 0 | 0 | 0 | Not allowed |

If one of these output modes is used the T2O alternate function of Port 4 must also be activated.

Timer 2 Compare and Compare Mode Registers

Timer 2 has two separate compare registers, T2CO1 for the 4-bit stage and T2CO2 for the 8-bit stage of Timer 2. The timer compares the contents of the compare register current counter value and if it matches it generates an output signal. Dependent on the timer mode, this signal is used to generate a timer interrupt, to toggle the output flip-flop as SSI clock or as a clock for the next counter

stage.

In the 12-bit timer mode, T2CO1 contains bits 0 to 3 and T2CO2 bits 4 to 11 of the 12-bit compare value. In all other modes, the two compare registers work independently as a 4- and 8-bit compare register.

When assigned to the compare register a compare event will be suppressed.

Timer 2 Compare Mode Register (T2CM)

Address: '7'hex – Subaddress: '3'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|--------------|---|--------------|-------------|-------------|---------------------------|
| T2CM | T2OTM | T2CTM | T2RM | T2IM | Reset value: 0000b |
| T2OTM | Timer 2 Overflow Toggle Mask bit T2OTM = 0, disable overflow toggle T2OTM = 1, enable overflow toggle, a counter overflow (OVF2) toggles output flip-flop (TOG2). If the T2OTM-bit is set, only a counter overflow can generate an interrupt except on the Timer 2 output mode 7. | | | | |
| T2CTM | Timer 2 Compare Toggle Mask bit T2CTM = 0, disable compare toggle T2CTM = 1, enable compare toggle, a match of the counter with the compare register toggles output flip-flop (TOG2). In Timer 2 output mode 7 and when the T2CTM-bit is set, only a match of the counter with the compare register can generate an interrupt. | | | | |
| T2RM | Timer 2 Reset Mask bit T2RM = 0, disable counter reset T2RM = 1, enable counter reset, a match of the counter with the compare register resets the counter | | | | |
| T2IM | Timer 2 Interrupt Mask bit T2IM = 0, disable Timer 2 interrupt T2IM = 1, enable Timer 2 interrupt | | | | |

| Timer 2 Output Mode | T2OTM | T2CTM | Timer 2 Interrupt Source |
|---------------------|-------|-------|--------------------------|
| 1, 2, 3, 4, 5 and 6 | 0 | x | Compare match (CM2) |
| 1, 2, 3, 4, 5 and 6 | 1 | x | Overflow (OVF2) |
| 7 | x | 1 | Compare match (CM2) |

Timer 2 COmpare Register 1 (T2CO1)

Address: '7'hex – Subaddress: '4'hex

| | | | | | | |
|--------------|--------------------|--------------|--------------|--------------|--------------|---------------------------|
| T2CO1 | Write cycle | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset value: 1111b |
|--------------|--------------------|--------------|--------------|--------------|--------------|---------------------------|

In prescaler mode the clock is bypassed if the compare register T2CO1 contains 0.

Timer 2 COmpare Register 2 (T2CO2) Byte Write

Address: '7'hex – Subaddress: '5'hex

| | | | | | | |
|--------------|---------------------------|--------------|--------------|--------------|--------------|---------------------------|
| T2CO2 | First write cycle | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset value: 1111b |
| | Second write cycle | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Reset value: 1111b |

3.3.3 Timer 3

Features

- 2 Compare Registers
- Capture Register
- Edge sensitive input with zero cross detection capability
- Trigger and single action modes
- Output control modes
- Automatically modulation and demodulation modes
- FSK modulation
- Pulse width modulation (PWM)
- Manchester demodulation together with SSI
- Biphase demodulation together with SSI
- Pulse-width demodulation together with SSI

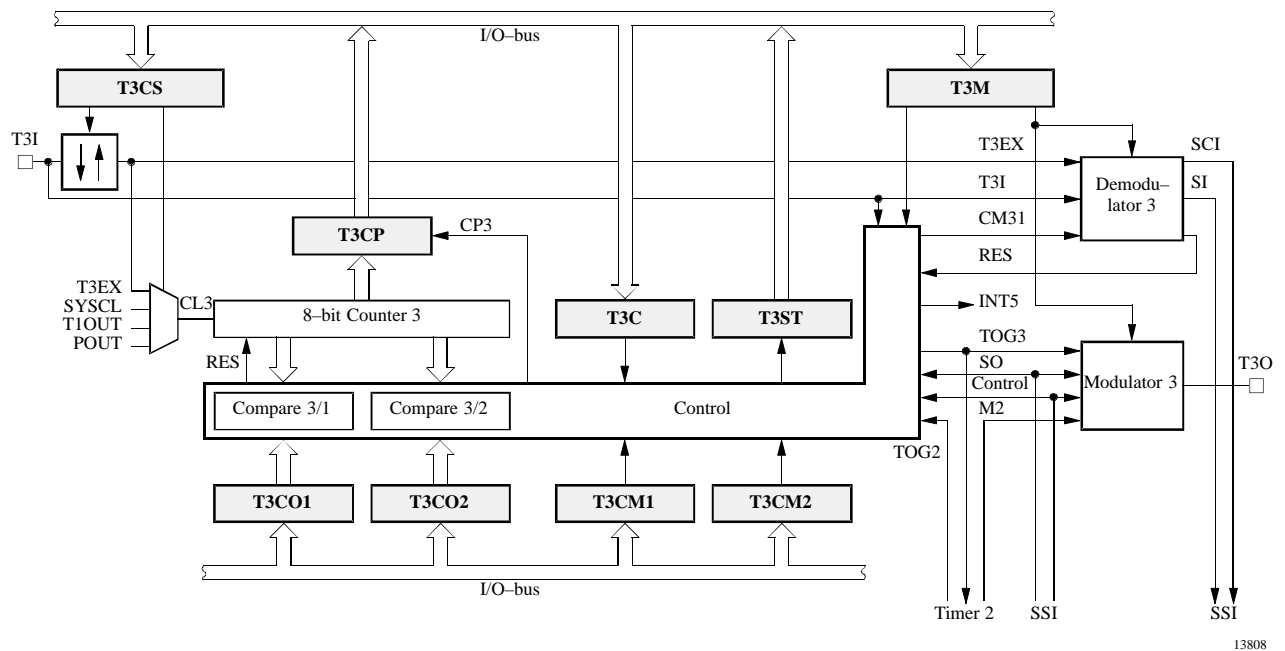
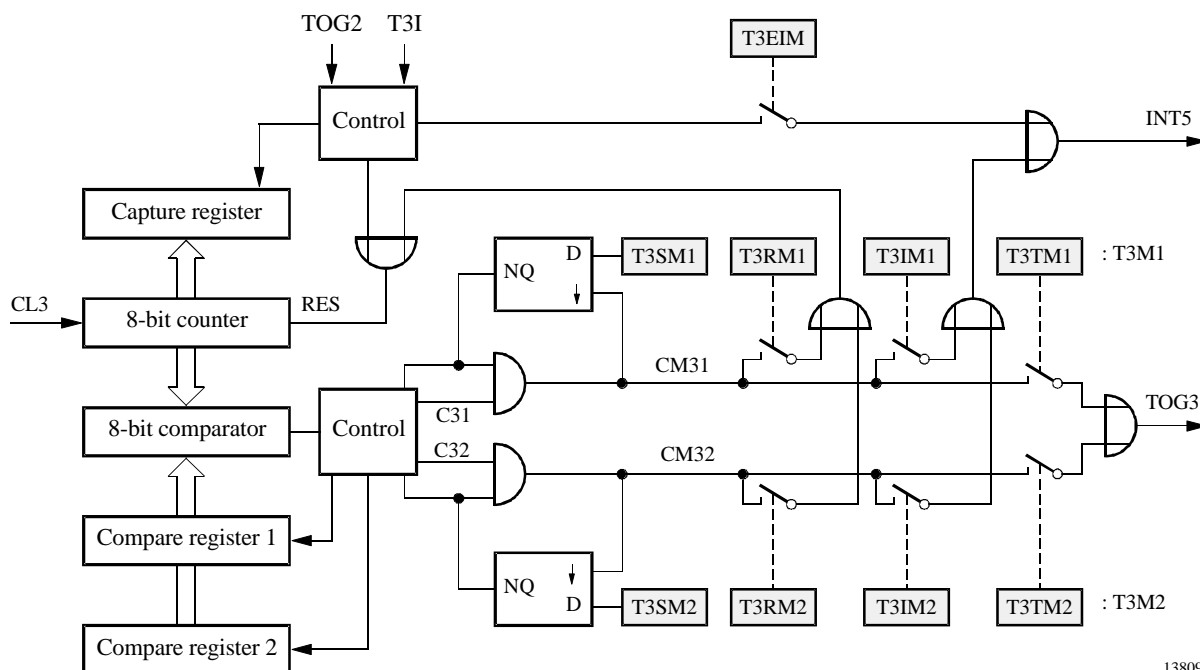


Figure 44. Timer 3

Timer 3 consists of an 8-bit up-counter with two compare registers and one capture register. The timer can be used as event counter, timer and signal generator. Its output can be programmed as modulator and demodulator for the serial interface. The two compare registers enable various modes of signal generation, modulation and demodulation. The counter can be driven by internal and external clock sources. For external clock sources, it has a programmable edge-sensitive input which can be used as counter input, capture signal input or trigger input. This timer input is synchronized with SYSCL. Therefore in the power-down mode SLEEP (CPU core → sleep and OSC-Stop → yes) this timer input is stopped too. The counter is readable via its capture register while it is running. In capture mode, the counter value can be captured by a programmable capture event from the Timer 3 input or Timer 2 output.

A special feature of this timer is the trigger- and single-action mode. In trigger mode, the counter starts counting triggered by the external signal at its input. In single-action mode, the counter counts only one time up to the programmed compare match event. These modes are very useful for modulation, demodulation, signal generation, signal measurement and phase controlling. For phase controlling, the timer input is protected against negative voltages and has zero-cross detection capability.

Timer 3 has a modulator output stage and input functions for demodulation. As modulator it works together with Timer 2 or the serial interface. When the shift register is used for modulation the data shifted out of the register is encoded bitwise. In all demodulation modes, the decoded data bits are shifted automatically into the shift register.



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Figure 45. Counter 3 stage

Timer / Counter Modes

Timer 3 has 6 timer modes and 6 modulator/demodulator modes. The mode is set via the Timer 3 Mode Register T3M.

In all these modes, the compare register and the compare-mode register belonging to it define the counter value for a compare match and the action of a compare match. A match of the current counter value with the content of one compare register triggers a counter reset, a Timer 3 interrupt or the toggling of the output flip-flop. The compare mode registers T3M1 and T3M2 contain the mask bits for enabling or disabling these actions.

The counter can also be enabled to execute single actions with one or both compare registers. If this mode is set the corresponding compare match event is generated only once after the counter start.

Most of the timer modes use its compare registers alternately. **After the start has been activated, the first comparison is carried out via the compare register 1, the second is carried out via the compare register 2, the third is carried out again via the compare register 1 and so on.** This makes it easy to generate signals with constant periods and variable duty cycle or to generate signals with variable pulse and space widths.

If single-action mode is set for one compare register, the comparison is always carried out after the first cycle via the other compare register.

The counter can be started and stopped via the control register T3C. This register also controls the initial level of the output before start. T3C contains the interrupt mask for a T3I input interrupt.

Via the Timer 3 clock-select register, the internal or external clock source can be selected. This register selects also the active edge of the external input. An edge at the external input T3I can generate also an interrupt if the T3EIM-bit is set and the Timer 3 is stopped (T3R = 0) in the T3C-register.

The status of the timer as well as the occurrence of a compare match or an edge detect of the input signal is indicated by the status register T2ST. This allows identification of the interrupt source because all these events share only one timer interrupt.

Timer 3 compare data values

The Timer 3 has two 8-bit compare registers (T3CO1, T3CO2). The compare data value can be 'm' for each of the Timer 3 compare registers.

The compare data value for the compare registers is:

$$m = x + 1 \quad 0 \leq x \leq 255$$

Timer 3 – Mode 1: Timer / Counter

The selected clock from an internal or external source increments the 8-bit counter. In this mode, the timer can be used as event counter for external clocks at T3I or as timer for generating interrupts and pulses at T3O. The counter value can be read by the software via the capture register.

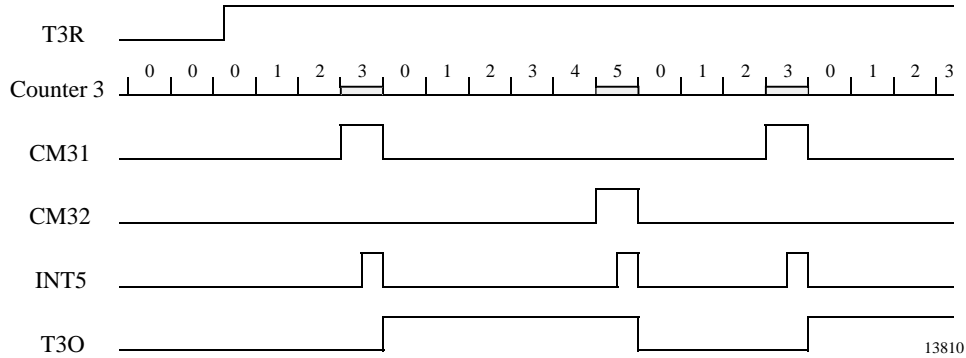


Figure 46. Counter reset with each compare match

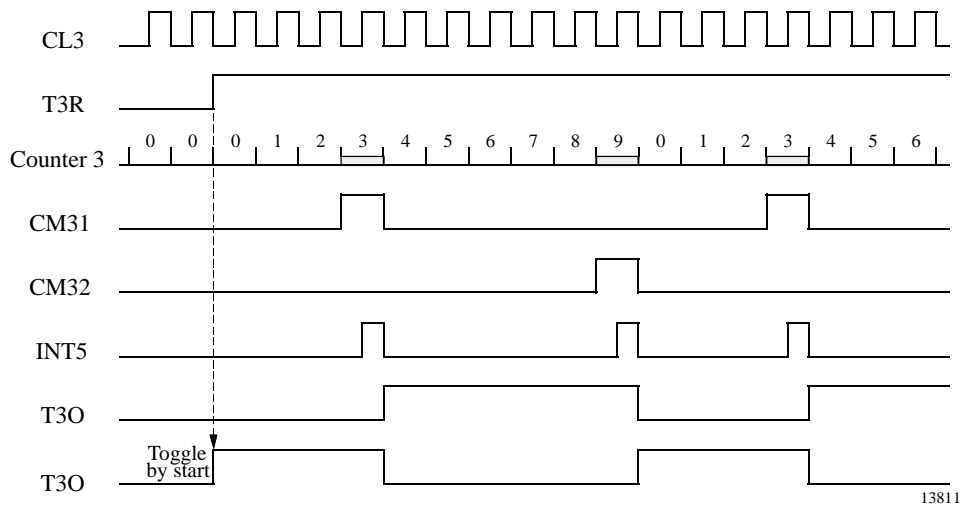


Figure 47. Counter reset with compare register 2 and toggle with start

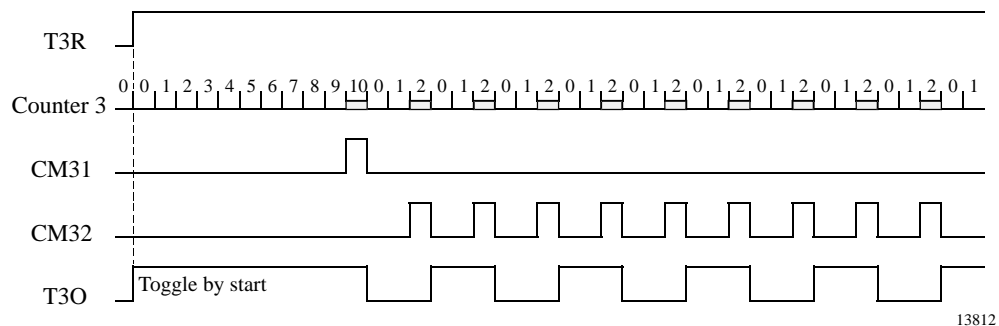


Figure 48. Single action of compare register 1

Timer 3 – Mode 2: Timer/Counter, Ext. Trigger Restart & Ext. Capture (with T3I Input)

The counter is driven by an internal clock source. After starting with T3R, the first edge from the external input T3I starts the counter. The following edges at T3I load the current counter value into the capture register, reset the counter and restart it. The edge can be selected by the programmable edge decoder of the timer input stage. If single-action mode is activated for one or both compare registers the trigger signal restarts the single action.

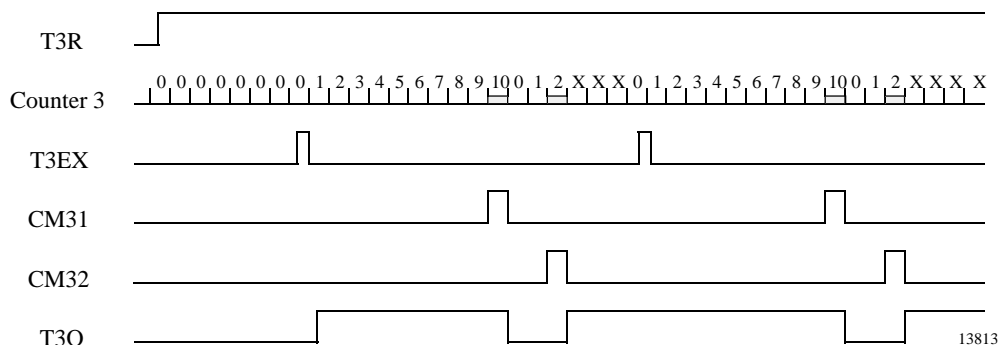


Figure 49. Externally triggered counter reset and start combined with single-action mode

Timer 3 – Mode 3: Timer/Counter, Int. Trigger Restart & Int. Capture (with TOG2)

The counter is driven by an internal or external (T3I) clock source. The output toggle signal of Timer 2 resets the counter. The counter value before the reset is saved in the capture register. If single-action mode is activated for one or both compare registers, the trigger signal restarts the single actions. This mode can be used for frequency measurements or as event counter with time gate (see combination mode 10).

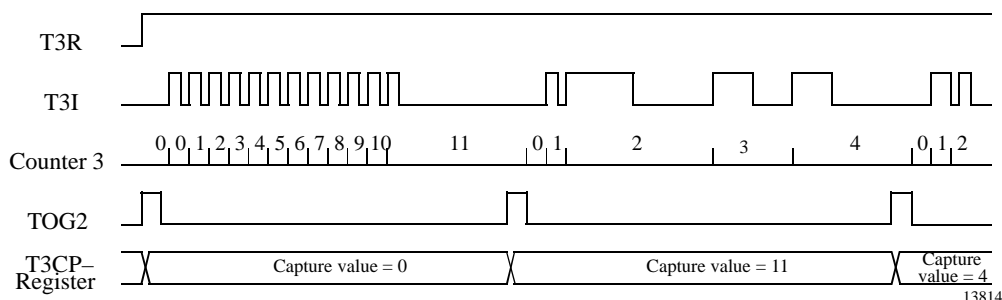


Figure 50. Event counter with time gate

Timer 3 – Mode 4: Timer/Counter

The timer runs as timer/counter in mode 1, but its output T3O is used as output for the Timer 2 output signal.

Timer 3 – Mode 5: Timer/Counter, Ext. Trigger Restart & Ext. Capture (with T3I Input)

The Timer 3 runs as timer/counter in mode 2, but its output T3O is used as output for the Timer 2 output signal.

Timer 3 Modulator / Demodulator Modes

Timer 3 – Mode 6: Carrier Frequency Burst Modulation Controlled by Timer 2 Output Toggle Flip-Flop (M2)

The Timer 3 counter is driven by an internal or external clock source. Its compare– and compare mode registers must be programmed to generate the carrier frequency via the output toggle flip-flop. The output toggle flip-flop of Timer 2 is used to enable or disable the Timer 3 output. Timer 2 can be driven by the toggle output signal of Timer 3 or any other clock source. (see combination mode 11)

Timer 3 – Mode 7: Carrier Frequency Burst Modulation Controlled by SSI Internal Output (SO)

The Timer 3 counter is driven by an internal or external clock source. Its compare– and compare mode registers must be programmed to generate the carrier frequency via the output toggle flip-flop. The output (SO) of the SSI is used to enable or disable the Timer 3 output. The SSI should be supplied with the toggle signal of Timer 2 (see combination mode 12).

Timer 3 – Mode 8: FSK Modulation with Shift Register Data (SO)

The two compare registers are used for generating two different time intervals. The SSI internal data output (SO) selects which compare register is used for the output frequency generation. A '0' level at the SSI data output enables the compare register 1. An '1' level enables compare register 2. The both compare- and compare mode registers must be programmed to generate the two frequencies via the output toggle flip-flop. The SSI can be supplied with the toggle signal of Timer 2. The Timer 3 counter is driven by an internal or external clock source. The Timer 2 counter is driven by the Counter 3 (TOG3) (see also combination mode 13).

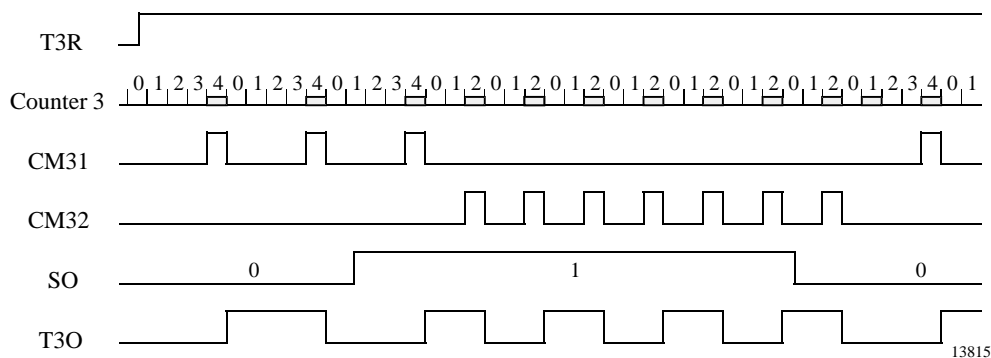


Figure 51. FSK modulation

Timer 3 – Mode 9: Pulse-Width Modulation with the Shift Register

The two compare registers are used for generating two different time intervals. The SSI internal data output (SO) selects which compare register is used for the output pulse generation. In this mode both compare- and compare mode registers must be programmed for generating the two pulse widths. It is also useful to enable the single-action mode for extreme duty cycles. Timer 2 is used as baudrate generator and for the trigger restart of Timer 3. The SSI must be supplied with a toggle signal of Timer 2. The counter is driven by an internal or external clock source (see combination mode 7).

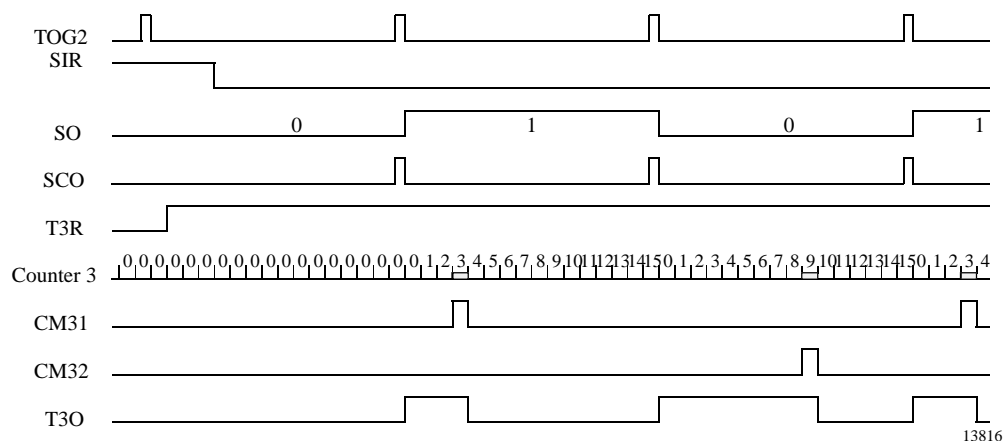


Figure 52. Pulse-width modulation

Timer 3 – Mode 10: Manchester demodulation / pulse-width demodulation

For Manchester demodulation, the edge detection stage must be programmed to detect each edge at the input. These edges are evaluated by the demodulator stage. The timer stage is used to generate the shift clock for the SSI. The compare register 1 match event defines the correct moment for shifting the state from the input T3I as the decoded bit into shift register – after that the demodulator waits for the next edge to synchronize the timer by a reset for the next bit. The compare register 2 can also be used to detect a time-out error and handle it with an interrupt routine (see also combination mode 8).

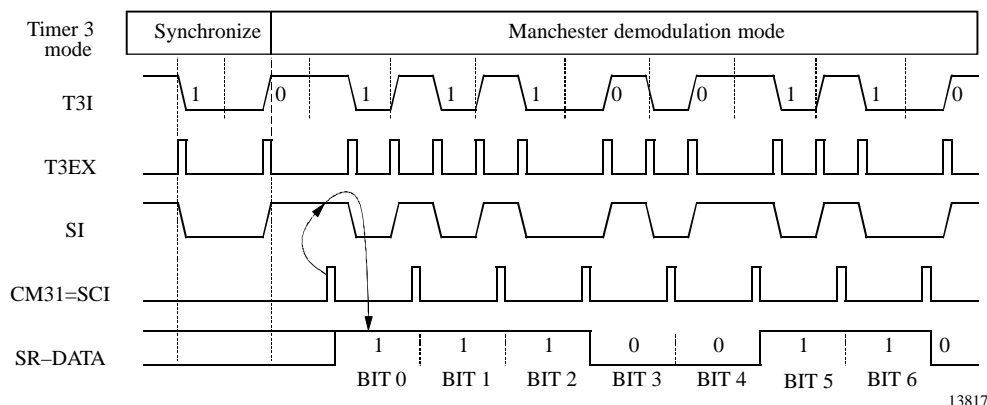


Figure 53. Manchester demodulation

Timer 3 – Mode 11: Biphase demodulation

In the Biphase demodulation mode, the timer operates like in Manchester demodulation mode. The difference is that the bits are decoded via a toggle flip-flop. This flip-flop samples the edge in the middle of the bitframe and the compare register 1 match event shifts the toggle flip-flop output into shift register (see also combined mode 9).

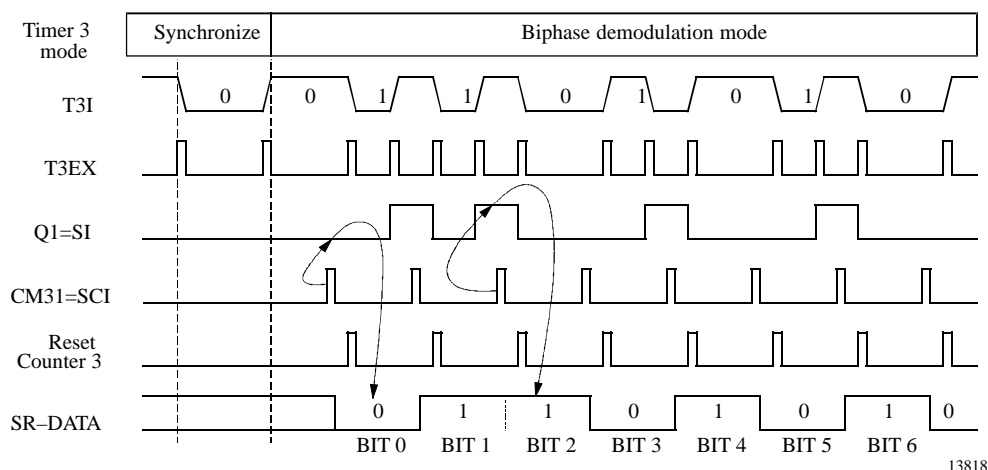


Figure 54. Biphase demodulation

Timer 3 – Mode 12: Timer / counter with external capture mode (T3I)

The counter is driven by an internal clock source and an edge at the external input T3I loads the counter value into the capture register. The edge can be selected with the programmable edge detector of the timer input stage. This mode can be used for signal and pulse measurements.

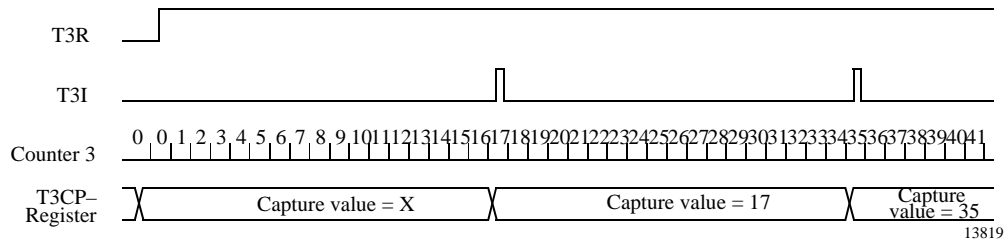
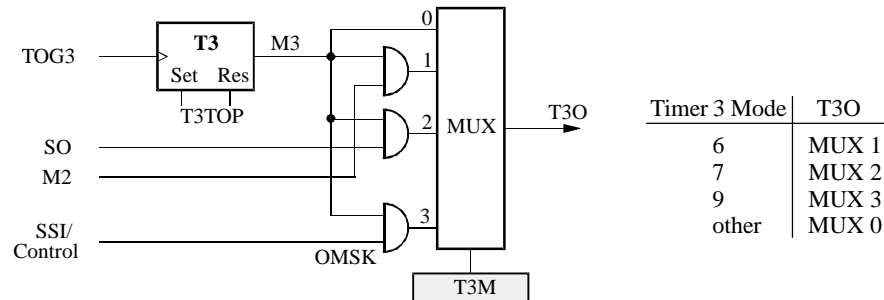


Figure 55. External capture mode

Timer 3 Modulator for Carrier Frequency Burst Modulation

If the output stage operates as pulse-width modulator for the shift register the output can be stopped with stage 1 of Timer 2. For this task, the timer mode 3 must be used and the prescaler must be supplied by the internal shift clock of the shift register.

The modulator can be started with the start of the shift register (SIR=0) and stopped either by a shift register stop (SIR=1) or compare match event of stage 1 of Timer 2. For this task, the Timer 2 must be used in mode 3 and the pre-scaler stage must be supplied by the internal shift clock of the shift register.

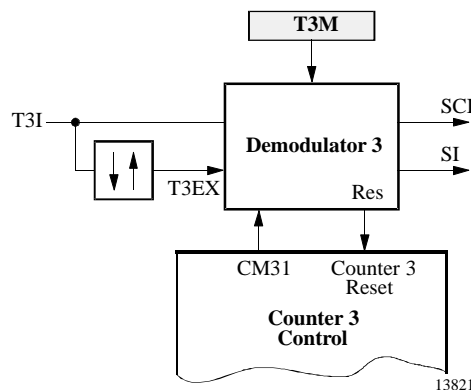


13820

Figure 56. Modulator 3

Timer 3 Demodulator for Biphase, Manchester and Pulse-Width-Modulated Signals

The demodulator stage of Timer 3 can be used to decode Biphase, Manchester and pulse-width-coded signals



13821

Figure 57. Timer 3 demodulator 3

Timer 3 Registers

Timer 3 Mode Register (T3M)

Address: 'B'hex – Subaddress: '0'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|------|---------------------------|-------|-------|-------|--------------------|
| T3M | T3M3 | T3M2 | T3M1 | T3M0 | Reset value: 1111b |
| T3M3 | Timer 3 Mode select bit 3 | | | | |
| T3M2 | Timer 3 Mode select bit 2 | | | | |
| T3M1 | Timer 3 Mode select bit 1 | | | | |
| T3M0 | Timer 3 Mode select bit 0 | | | | |

| Mode | T3M3 | T3M2 | T3M1 | T3M0 | Timer 3 Modes |
|------|------|------|------|------|--|
| 1 | 1 | 1 | 1 | 1 | Timer / counter with a read access |
| 2 | 1 | 1 | 1 | 0 | Timer / counter, external capture & external trigger restart mode (T3I) |
| 3 | 1 | 1 | 0 | 1 | Timer / counter, internal capture & internal trigger restart mode (TOG2) |
| 4 | 1 | 1 | 0 | 0 | Timer / counter mode 1 without output (T2O → T3O) |
| 5 | 1 | 0 | 1 | 1 | Timer / counter mode 2 without output (T2O → T3O) |
| 6 | 1 | 0 | 1 | 0 | Burst modulation with Timer 2 (M2) |
| 7 | 1 | 0 | 0 | 1 | Burst modulation with shift register (SO) |
| 8 | 1 | 0 | 0 | 0 | FSK modulation with shift register (SO) |
| 9 | 0 | 1 | 1 | 1 | Pulse-width modulation with shift register (SO) & Timer 2 (TOG2), internal trigger restart (SCO) → counter reset |
| 10 | 0 | 1 | 1 | 0 | Manchester demodulation / pulse-width demodulation * (T2O → T3O) |
| 11 | 0 | 1 | 0 | 1 | Biphase demodulation * (T2O → T3O) |
| 12 | 0 | 1 | 0 | 0 | Timer / counter with external capture mode (T3I) |
| 13 | 0 | 0 | 1 | 1 | Not allowed |
| 14 | 0 | 0 | 1 | 0 | Not allowed |
| 15 | 0 | 0 | 0 | 1 | Not allowed |
| 16 | 0 | 0 | 0 | 0 | Not allowed |

* In this mode, the SSI can be used only as demodulator (8-bit NRZ rising edge). All other SSI modes are not allowed.

Timer 3 Control Register 1 (T3C) Write

Primary register address: 'C'hex – Write

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|-------|---|-------|-------|-------|--------------------|
| T3C | T3EIM | T3TOP | T3TS | T3R | Reset value: 0000b |
| T3EIM | Timer 3 Edge Interrupt Mask | | | | |
| | T3EIM = 0, disables the interrupt when an edge event for Timer 3 occurs (T3I) | | | | |
| | T3EIM = 1, enables the interrupt when an edge event for Timer 3 occurs (T3I) | | | | |
| T3TOP | Timer 3 Toggle Output Preset | | | | |
| | T3TOP = 0, sets toggle output (M3) to '0' | | | | |
| | T3TOP = 1, sets toggle output (M3) to '1' | | | | |
| | Note: If T3R = 1, no output preset is possible | | | | |
| T3TS | Timer 3 Toggle with Start | | | | |
| | T3TS = 0, Timer 3 output is not toggled during the start | | | | |
| | T3TS = 1, Timer 3 output is toggled if it is started with T3R | | | | |
| T3R | Timer 3 Run | | | | |
| | T3R = 0, Timer 3 stop and reset | | | | |
| | T3R = 1, Timer 3 run | | | | |

Timer 3 Status Register 1 (T3ST) Read

Primary register address: 'C'hex – Read

| | | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|-------------|-------------|-------|-------------|-------------|-------------|---------------------------|
| T3ST | Read | --- | T3ED | T3C2 | T3C1 | Reset value: x000b |

| | |
|-------------|--|
| T3ED | Timer 3 Edge Detect This bit will be set by the edge-detect logic of Timer 3 input (T3I) |
| T3C2 | Timer 3 Compare 2 This bit will be set when a match occurs between Counter 3 and T3CO2 |
| T3C1 | Timer 3 Compare 1 This bit will be set when a match occurs between Counter 3 and T3CO1 |

Note: The status bits T3C1, T3C2 and T3ED will be reset after a READ access to T3ST.

Timer 3 Clock Select Register (T3CS)

Address: 'B'hex – Subaddress: '1'hex

| | | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|-------------|--|-------------|-------------|--------------|--------------|---------------------------|
| T3CS | | T3E1 | T3E0 | T3CS1 | T3CS0 | Reset value: 1111b |

| | | | | |
|-------------|----------------------------------|-------------|-------------|--|
| T3E1 | Timer 3 Edge select bit 1 | T3E1 | T3E0 | Timer 3 Input Edge Select (T3I) |
| T3E0 | Timer 3 Edge select bit 0 | 1 | 1 | --- |
| | | 1 | 0 | Positive edge at T3I pin |
| | | 0 | 1 | Negative edge at T3I pin |
| | | 0 | 0 | Each edge at T3I pin |

| | | | | |
|--------------|--|--------------|--------------|--|
| T3CS1 | Timer 3 Clock Source select bit 1 | T3CS1 | T3CS0 | Counter 3 Input Signal (CL3) |
| T3CS0 | Timer 3 Clock Source select bit 0 | 1 | 1 | System clock (SYSCL) |
| | | 1 | 0 | Output signal of Timer 2 (POUT) |
| | | 0 | 1 | Output signal of Timer 1 (T1OUT) |
| | | 0 | 0 | External input signal from T3I edge detect |

Timer 3 Compare- and Compare Mode Register

Timer 3 has two separate compare registers T3CO1 and T3CO2 for the 8-bit stage of Timer 3. The timer compares the content of the compare register with the current counter value. If both match, it generates a signal. This signal can be used for the counter reset, to generate a timer interrupt, for toggling the output flip-flop, as SSI clock or as clock for the next counter stage. For each compare register an compare-mode register exists. This registers contain mask bits to enable or disable the generation of an interrupt, a counter reset, or an output toggling with the occurrence of a compare match of the corresponding compare register. The mask bits for activating the single-action mode can also be located in the compare mode registers. When assigned to the compare register a compare event will be suppressed.

Timer 3 Compare Mode Register 1 (T3CM1)

Address: 'B'hex – Subaddress: '2'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|--------------|--|--------------|--------------|--------------|---------------------------|
| T3CM1 | T3SM1 | T3TM1 | T3RM1 | T3IM1 | Reset value: 0000b |
| T3SM1 | Timer 3 Single action Mask bit 1 T3SM1 = 0, disables single-action compare mode T3SM1 = 1, enables single-compare mode. After this bit is set, the compare register (T3CO1) is used until the next compare match. | | | | |
| T3TM1 | Timer 3 compare Toggle action Mask bit 1 T3TM1 = 0, disables compare toggle T3TM1 = 1, enables compare toggle. A match of Counter 3 with the compare register (T3CO1) toggles the output flip-flop (TOG3). | | | | |
| T3RM1 | Timer 3 Reset Mask bit 1 T3RM1 = 0, disables counter reset T3RM1 = 1, enables counter reset. A match of Counter 3 with the compare register (T3CO1) resets the Counter 3. | | | | |
| T3IM1 | Timer 3 Interrupt Mask bit 1 T3RM1 = 0, disables Timer 3 interrupt for T3CO1 register. T3RM1 = 1, enables Timer 3 interrupt for T3CO1 register. | | | | |

T3CM1 contains the mask bits for the match event of the Counter 3 compare register 1

Timer 3 Compare Mode Register 2 (T3CM2)

Address: 'B'hex – Subaddress: '3'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|--------------|--|--------------|--------------|--------------|---------------------------|
| T3CM2 | T3SM2 | T3TM2 | T3RM2 | T3IM2 | Reset value: 0000b |
| T3SM2 | Timer 3 Single action Mask bit 2 T3SM2 = 0, disables single-action compare mode T3SM2 = 1, enables single-compare mode. After this bit is set, the compare register (T3CO2) is used until the next compare match. | | | | |
| T3TM2 | Timer 3 compare Toggle action Mask bit 2 T3TM2 = 0, disables compare toggle T3TM2 = 1, enables compare toggle. A match of Counter 3 with the compare register (T3CO2) toggles the output flip-flop (TOG3). | | | | |
| T3RM2 | Timer 3 Reset Mask bit 2 T3RM2 = 0, disables counter reset T3RM2 = 1, enables counter reset. A match of Counter 3 with the compare register (T3CO2) resets the Counter 3. | | | | |
| T3IM2 | Timer 3 Interrupt Mask bit 2 T3RM2 = 0, disables Timer 3 interrupt for T3CO2 register. T3RM2 = 1, enables Timer 3 interrupt for T3CO2 register. | | | | |

T3CM2 contains the mask bits for the match event of Counter 3 compare register 2

The compare registers and corresponding counter reset masks can be used to program the counter time intervals and the toggle masks can be used to program output signal. The single-action mask can also be used in this mode. It starts operating after the timer started with T3R.

Timer 3 COmpare Register 1 (T3CO1) Byte Write

Address: 'B'hex – Subaddress: '4'hex

| | | | | | | |
|-------|--------------------|-------------|-------|-------|-------|--------------------|
| T3CO1 | Second write cycle | High nibble | | | | Reset value: 1111b |
| | | Bit 7 | Bit 6 | Bit 5 | Bit 4 | |
| | First write cycle | Low nibble | | | | Reset value: 1111b |
| | | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |

Timer 3 COmpare Register 2 (T3CO2) Byte Write

Address: 'B'hex – Subaddress: '5'hex

| | | | | | | |
|-------|--------------------|-------------|-------|-------|-------|--------------------|
| T3CO2 | Second write cycle | High nibble | | | | Reset value: 1111b |
| | | Bit 7 | Bit 6 | Bit 5 | Bit 4 | |
| | First write cycle | Low nibble | | | | Reset value: 1111b |
| | | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |

Timer 3 Capture register

The counter content can be read via the capture register. There are two ways to use the capture register. In modes 1 and 4, it is possible to read the current counter value directly out of the capture register. In the capture modes 2, 3, 5 and 12, a capture event like an edge at the Timer 3 input or a signal from Timer 2 stores the current counter value into the capture register. This counter value can be read from the capture register.

Timer 3 CaPture Register (T3CP) Byte Read

Address: 'B'hex – Subaddress: '4'hex

| | | | | | | |
|------|-------------------|-------------|-------|-------|-------|--------------------|
| T3CP | First read cycle | High nibble | | | | Reset value: xxxxb |
| | | Bit 7 | Bit 6 | Bit 5 | Bit 4 | |
| | Second read cycle | Low nibble | | | | Reset value: xxxxb |
| | | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |

3.3.4 Synchronous Serial Interface (SSI)

SSI Features:

2- and 3-wire NRZ

2-wire mode (I²C compatible)

(additional internal 2-wire link for multi-chip packaging solutions)

- With Timer 2:

Biphase modulation

Manchester modulation

Pulse-width demodulation

Burst modulation

- With Timer 3:

Pulse-width modulation (PWM)

FSK modulation

Biphase demodulation

Manchester demodulation

Pulse-width demodulation

Pulse position Demodulation

SSI Peripheral Configuration

The synchronous serial interface (SSI) can be used either for serial communication with external devices such as EEPROMs, shift registers, display drivers, other microcontrollers, or as a means for generating and capturing on-chip serial streams of data. External data communication takes place via the Port 4 (BP4) multi-functional port which can be software configured by writing the appropriate control word into the P4CR

register. The SSI can be configured in any one of the following ways:

a) 2-wire external interface for bidirectional data communication with one data terminal and one shift clock. The SSI uses the Port BP43 as a bidirectional serial data line (SD) and BP40 as shift clock line (SC).

b) 3-wire external interface for simultaneous input and output of serial data, with a serial input data terminal (SI), a serial output data terminal (SO) and a shift clock (SC). The SSI uses BP40 as shift clock (SC), while the serial data input (SI) is applied to BP43 (configured in P4CR as input!). Serial output data (SO) in this case is passed through to BP42 (configured in P4CR to T2O) via the Timer 2 output stage (T2M2 configured in mode 6).

c) Timer/SSI combined modes – the SSI used together with Timer 2 or Timer 3 is capable of performing a variety of data modulation and demodulation functions (see Timer Section). The modulating data is converted by the SSI into a continuous serial stream of data which is in turn modulated in one of the timer functional blocks. Serial demodulated data can be serially captured in the SSI and read by the controller. In the Timer 3 modes 10 and 11 (demodulation modes) the SSI can only be used as demodulator.

d) Multi-chip link (MCL) – the SSI can also be used as an interchip data interface for use in single package multi-chip modules or hybrids. For such applications, the SSI is provided with two dedicated pads (MCL_SD and MCL_SC) which act as a two-wire chip-to-chip link. The MCL can be activated by the MCL control bit. Should these MCL pads be used by the SSI, the standard SD and SC pins are not required and the corresponding Port 4 ports are available as conventional data ports.

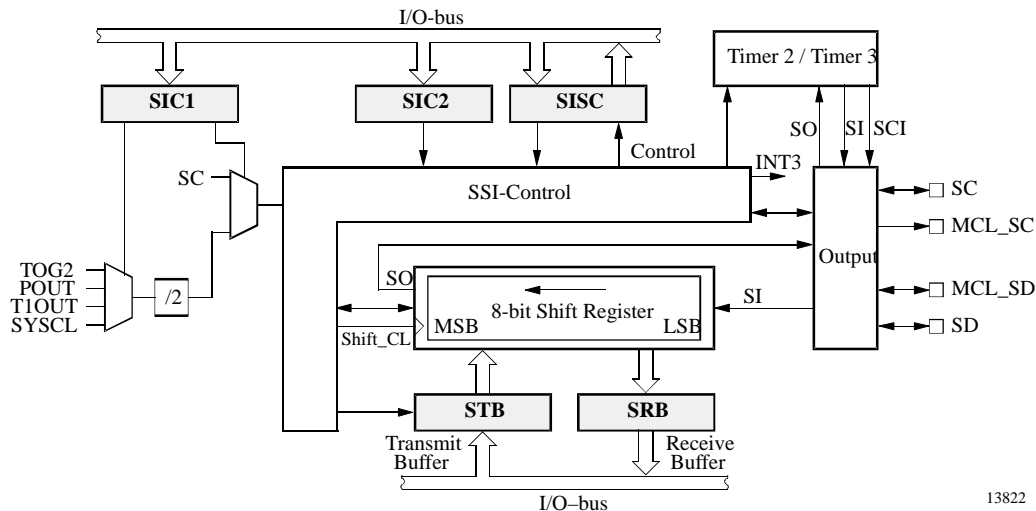


Figure 58. Block diagram of the synchronous serial interface

General SSI Operation

The SSI is comprised essentially of an 8-bit shift register with two associated 8-bit buffers – the receive buffer (SRB) for capturing the incoming serial data and a transmit buffer (STB) for intermediate storage of data to be serially output. Both buffers are directly accessible by software. Transferring the parallel buffer data into and out of the shift register is controlled automatically by the SSI control, so that both single byte transfers or continuous bit streams can be supported.

The SSI can generate the shift clock (SC) either from one of several on-chip clock sources or accept an external clock. The external shift clock is output on, or applied to the Port BP40. Selection of an external clock source is performed by the Serial Clock Direction control bit (SCD). In the combinational modes, the required clock is selected by the corresponding timer mode.

The SSI can operate in three data transfer modes – synchronous 8-bit shift mode, I²C compatible 9-bit shift modes or 8-bit pseudo I²C protocol (without acknowledge-bit).

External SSI clocking is not supported in these modes. The SSI should thus generate and has full control over the shift clock so that it can always be regarded as an I²C Bus Master device.

All directional control of the external data port used by the SSI is handled automatically and is dependent on the transmission direction set by the Serial Data Direction (SDD) control bit. This control bit defines whether the SSI is currently operating in Transmit (TX) mode or Receive (RX) mode.

Serial data is organized in 8-bit telegrams which are shifted with the most significant bit first. In the 9-bit I²C

mode, an additional acknowledge bit is appended to the end of the telegram for handshaking purposes (see I²C protocol).

At the beginning of every telegram, the SSI control loads the transmit buffer into the shift register and proceeds immediately to shift data serially out. At the same time, incoming data is shifted into the shift register input. This incoming data is automatically loaded into the receive buffer when the complete telegram has been received. Data can, if required thus be simultaneously received and transmitted.

Before data can be transferred, the SSI must first be activated. This is performed by means of the SSI reset control (SIR) bit. All further operation then depends on the data directional mode (TX/RX) and the present status of the SSI buffer registers shown by the Serial Interface Ready Status Flag (SRDY). This SRDY flag indicates the (empty/full) status of either the transmit buffer (in TX mode), or the receive buffer (in RX mode). The control logic ensures that data shifting is temporarily halted at any time, if the appropriate receive/transmit buffer is not ready (SRDY = 0). The SRDY status will then automatically be set back to '1' and data shifting resumed as soon as the application software loads the new data into the transmit register (in TX mode) or frees the shift register by reading it into the receive buffer (in RX mode).

A further activity status (ACT) bit indicates the present status of the serial communication. The ACT bit remains high for the duration of the serial telegram or if I²C stop or start conditions are currently being generated. Both the current SRDY and ACT status can be read in the SSI status register. To deactivate the SSI, the SIR bit must be set high.

8-bit Synchronous Mode

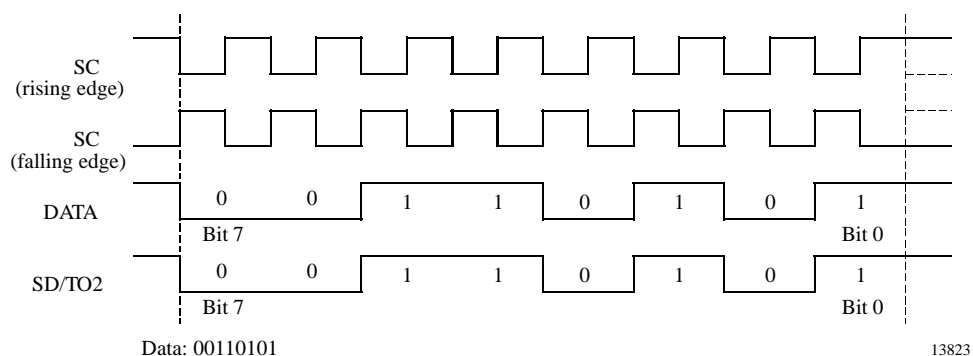


Figure 59. 8-bit synchronous mode

In the 8-bit synchronous mode, the SSI can operate as either a 2- or 3-wire interface (see SSI peripheral configuration). The serial data (SD) is received or transmitted in NRZ format, synchronised to either the rising or falling edge of the shift clock (SC). The choice of clock edge is defined by the Serial Mode Control bits (SM0, SM1). It should be noted that the transmission edge refers to the SC clock edge with which the SD changes. To avoid clock skew problems, the incoming serial input data is shifted in with the opposite edge.

When used together with one of the timer modulator or demodulator stages, the SSI must be set in the 8-bit synchronous mode 1.

In RX mode, as soon as the SSI is activated (SIR= 0), 8 shift clocks are generated and the incoming serial data is shifted into the shift register. This first telegram is automatically transferred into the receive buffer and the SRDY set to 0 indicating that the receive buffer contains valid data. At the same time an interrupt (if enabled) is

generated. The SSI then continues shifting in the following 8-bit telegram. If, during this time the first telegram has been read by the controller, the second telegram will also be transferred in the same way into the receive buffer and the SSI will continue clocking in the next telegram. Should, however, the first telegram not have been read (SRDY=1), then the SSI will stop, temporarily holding the second telegram in the shift register until a certain point of time when the controller is able to service the receive buffer. In this way no data is lost or overwritten.

Deactivating the SSI (SIR=1) in mid-telegram will immediately stop the shift clock and latch the present contents of the shift register into the receive buffer. This can be used for clocking in a data telegram of less than 8 bits in length. Care should be taken to read out the final complete 8-bit data telegram of a multiple word message before deactivating the SSI (SIR=1) and terminating the reception. After termination, the shift register contents will overwrite the receive buffer.

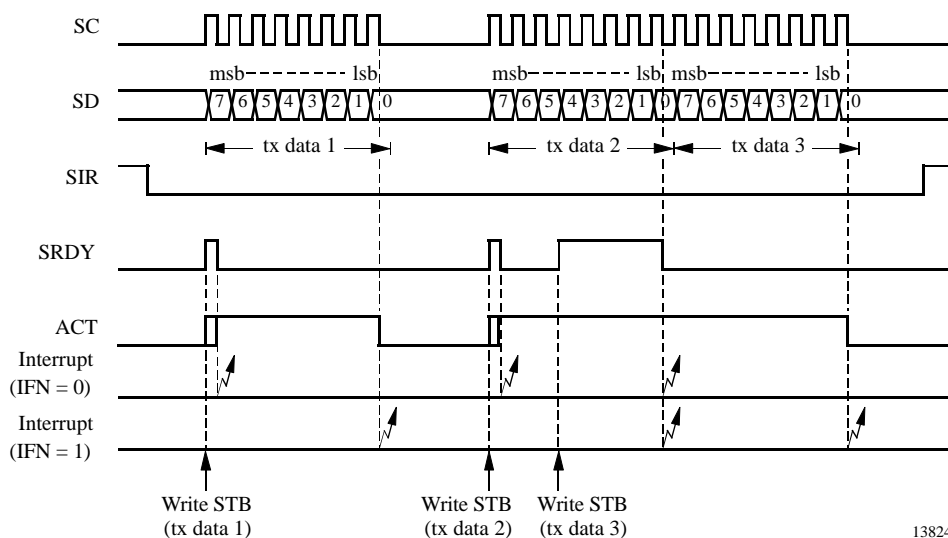
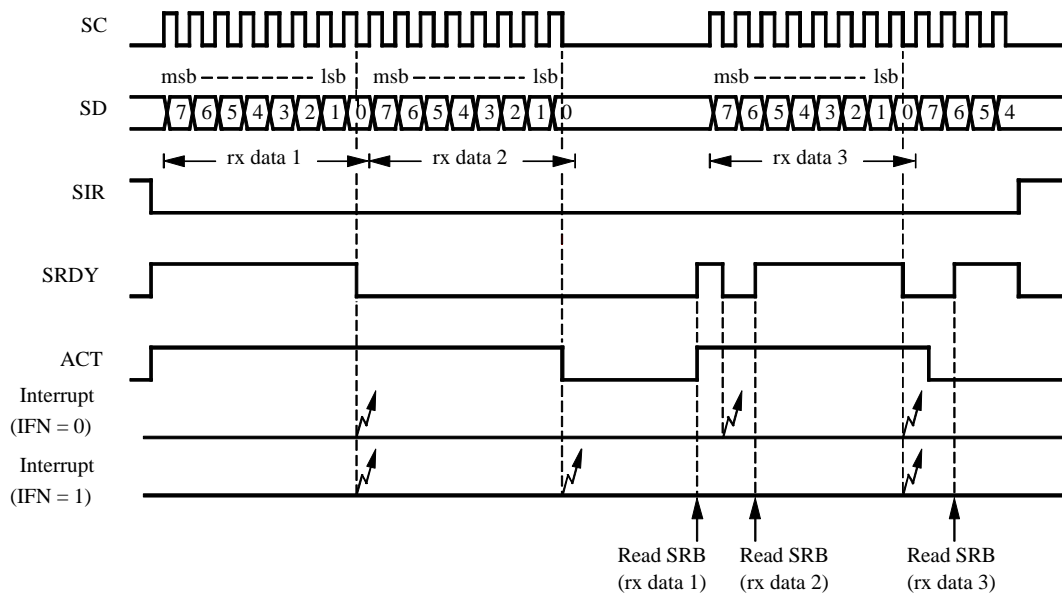


Figure 60. Example of 8-bit synchronous transmit operation



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Figure 61. Example of 8-bit synchronous receive operation

9-bit Shift Mode (I²C compatible)

In the 9-bit shift mode, the SSI is able to handle the I²C protocol (described below). It always operates as an I²C master device, i.e., SC is always generated and output by the SSI. Both the I²C start and stop conditions are automatically generated whenever the SSI is activated or deactivated by the SIR-bit. In accordance with the I²C protocol, the output data is always changed in the clock low phase and shifted in on the high phase.

Before activating the SSI (SIR=0) and commencing an I²C dialog, the appropriate data direction for the first word must be set using the SDD control bit. The state of this bit controls the direction of the data port (BP43 or MCL_SD). Once started, the 8 data bits are, depending on the selected direction, either clocked into or out of the shift register. During the 9th clock period, the port direction is automatically switched over so that the

corresponding acknowledge bit can be shifted out or read in. In transmit mode, the acknowledge bit received from the slave device is captured in the SSI Status Register (TACK) where it can be read by the controller. and in receive mode, the state of the acknowledge bit to be returned to the slave device is predetermined by the SSI Status Register (RACK).

Changing the directional mode (TX/RX) should not be performed during the transfer of an I²C telegram. One should wait until the end of the telegram which can be detected using the SSI interrupt (IFN =1) or by interrogating the ACT status.

A 9-bit telegram, once started will always run to completion and will not be prematurely terminated by the SIR bit. So, if the SIR-bit is set to '1' in mit telegram, the SSI will complete the current transfer and terminate the dialog with an I²C stop condition.

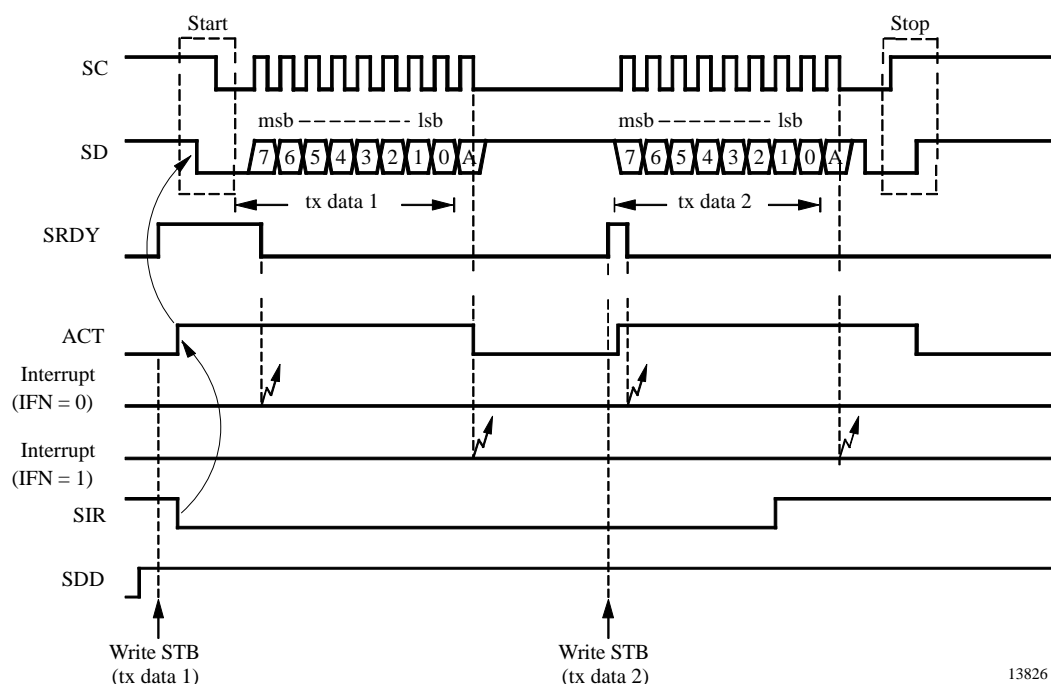


Figure 62. Example of I²C transmit dialog

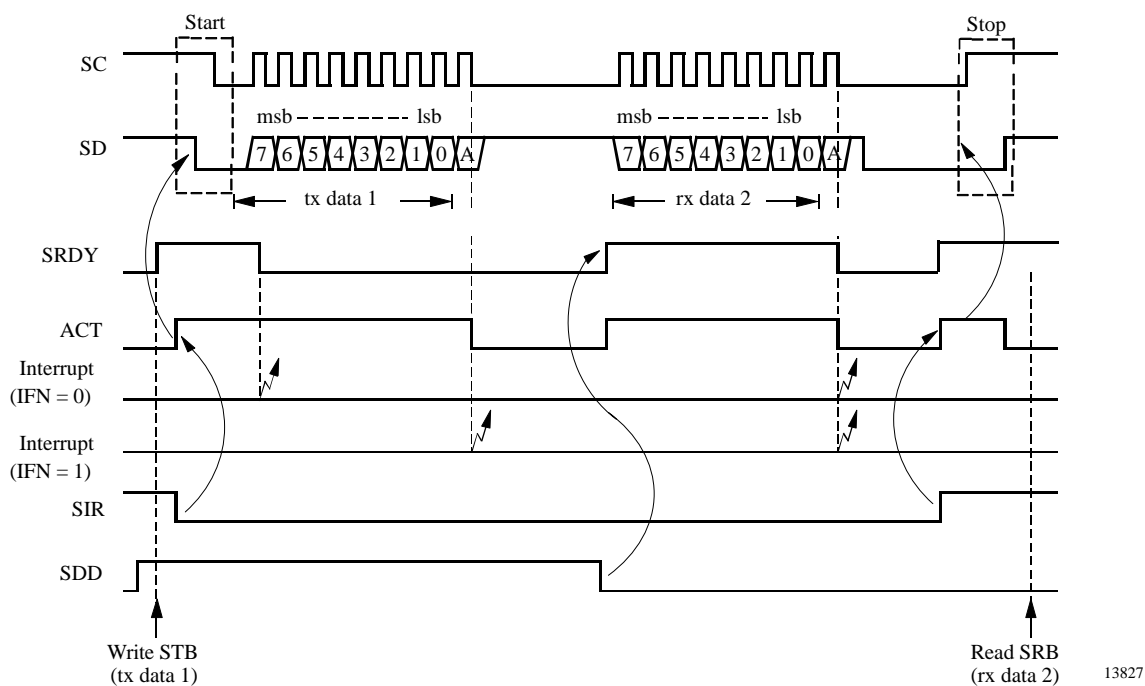


Figure 63. Example of I²C receive dialog

8-bit Pseudo I²C Mode

In this mode, the SSI exhibits all the typical I²C operational features except for the acknowledge-bit which is never expected or transmitted.

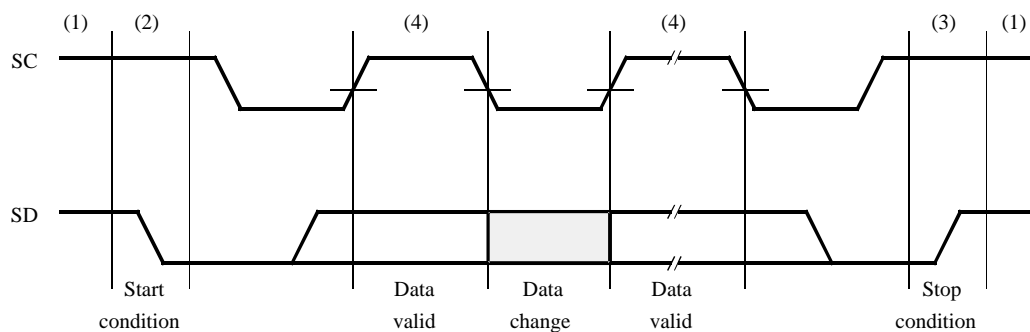
I²C Bus Protocol

The I²C protocol constitutes a simple 2-wire bidirectional communication highway via which devices can communicate control and data information. Although the I²C protocol can support multi-master bus

configurations, the SSI, in I²C mode is intended for use purely as a master controller on a single master bus system. So all reference to multiple bus control and bus contention will be omitted at this point.

All data is packaged into 8-bit telegrams plus a trailing handshaking or acknowledge-bit. Normally the communication channel is opened with a so-called start condition, which initializes all devices connected to the bus. This is then followed by a data telegram, transmitted by the master controller device. This telegram usually contains an 8-bit address code to activate a single slave

device connected onto the I²C bus. Each slave receives this address and compares it with its own unique address. The addressed slave device, if ready to receive data will respond by pulling the SD line low during the 9th clock pulse. This represents a so-called I²C acknowledge. The controller on detecting this affirmative acknowledge then opens a connection to the required slave. Data can then be passed back and forth by the master controller, each 8-bit telegram being acknowledged by the respective recipient. The communication is finally closed by the master device and the slave device put back into standby by applying a stop condition onto the bus.



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Figure 64. I²C bus protocol 1

Bus not busy (1)

Both data and clock lines remain HIGH.

Start data transfer (2)

A HIGH to LOW transition of the SD line while the clock (SC) is HIGH defines a START condition.

Stop data transfer (3)

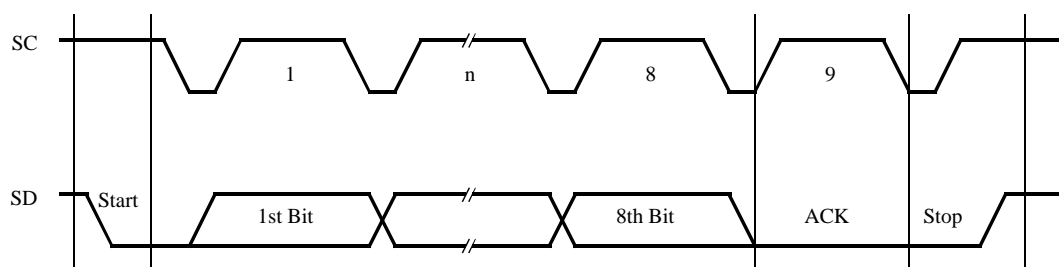
A LOW to HIGH transition of the SD line while the clock (SC) is HIGH defines a STOP condition.

Data valid (4)

The state of the data line represents valid data when, after START condition, the data line is stable for the duration of the HIGH period of the clock signal.

Acknowledge

All address and data words are serially transmitted to and from device in eight-bit words. The receiving device returns a zero on the data line during the ninth clock cycle to acknowledge word receipt.



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Figure 65. I²C bus protocol 2

SSI Interrupt

The SSI interrupt INT3 can be generated either by an SSI buffer register status (i.e., transmit buffer empty or receive buffer full) end of SSI data telegram or on the falling edge of the SC/SD pins on Port 4 (see P4CR). SSI interrupt selection is performed by the Interrupt Function control bit (IFN). The SSI interrupt is usually used to synchronize the software control of the SSI and inform the controller of the present SSI status. The Port 4 interrupts can be used together with the SSI or, if the SSI itself is not required, as additional external interrupt sources. In either case this interrupt is capable of waking the controller out of sleep mode.

To enable and select the SSI relevant interrupts use the SSI interrupt mask (SIM) and the Interrupt Function (IFN) while the Port 4 interrupts are enabled by setting appropriate control bits in P4CR register.

Modulation and Demodulation

If the shift register is used together with Timer 2 or Timer 3 for modulation or demodulation purposes, the 8-bit synchronous mode must be used. In this case, the unused Port 4 pins can be used as conventional bidirectional ports.

The modulation and demodulation stages, if enabled, operate as soon as the SSI is activated (SIR=0) and cease when deactivated (SIR=1).

Due to the byte-orientated data control, the SSI when running normally generates serial bit streams which are submultiples of 8 bits. An SSI output masking (OMSK) function permits, however, the generation of bit streams

of any length. The OMSK signal is derived indirectly from the 4-bit prescaler of the Timer 2 and masks out a programmable number of unrequired trailing data bits during the shifting out of the final data word in the bit stream. The number of non-masked data bits is defined by the value pre-programmed in the prescaler compare register. To use output masking, the modulator stop mode bit (MSM) must be set to '0' before programming the final data word into the SSI transmit buffer. This in turn, enables shift clocks to the prescaler when this final word is shifted out. On reaching the compare value, the prescaler triggers the OMSK signal and all following data bits are blanked.

Internal 2-Wire Multi-Chip Link

Two additional on-chip pads (MCL_SC and MCL_SD) for the SC and the SD line can be used as chip-to-chip link for multi-chip applications. These pads can be activated by setting the MCL-bit in the SISC-register.

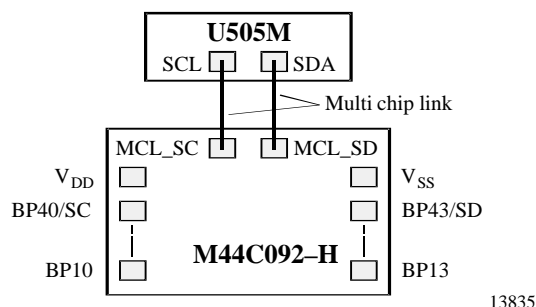


Figure 66. Multi-chip link

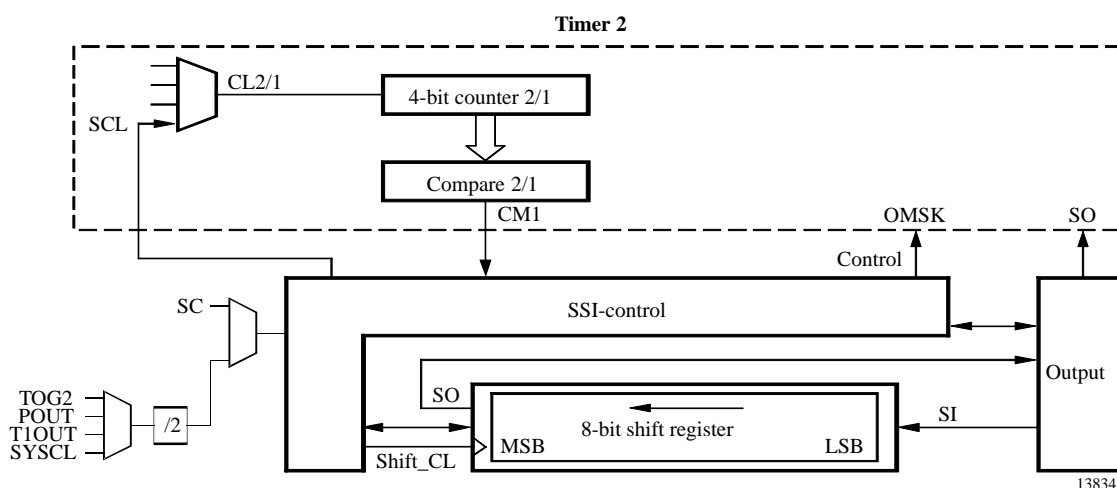


Figure 67. SSI output masking function

Serial Interface Registers

Serial Interface Control Register 1 (SIC1)

Auxiliary register address: '9'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|-------------|------------|------------|-------------|-------------|---------------------------|
| SIC1 | SIR | SCD | SCS1 | SCS0 | Reset value: 1111b |

SIR Serial Interface **R**eset
 SIR = 1, SSI inactive
 SIR = 0, SSI active

SCD Serial Clock **D**irection
 SCD = 1, SC line used as output
 SCD = 0, SC line used as input
Note: This bit has to be set to '1' during the I²C mode and the Timer 3 mode 10 or 11

| SCS1 | Serial Clock source Select bit 1 | SCS1 | SCS0 | Internal Clock for SSI |
|------|----------------------------------|------|------|------------------------|
| SCS0 | Serial Clock source Select bit 0 | 1 | 1 | SYSCL / 2 |
| | | 1 | 0 | T1OUT / 2 |
| | | 0 | 1 | POUT / 2 |
| | | 0 | 0 | TOG2 / 2 |

Note: with SCD = '0' the bits SCS1 and SCS0 are insignificant

- In Transmit mode (SDD = 1) shifting starts only if the transmit buffer has been loaded (SRDY = 1).
- Setting SIR-bit loads the contents of the shift register into the receive buffer (synchronous 8-bit mode only).
- In I²C modes, writing a 0 to SIR generates a start condition and writing a 1 generates a stop condition.

Serial Interface Control Register 2 (SIC2)

Auxiliary register address: 'A'hex

| | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|-------------|------------|------------|------------|------------|---------------------------|
| SIC2 | MSM | SM1 | SM0 | SDD | Reset value: 1111b |

MSM Modular Stop Mode
 MSM = 1, modulator stop mode disabled (output masking off)
 MSM = 0, modulator stop mode enabled (output masking on) – used in modulation modes for generating bit streams which are not sub-multiples of 8 bit.

| SM1 | Serial Mode control bit 1 | Mode | SM1 | SM0 | SSI Mode |
|-----|---------------------------|------|-----|-----|--|
| SM0 | Serial Mode control bit 0 | 1 | 1 | 1 | 8-bit NRZ-Data changes with the rising edge of SC |
| | | 2 | 1 | 0 | 8-bit NRZ-Data changes with the falling edge of SC |
| | | 3 | 0 | 1 | 9-bit two-wire I ² C compatible |
| | | 4 | 0 | 0 | 8-bit two-wire pseudo I ² C compatible (no acknowledge) |

SDD Serial Data **D**irection
 SDD = 1, transmit mode – SD line used as output (transmit data). SRDY is set by a transmit buffer write access.
 SDD = 0, receive mode – SD line used as input (receive data). SRDY is set by a receive buffer read access

Note: SDD controls port directional control and defines the reset function for the SRDY–flag

Serial Interface Status and Control Register (SISC)

Primary register address: 'A'hex

| | | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |
|------|---|-------|-------|-------|-------|--------------------|
| SISC | write | MCL | RACK | SIM | IFN | Reset value: 1111b |
| SISC | read | --- | TACK | ACT | SRDY | Reset value: xxxxb |
| MCL | Multi-Chip Link activation MCL = 1, multi-chip link disabled. This bit has to be set to '0' during transactions to/from EEPROM of the M44C892–H MCL = 0, connects SC and SD additional to the internal multi-chip link pads | | | | | |
| RACK | Receive ACKnowledge status/control bit for I²C mode RACK = 0, transmit acknowledge in next receive telegram RACK = 1, transmit no acknowledge in last receive telegram | | | | | |
| TACK | Transmit ACKnowledge status/control bit for I²C mode TACK = 0, acknowledge received in last transmit telegram TACK = 1, no acknowledge received in last transmit telegram | | | | | |
| SIM | Serial Interrupt Mask SIM = 1, disable interrupts SIM = 0, enable serial interrupt. An interrupt is generated. | | | | | |
| IFN | Interrupt FuNction IFN = 1, the serial interrupt is generated at the end of telegram IFN = 0, the serial interrupt is generated when the SRDY goes low (i.e., buffer becomes empty/full in transmit/receive mode) | | | | | |
| SRDY | Serial interface buffer ReaDY status flag SRDY = 1, in receive mode: receive buffer empty in transmit mode: transmit buffer full SRDY = 0, in receive mode: receive buffer full in transmit mode: transmit buffer empty | | | | | |
| ACT | Transmission ACTive status flag ACT = 1, transmission is active, i.e., serial data transfer. Stop or start conditions are currently in progress. ACT = 0, transmission is inactive | | | | | |

Serial Transmit Buffer (STB) – Byte Write

Primary register address: '9'hex

| | | | | | | |
|-----|--------------------|-------|-------|-------|-------|--------------------|
| STB | First write cycle | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset value: xxxxb |
| | Second write cycle | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Reset value: xxxxb |

The STB is the transmit buffer of the SSI. The SSI transfers the transmit buffer into the shift register and starts shifting with the most significant bit.

Serial Receive Buffer (SRB) – Byte Read

Primary register address: '9'hex

| | | | | | | |
|-----|-------------------|-------|-------|-------|-------|--------------------|
| SRB | First read cycle | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Reset value: xxxxb |
| | Second read cycle | Bit 3 | Bit 2 | Bit 1 | Bit 0 | |

The SRB is the receive buffer of the SSI. The shift register clocks serial data in (most significant bit first) and loads content into the receive buffer when complete telegram has been received.

3.3.5 Combination Modes

The UTCM consists of two timers (Timer 2 and Timer 3) and a serial interface. There is a multitude of modes in which the timers and serial interface can work together.

The 8-bit wide serial interface operates as shift register for modulation and demodulation. The modulator and demodulator units work together with the timers and shift the data bits into or out of the shift register.

Combination Mode Timer 2 and SSI

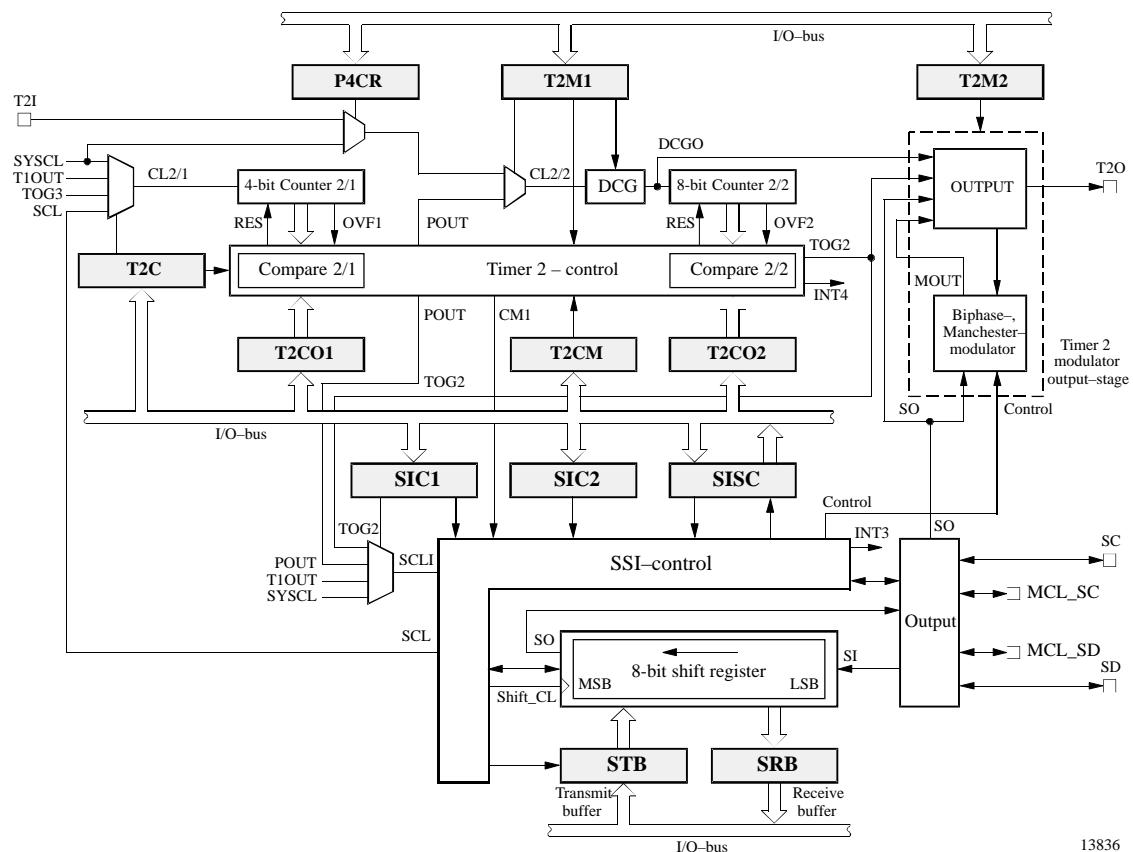
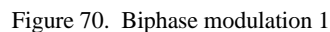
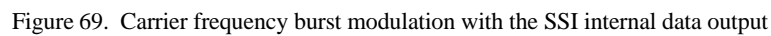


Figure 68. Combination Timer 2 and SSI

Timer 2 output mode 3: Duty cycle burst generator



Combination mode 4: Manchester modulation 2

- SSI mode 1: 8-bit shift register internal data output (SO) to the Timer 2 modulator stage
- Timer 2 mode 3: 8-bit compare counter and 4-bit prescaler
- Timer 2 output mode 5: The modulator 2 of Timer 2 modulates the SSI data output to Manchester code

The 4 bit stage can be used as prescaler for the SSI to generate the stop signal for modulator 2. The SSI has a special mode to supply the prescaler with the shiftclock. The control output signal (OMSK) of the SSI is used as stop signal for the modulator. This is an example for a 12-bit Manchester telegram:

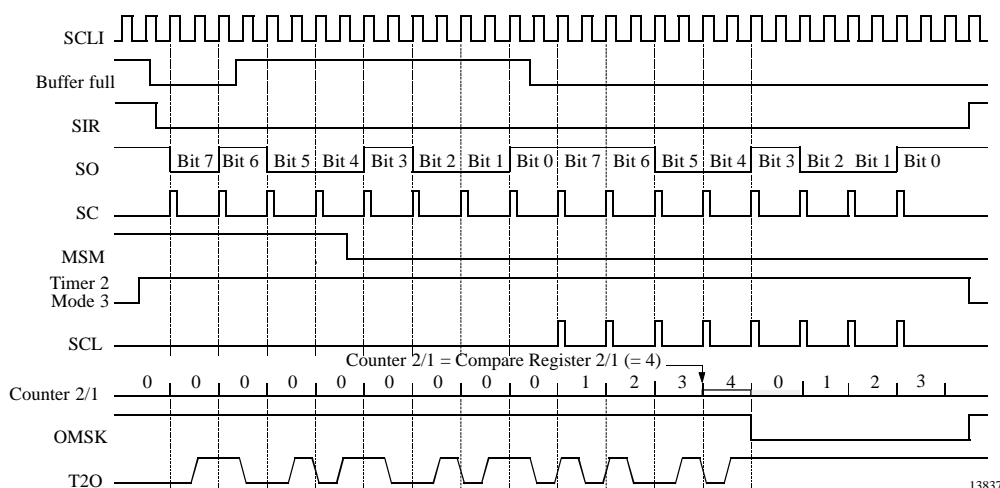


Figure 72. Manchester modulation 2

Combination mode 5: Biphase modulation 2

- SSI mode 1: 8-bit shift register internal data output (SO) to the Timer 2 modulator stage
- Timer 2 mode 3: 8-bit compare counter and 4-bit prescaler
- Timer 2 output mode 4: The modulator 2 of Timer 2 modulates the SSI data output to Biphase code

The 4-bit stage can be used as prescaler for the SSI to generate the stop signal for modulator 2. The SSI has a special mode to supply the prescaler via the shift-clock. The control output signal (OMSK) of the SSI is used as stop signal for the modulator. This is an example for a 13-bit Biphase telegram:

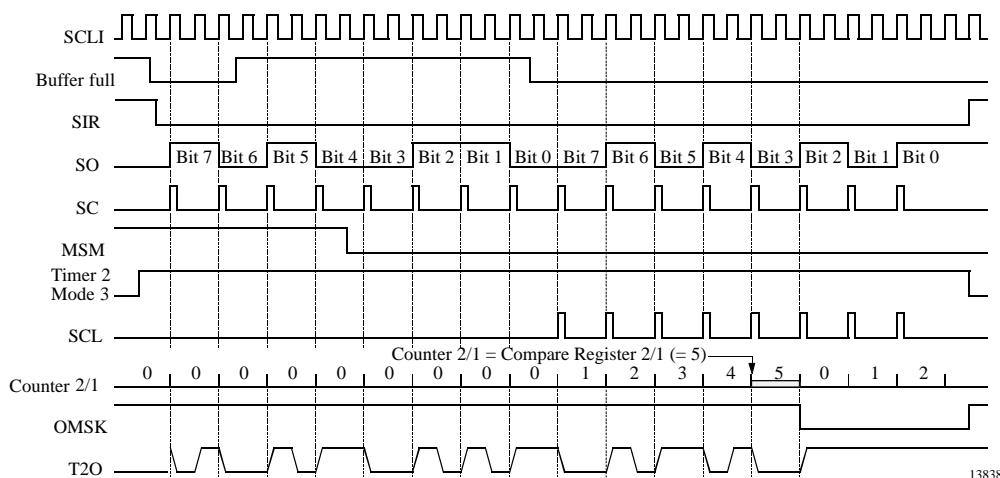


Figure 73. Biphase modulation

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[illegible]

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Combination mode 7: Pulse width modulation (PWM)

SSI mode 1: 8-bit shift register internal data output (SO) to the Timer 3

Timer 3 mode 9: Pulse width modulation with the shift register data (SO)

The two compare registers are used to generate two varied time intervals. The SSI data output selects which compare register is used for the output pulse generation. In this mode both compare and compare mode registers must be programmed to generate the two pulse width. It is also useful to enable the single action mode for extreme duty cycles. Timer 2 is used as baudrate generator and for the triggered restart of Timer 3. The SSI must be supplied with the toggle signal of Timer 2. The counter is driven by an internal or external clock source.

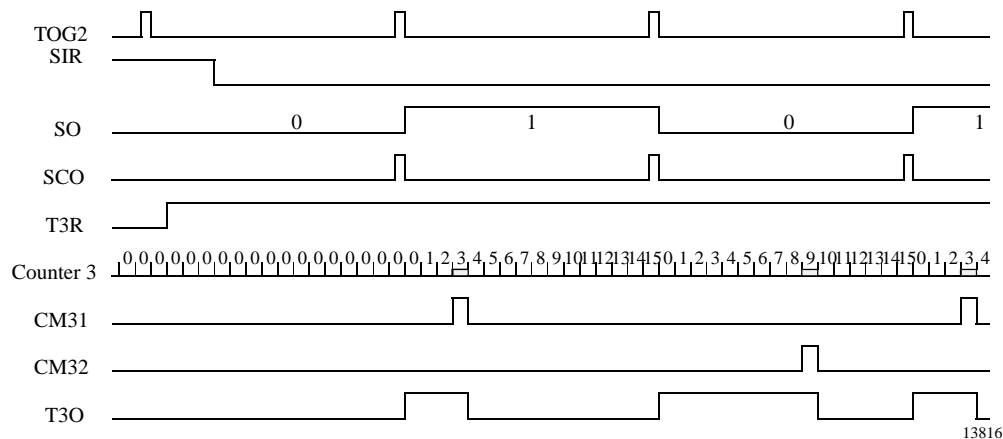


Figure 76. Pulse-width modulation

Combination mode 8: Manchester demodulation / pulse width demodulation

SSI mode 1: 8-bit shift register internal data input (SI) and the internal shift clock (SCI) from the Timer 3

Timer 3 mode 10: Manchester demodulation / pulse width demodulation with Timer 3

For Manchester demodulation the edge detection stage must be programmed to detect each edge at the input. These edges are evaluated by the demodulator stage. The timer stage is used to generate the shift clock for the SSI. A compare register 1 match event defines the correct moment for shifting the state from the input T3I as the decoded bit into shift register. After that the demodulator waits for the next edge to synchronize the timer by a reset for the next bit. The compare register 2 can be used to detect a time error and handle it with an interrupt routine.

Before activating the demodulator mode the timer and the demodulator stage must be synchronized with the bitstream. The Manchester code timing consists of parts with the half bitlength and the complete bitlength. A synchronization routine must start the demodulator after an interval with the complete bitlength.

The counter can be driven by any internal clock source. The output T3O can be used by Timer 2 in this mode. The Manchester decoder can also be used for pulse-width demodulation. The input must be programmed to detect the positive edge. The demodulator and timer must be synchronized with the leading edge of the pulse. After that a counter match with the compare register 1 shifts the state at the input T3I into the shift register. The next positive edge at the input restarts the timer.

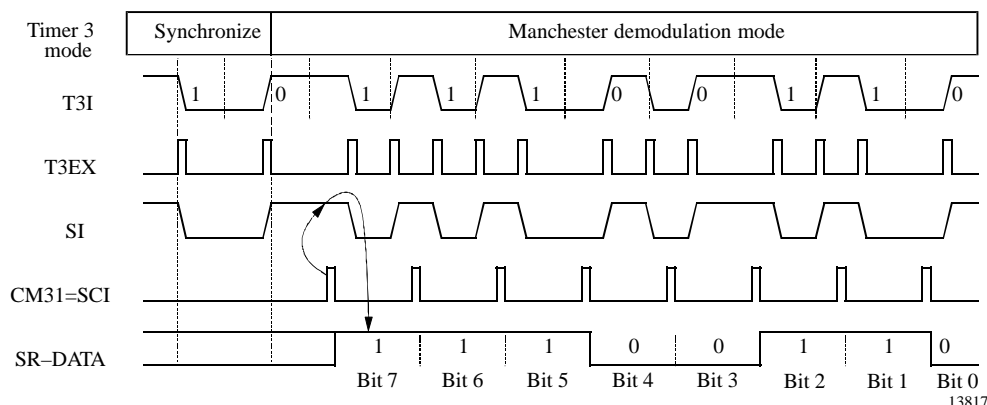


Figure 77. Manchester demodulation

Combination mode 9: Biphase demodulation

SSI mode 1: 8-bit shift register internal data input (SI) and the internal shift clock (SCI) from the Timer 3

Timer 3 mode 11: Biphase demodulation with Timer 3

In the Biphase demodulation mode the timer works like in the Manchester demodulation mode. The difference is that the bits are decoded with the toggle flip-flop. This flip-flop samples the edge in the middle of the bitframe and the compare register 1 match event shifts the toggle flip-flop output into shift register. Before activating the demodulation the timer and the demodulation stage must be synchronized with the bitstream. The Biphase code timing consists of parts with the half bitlength and the complete bitlength. The synchronization routine must start the demodulator after an interval with the complete bitlength.

The counter can be driven by any internal clock source and the output T3O can be used by Timer 2 in this mode.

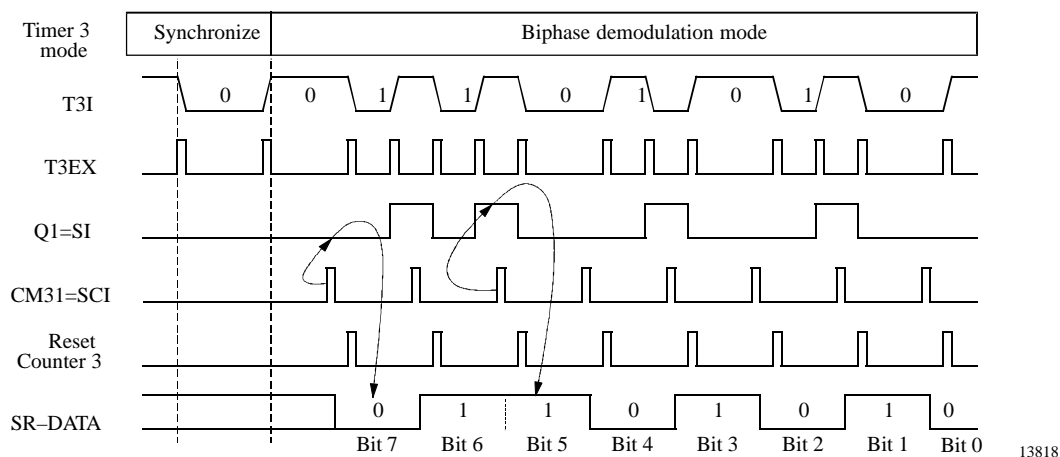


Figure 78. Biphase demodulation

Combination Mode Timer 2 and Timer 3

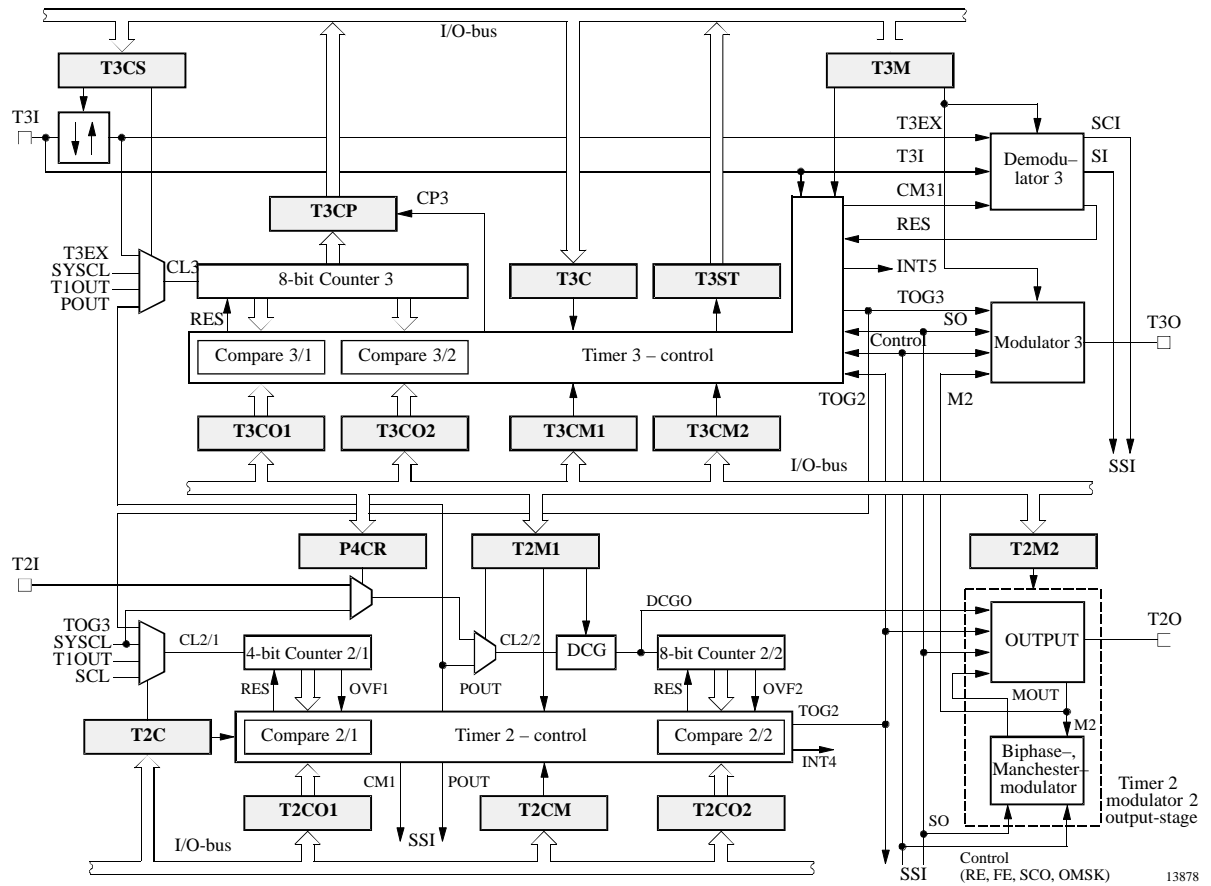


Figure 79. Combination Timer 3 and Timer 2

Combination mode 10: Frequency measurement or event counter with time gate

Timer 2 mode 1/2: 12-bit compare counter / 8-bit compare counter and 4-bit prescaler

Timer 2 output mode 1/6: Timer 2 compare match toggles (TOG2) to the Timer 3

Timer 3 mode 3: Timer / Counter; int. trigger restart & int. capture (with Timer 2 TOG2-signal)

The counter is driven by an external (T3I) clock source. The output signal (TOG2) of Timer 2 resets the counter. The counter value before reset is saved in the capture register. If single-action mode is activated for one or both compare registers, the trigger signal restarts also the single actions. This mode can be used for frequency measurements or as event counter with time gate.

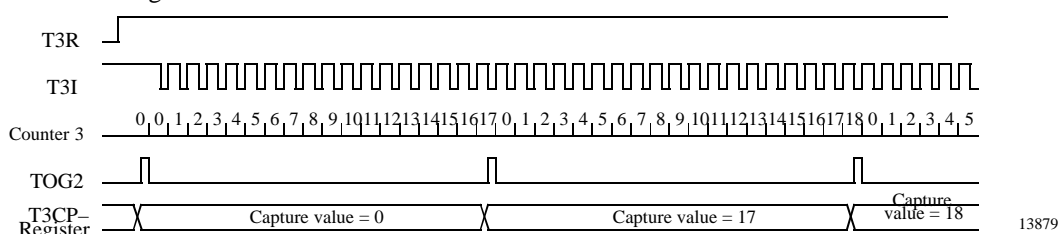


Figure 80. Frequency measurements

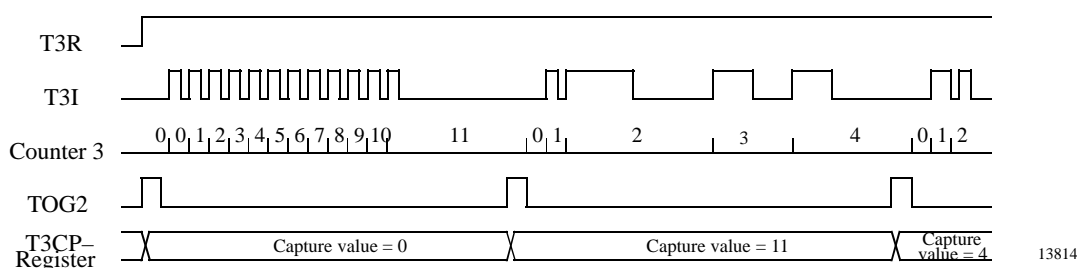


Figure 81. Event counter with time gate

Combination mode 11: Burst modulation 1

Timer 2 mode 1/2: 12-bit compare counter / 8-bit compare counter and 4-bit prescaler

Timer 2 output mode 1/6: Timer 2 compare match toggles the output flip-flop (M2) to the Timer 3

Timer 3 mode 6: Carrier frequency burst modulation controlled by Timer 2 output (M2)

The Timer 3 counter is driven by an internal or external clock source. Its compare- and compare mode registers must be programmed to generate the carrier frequency with the output toggle flip-flop. The output toggle flip-flop (M2) of Timer 2 is used to enable and disable the Timer 3 output. The Timer 2 can be driven by the toggle output signal of Timer 3 (TOG3) or any other clock source.

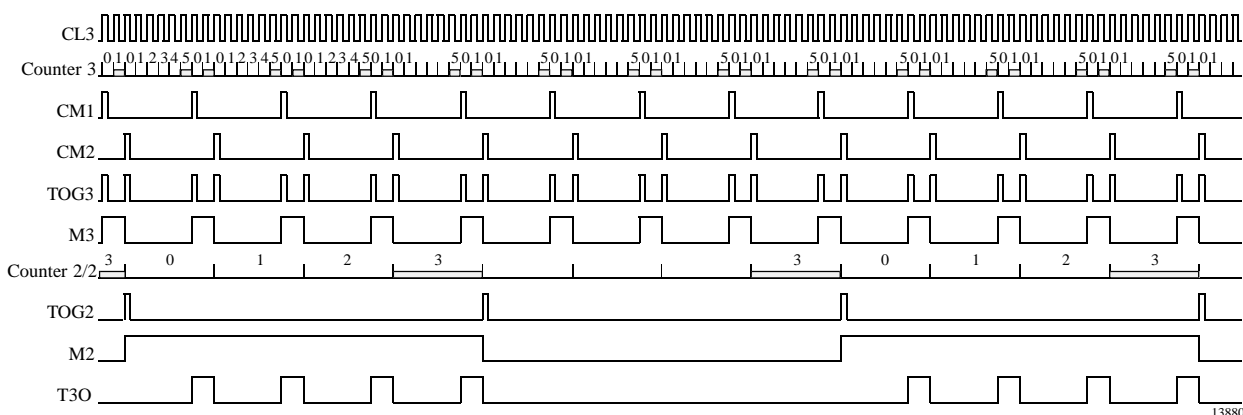


Figure 82. Burst modulation 1

Combination Mode Timer 2, Timer 3 and SSI

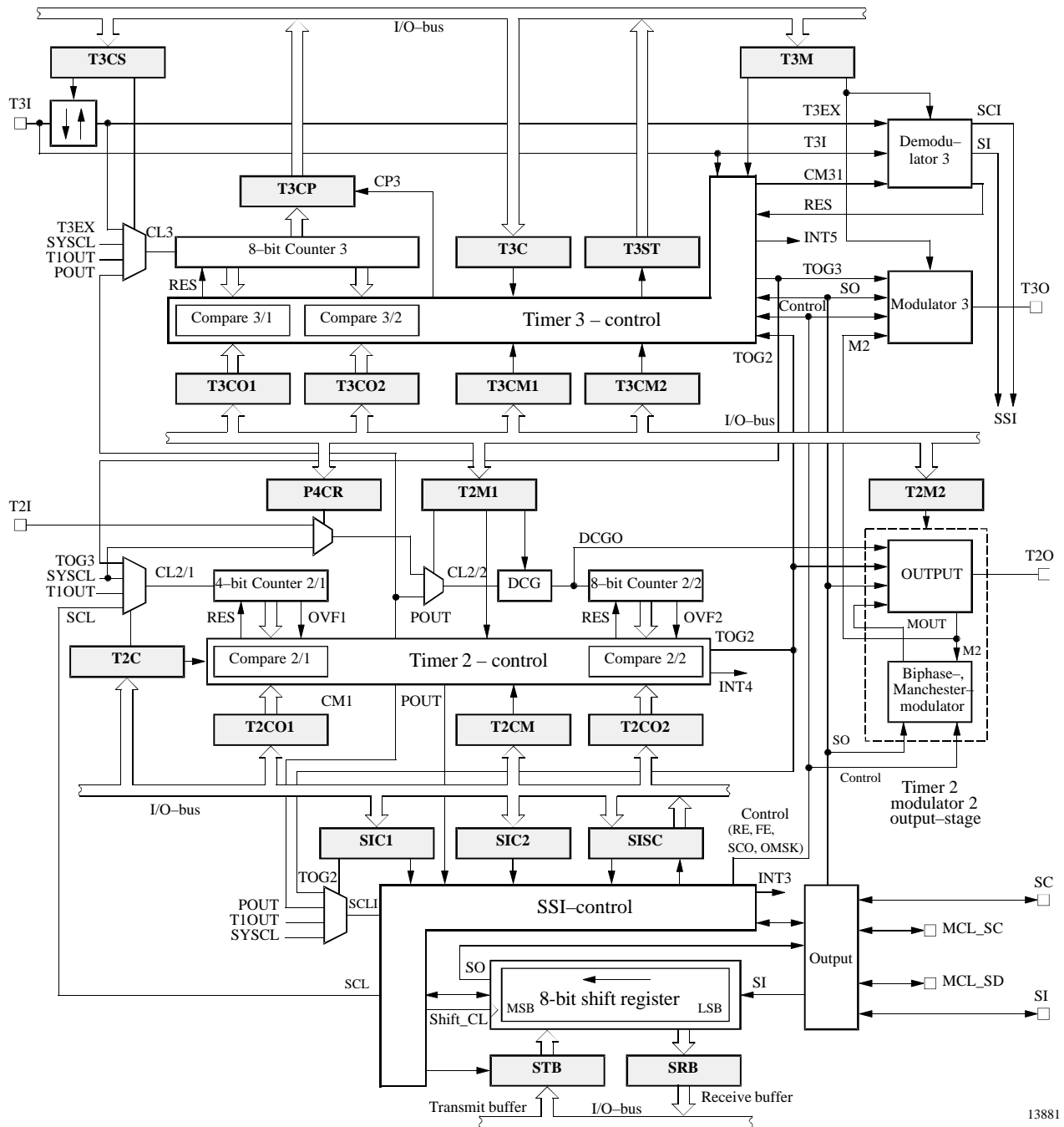


Figure 83. Combination Timer 2, Timer 3 and SSI

4 M44C892-H

The M44C892-H is a multi-chip product which offers a combination of a MARC4-based microcontroller and a serial E2PROM data memory in a single package. As microcontroller the M44C092-H is used and as serial E2PROM the U505M. Two internal lines can be used as chip-to-chip link in a single package. The maximum internal data communication frequency between the M44C092-H and the U505M over the chip link (MCL_SC and MCL_SD) is $f_{SC_MCL} = 500 \text{ kHz}$.

The microcontroller and the EEPROM portions of this multi-chip device are equivalent to their respective individual component chips, except for the electrical specification.

Internal 2-wire multi-chip link

Two additional on-chip pads (MCL_SC and MCL_SD) for the SC and the SD line can be used as chip-to-chip link for multi-chip applications. These pads can be activated by setting the MCL-bit in the SISC-register.

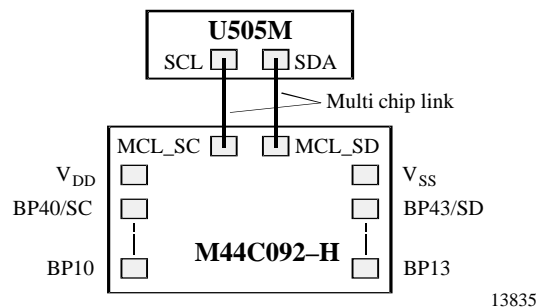


Figure 86. Link between M44C092-H and U505M

4.1 U505M EEPROM

The U505M is a 512-bit EEPROM internally organized 32 x 16 bit. The programming voltage as well as the write-cycle timing is generated on-chip. The U505M features a serial interface allowing operation on a simple two-wire bus with an I²C-compatible protocol. Its low power consumption makes it well suited for battery applications.

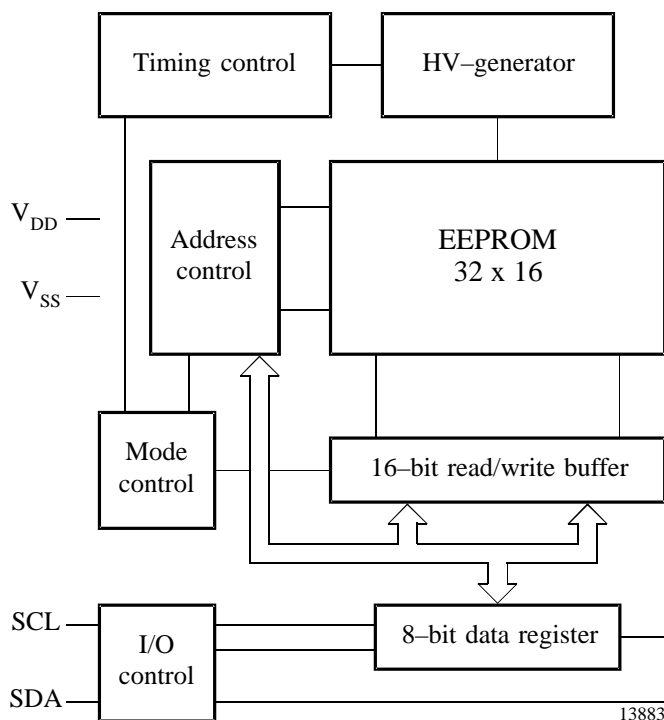


Figure 87. Block diagram EEPROM

4.1.1 Serial Interface

The U505M has an I²C-like two-wire serial interface to the microcontroller for read and write accesses to the EEPROM. The U505M is considered to be a slave in all these applications. That means, the controller has to be the master that initiates the data transfer and provides the clock for transmit and receive operations.

The serial interface is controlled by the M44C892-H microcontroller which generates the serial clock and controls the access via the SCL-line and SDA-line. SCL is used to clock the data into and out of the device. SDA is a bidirectional line that is used to transfer data into and out of the device. The following protocol is used for the data transfers.

Serial Protocol

- Data states on the SDA-line changing only while SCL is low.
- Changes on the SDA-line while SCL is high are interpreted as START or STOP condition.

- A START condition is defined as high to low transition on the SDA-line while the SCL-line is high.
- A STOP condition is defined as low to high transition on the SDA-line while the SCL-line is high.
- Each data transfer must be initialized with a START condition and terminated with a STOP condition. The START condition wakes the device from standby mode and the STOP condition returns the device to standby mode.
- A receiving device generates an acknowledge (A) after the reception of each byte. This requires an additional clock pulse, generated by the master. If the reception was successful the receiving master or slave device pulls down the SDA-line during that clock cycle. If an acknowledge is not detected (N) by the interface in transmit mode, it will terminate further data transmissions and go into receive mode. A master device must finish its read operation by a non-acknowledge and then send a stop condition to bring the device into a known state.

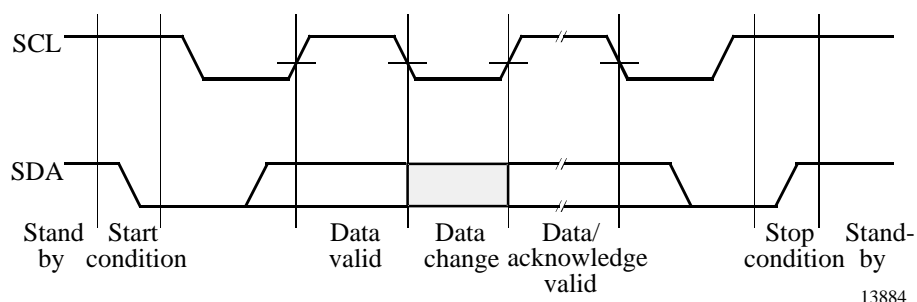


Figure 88. I²C protocol

- Before the START condition and after the STOP condition the device is in stand-by mode and the SDA line is switched as input with pull-up resistor.
- The control byte that follows the START condition de-

termines the following operation. It consists of the 5-bit row address, 2 mode control bits and the READ / NWRITE bit that is used to control the direction of the following transfer. A "0" defines a write access and a "1" a read access.

- **Control byte format:**

| EEPROM address | | | | | | Mode control bits | | Read/ NWrite | |
|----------------|----|----|----|----|----|-------------------|----|--------------|------|
| Start | A4 | A3 | A2 | A1 | A0 | C1 | C0 | R/NW | Ackn |

- **Control byte format:**

| | | | | | | | |
|-------|--------------|------|-----------|------|-----------|------|------|
| Start | Control byte | Ackn | Data byte | Ackn | Data byte | Ackn | Stop |
|-------|--------------|------|-----------|------|-----------|------|------|

4.1.2 EEPROM

The EEPROM has a size of 512 bits and is organized as 32 x 16-bit matrix. To read and write data to and from the EEPROM the serial interface must be used. The interface supports one and two byte write accesses and one to n-byte read accesses to the EEPROM.

EEPROM – Operating Modes

The operating modes of the EEPROM are defined via the control byte. The control byte contains the row address, the mode control bits and the read/not-write bit that is used to control the direction of the following transfer. A "0" defines a write access and a "1" a read access. The five address bits select one of the 32 rows of the EEPROM memory to be accessed. For all accesses the complete 16-bit word of the selected row is loaded into a buffer. The buffer must be read or overwritten via the serial interface. The two mode control bits C1 and C2 define in which order the accesses to the buffer are performed: High byte – low byte or low byte – high byte. The EEPROM also supports autoincrement and autodecrement read operations. After sending the start address with the corresponding mode, consecutive memory cells can be read row by row without transmission of the row addresses.

Write One Data Byte

| | | | | | |
|-------|--------------|---|-------------|---|------|
| Start | Control byte | A | Data byte 1 | A | Stop |
|-------|--------------|---|-------------|---|------|

Write Two Data Bytes

| | | | | | | | |
|-------|--------------|---|-------------|---|-------------|---|------|
| Start | Control byte | A | Data byte 1 | A | Data byte 2 | A | Stop |
|-------|--------------|---|-------------|---|-------------|---|------|

Write Control Byte Only

| | | | |
|-------|--------------|---|------|
| Start | Control byte | A | Stop |
|-------|--------------|---|------|

Write Control Bytes

Write low byte first

| MSB | | | | | LSB | | |
|-------------|----|----|----|----|-----|----|------|
| A4 | A3 | A2 | A1 | A0 | C1 | C0 | R/NW |
| Row address | | | | | 0 | 1 | 0 |

Byte order

| | |
|-------|-------|
| LB(R) | HB(R) |
|-------|-------|

Write high byte first

| MSB | | | | | LSB | | |
|-------------|----|----|----|----|-----|----|------|
| A4 | A3 | A2 | A1 | A0 | C1 | C0 | R/NW |
| Row address | | | | | 1 | 0 | 0 |

Byte order

| | |
|-------|-------|
| HB(R) | LB(R) |
|-------|-------|

A → acknowledge; HB: high byte; LB: low byte; R: row address

Two special control bytes enable the complete initialization of EEPROM with "0" or with "1".

Write Operations

The EEPROM permits 8-bit and 16-bit write operations. A write access starts with the START condition followed by a write control byte and one or two data bytes from the master. It is completed via the STOP condition from the master after the acknowledge cycle.

The programming cycle consists of an erase cycle (write "zeros") and the write cycle (write "ones"). Both cycles together take about 10 ms.

Acknowledge Polling

If the EEPROM is busy with an internal write cycle, all inputs are disabled and the EEPROM will not acknowledge until the write cycle is finished. This can be used to detect the end of the write cycle. The master must perform acknowledge polling by sending a start condition followed by the control byte. If the device is still busy with the write cycle, it will not return an acknowledge and the master has to generate a stop condition or perform further acknowledge polling sequences. If the cycle is complete, it returns an acknowledge and the master can proceed with the next read or write cycle.

Read Operations

The EEPROM allows byte-, word- and current address read operations. The read operations are initiated in the same way as write operations. Every read access is initiated by sending the START condition followed by the control byte which contains the address and the read mode. After the device receives a read command it returns an acknowledge, loads the addressed word into the read/write buffer and sends the selected data byte to the master. The master has to acknowledge the received byte

if it wants to proceed the read operation. If two bytes are read out from the buffer the device increments respectively decrements the word address automatically and loads the buffer with the next word. The read mode bits determines if the low or high byte is read first from the buffer and if the word address is incremented or decremented for the next read access. If the memory address limit is reached, the data word address will "roll over" and the sequential read will continue. The master can terminate the read operation after every byte by not responding with an acknowledge (N) and by issuing a stop condition.

Read One Data Byte

| | | | | | |
|-------|--------------|---|-------------|---|------|
| Start | Control byte | A | Data byte 1 | N | Stop |
|-------|--------------|---|-------------|---|------|

Read Two Data Bytes

| | | | | | | | |
|-------|--------------|---|-------------|---|-------------|---|------|
| Start | Control byte | A | Data byte 1 | A | Data byte 2 | N | Stop |
|-------|--------------|---|-------------|---|-------------|---|------|

Read n Data Bytes

| | | | | | | | | | | |
|-------|--------------|---|-------------|---|-------------|---|-----|-------------|---|------|
| Start | Control byte | A | Data byte 1 | A | Data byte 2 | A | --- | Data byte n | N | Stop |
|-------|--------------|---|-------------|---|-------------|---|-----|-------------|---|------|

Read Control Bytes

| Read low byte first, address increment | MSB | | | | | LSB | | |
|--|-------------|----|----|----|----|-----|----|------|
| | A4 | A3 | A2 | A1 | A0 | C1 | C0 | R/NW |
| | Row address | | | | | 0 | 1 | 1 |

Byte order

| | | | | | | |
|-------|-------|---------|---------|-----|---------|---------|
| LB(R) | HB(R) | LB(R+1) | HB(R+1) | --- | LB(R+n) | HB(R+n) |
|-------|-------|---------|---------|-----|---------|---------|

| Read high byte first, addr. decrement | MSB | | | | | LSB | | |
|---------------------------------------|-------------|----|----|----|----|-----|----|------|
| | A4 | A3 | A2 | A1 | A0 | C1 | C0 | R/NW |
| | Row address | | | | | 1 | 0 | 1 |

Byte order

| | | | | | | |
|-------|-------|---------|---------|-----|---------|---------|
| HB(R) | LB(R) | HB(R–1) | LB(R–1) | --- | HB(R–n) | LB(R–n) |
|-------|-------|---------|---------|-----|---------|---------|

A → acknowledge, N → no acknowledge; HB: high byte; LB: low byte, R: row address

Initialization after a Reset Condition

The EEPROM with the serial interface has its own reset circuitry. In systems with microcontrollers that have their own reset circuitry for power on reset, watchdog reset or brown-out reset, it may be necessary to bring the U505M into a known state independent of its internal reset. This is performed by writing:

| | | | | | |
|-------|--------------|---|-------------|---|------|
| Start | Control byte | A | Data byte 1 | N | Stop |
|-------|--------------|---|-------------|---|------|

to the serial interface. If the U505M acknowledges this sequence it is in a defined state. Maybe it is necessary to perform this sequence twice.

5 Electrical Characteristics

5.1 Absolute Maximum Ratings

Voltages are given relative to V_{SS}

| Parameters | Symbol | Value | Unit |
|--|-------------|--|------|
| Supply voltage | V_{DD} | –0.3 to + 6.5 | V |
| Input voltage (on any pin) | V_{IN} | $V_{SS} - 0.3 \leq V_{IN} \leq V_{DD} + 0.3$ | V |
| Output short circuit duration | t_{short} | indefinite | s |
| Operating temperature range | T_{amb} | –40 to +105 | °C |
| Storage temperature range | T_{stg} | –40 to +130 | °C |
| Thermal resistance (SSO20) | R_{thJA} | 140 | K/W |
| Soldering temperature ($t \leq 10$ s) | T_{sld} | 260 | °C |

Stresses greater than those listed under absolute maximum ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at any condition above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating condition for an extended period may affect device

reliability. All inputs and outputs are protected against high electrostatic voltages or electric fields. However, precautions to minimize the build-up of electrostatic charges during handling are recommended. Reliability of operation is enhanced if unused inputs are connected to an appropriate logic voltage level (e.g. V_{DD}).

5.2 DC Operating Characteristics

$V_{SS} = 0$ V, $T_{amb} = -40$ to 105°C unless otherwise specified.

| Parameters | Test Conditions / Pins | Symbol | Min. | Typ. | Max. | Unit |
|--|--|-------------|-----------|--------------------------|-------------------|--|
| Power supply | | | | | | |
| Operating voltage at V_{DD} | | V_{DD} | V_{POR} | | 6.5 | V |
| Active current CPU active | $f_{SYSCL} = 1$ MHz $V_{DD} = 1.8$ V $V_{DD} = 3.0$ V $V_{DD} = 6.5$ V | I_{DD} | | 200 300 700 | 450 | μA μA μA |
| Power down current (CPU sleep, RC oscillator active, 4-MHz quartz-osc. active) | $f_{SYSCL} = 1$ MHz $V_{DD} = 1.8$ V $V_{DD} = 3.0$ V $V_{DD} = 6.5$ V | I_{PD} | | 40 70 200 | 180 | μA μA μA |
| Sleep current (CPU sleep, 32-kHz quartz-osc. active 4-MHz quartz-osc. inactive) | $V_{DD} = 1.8$ V $V_{DD} = 3.0$ V $V_{DD} = 6.5$ V | I_{Sleep} | | 0.4 0.6 0.8 | 1.3 1.8 | μA μA μA |
| Sleep current (CPU sleep, 32-kHz quartz-osc. inactive 4-MHz quartz-osc. inactive) | $V_{DD} = 1.8$ V $V_{DD} = 3.0$ V $V_{DD} = 6.5$ V for M44C092–H $V_{DD} = 6.5$ V for M44C892–H | I_{Sleep} | | 0.1 0.3 0.5 0.6 | 0.5 0.8 1.0 | μA μA μA μA |
| Pin capacitance | Any pin to V_{SS} | C_L | | 7 | 10 | pF |

$V_{SS} = 0\text{ V}$, $T_{amb} = -40$ to 105°C unless otherwise specified.

| Parameters | Test Conditions / Pins | Symbol | Min. | Typ. | Max. | Unit |
|--|--------------------------------------|------------|------|------|------|------|
| Power-on reset threshold voltage | | | | | | |
| POR threshold voltage | BOT = 1 | V_{POR} | 1.6 | 1.7 | 1.8 | V |
| POR threshold voltage | BOT = 0 | V_{POR} | 1.85 | 2.0 | 2.15 | V |
| POR hysteresis | | V_{POR} | | 50 | | mV |
| Voltage monitor threshold voltage | | | | | | |
| VM high threshold voltage | $V_{DD} > V_M$, $V_{MS} = 1$ | V_{MThh} | | 3.0 | 3.25 | V |
| VM high threshold voltage | $V_{DD} < V_M$, $V_{MS} = 0$ | V_{MThh} | 2.8 | 3.0 | | V |
| VM middle thresh. voltage | $V_{DD} > V_M$, $V_{MS} = 1$ | V_{MThm} | | 2.6 | 2.8 | V |
| VM middle thresh. voltage | $V_{DD} < V_M$, $V_{MS} = 0$ | V_{MThm} | 2.4 | 2.6 | | V |
| VM low threshold voltage | $V_{DD} > V_M$, $V_{MS} = 1$ | V_{MThl} | | 2.2 | 2.4 | V |
| VM low threshold voltage | $V_{DD} < V_M$, $V_{MS} = 0$ | V_{MThl} | 2.0 | 2.2 | | V |
| External input voltage | | | | | | |
| VMI | $V_{DD} = 3\text{ V}$, $V_{MS} = 1$ | V_{VMI} | | 1.3 | 1.4 | V |
| VMI | $V_{DD} = 3\text{ V}$, $V_{MS} = 0$ | V_{VMI} | 1.2 | 1.3 | | V |

All Bidirectional Ports

$V_{SS} = 0\text{ V}$, $T_{amb} = -40$ to 105°C unless otherwise specified.

| Parameters | Test Conditions / Pins | Symbol | Min. | Typ. | Max. | Unit |
|--|---|----------|--------------------|------|--------------------|---------------|
| Input voltage LOW | $V_{DD} = 1.8$ to 6.5 V | V_{IL} | V_{SS} | | $0.2 \cdot V_{DD}$ | V |
| Input voltage HIGH | $V_{DD} = 1.8$ to 6.5 V | V_{IH} | $0.8 \cdot V_{DD}$ | | V_{DD} | V |
| Input LOW current (switched pull-up) | $V_{DD} = 2.0\text{ V}$ | I_{IL} | -2 | -4 | -12 | μA |
| | $V_{DD} = 3.0\text{ V}$, $V_{IL} = V_{SS}$ | | -10 | -20 | -40 | μA |
| | $V_{DD} = 6.5\text{ V}$ | | -50 | -100 | -200 | μA |
| Input HIGH current (switched pull-down) | $V_{DD} = 2.0\text{ V}$ | I_{IH} | 2 | 4 | 12 | μA |
| | $V_{DD} = 3.0\text{ V}$, $V_{IH} = V_{DD}$ | | 10 | 20 | 40 | μA |
| | $V_{DD} = 6.5\text{ V}$ | | 50 | 100 | 200 | μA |
| Input LOW current (static pull-up) | $V_{DD} = 2.0\text{ V}$ | I_{IL} | -20 | -50 | -100 | μA |
| | $V_{DD} = 3.0\text{ V}$, $V_{IL} = V_{SS}$ | | -80 | -160 | -320 | μA |
| | $V_{DD} = 6.5\text{ V}$ | | -300 | -600 | -1200 | μA |
| Input LOW current (static pull-down) | $V_{DD} = 2.0\text{ V}$ | I_{IH} | 20 | 50 | 100 | μA |
| | $V_{DD} = 3.0\text{ V}$, $V_{IH} = V_{DD}$ | | 80 | 160 | 320 | μA |
| | $V_{DD} = 6.5\text{ V}$ | | 300 | 600 | 1200 | μA |
| Input leakage current | $V_{IL} = V_{SS}$ | I_{IL} | | | 100 | nA |
| Input leakage current | $V_{IH} = V_{DD}$ | I_{IH} | | | 100 | nA |
| Output LOW current | $V_{OL} = 0.2 \cdot V_{DD}$ | I_{OL} | | | | |
| | $V_{DD} = 2.0\text{ V}$ | | 0.6 | 1.2 | 2.5 | mA |
| | $V_{DD} = 3.0\text{ V}$ | | 3 | 5 | 8 | mA |
| | $V_{DD} = 6.5\text{ V}$ | | 8 | 15 | 22 | mA |
| Output HIGH current | $V_{OH} = 0.8 \cdot V_{DD}$ | I_{OH} | | | | |
| | $V_{DD} = 2.0\text{ V}$ | | -0.6 | -1.2 | -2.5 | mA |
| | $V_{DD} = 3.0\text{ V}$ | | -3 | -5 | -8 | mA |
| | $V_{DD} = 6.5\text{ V}$ | | -8 | -16 | -24 | mA |

Note: The Pin BP20/NTE has a static pull-up resistor during the reset-phase of the microcontroller

5.3 AC Characteristics

Supply voltage $V_{DD} = 1.8$ to 6.5 V, $V_{SS} = 0$ V, $T_{amb} = 25^{\circ}\text{C}$ unless otherwise specified.

| Parameters | Test Conditions / Pins | Symbol | Min. | Typ. | Max. | Unit |
|---|---|--------------|----------------------|------|-----------|---------------|
| System clock cycle | $V_{DD} = 1.8$ to 6.5 V $T_{amb} = -40$ to 105°C | t_{SYSCL} | 500 | | 4000 | ns |
| | $V_{DD} = 2.4$ to 6.5 V $T_{amb} = -40$ to 105°C | t_{SYSCL} | 250 | | 4000 | ns |
| | $V_{DD} = 2.4$ to 6.5 V $T_{amb} = -40$ to 85°C | t_{SYSCL} | 500 | | 2000 | ns |
| Timer 2 input timing Pin T2I | | | | | | |
| Timer 2 input clock | | f_{T2I} | | | 5 | MHz |
| Timer 2 input LOW time | Rise / fall time < 10 ns | t_{T2IL} | 100 | | | ns |
| Timer 2 input HIGH time | Rise / fall time < 10 ns | t_{T2IH} | 100 | | | ns |
| Timer 3 input timing Pin T3I | | | | | | |
| Timer 3 input clock | | f_{T3I} | | | $SYSCL/2$ | MHz |
| Timer 3 input LOW time | Rise / fall time < 10 ns | t_{T3IL} | $2 \times t_{SYSCL}$ | | | ns |
| Timer 3 input HIGH time | Rise / fall time < 10 ns | t_{T3IH} | $2 \times t_{SYSCL}$ | | | ns |
| Interrupt request input timing | | | | | | |
| Int. request LOW time | Rise / fall time < 10 ns | t_{IRL} | 100 | | | ns |
| Int. request HIGH time | Rise / fall time < 10 ns | t_{IRH} | 100 | | | ns |
| External system clock | | | | | | |
| EXSCL at OSC1, ECM = EN | Rise / fall time < 10 ns | f_{EXSCL} | 0.5 | | 4 | MHz |
| EXSCL at OSC1, ECM = DI | Rise / fall time < 10 ns | f_{EXSCL} | 0.02 | | 4 | MHz |
| Input HIGH time | Rise / fall time < 10 ns | t_{IH} | 0.1 | | | μs |
| Reset timing | | | | | | |
| Power-on reset time | $V_{DD} > V_{POR}$ | t_{POR} | | 1.5 | 5 | ms |
| RC oscillator 1 | | | | | | |
| Frequency | | f_{RCOut1} | | 3.8 | | MHz |
| Stability | $V_{DD} = 2.0$ to 6.5 V $T_{amb} = -40$ to 105°C | $\Delta f/f$ | | | ± 50 | % |
| RC oscillator 2 – external resistor | | | | | | |
| Frequency | $R_{ext} = 170$ k Ω | f_{RCOut2} | | 4 | | MHz |
| Stability | $V_{DD} = 2.0$ to 6.5 V $T_{amb} = -40$ to 85°C | $\Delta f/f$ | | | ± 15 | % |
| Stabilization time | | t_S | | | 10 | μs |
| 4-MHz crystal oscillator (operating range $V_{DD} = 2.2$ V to 6.5 V) | | | | | | |
| Frequency | | f_X | | 4 | | MHz |
| Start-up time | | t_{SQ} | | 5 | | ms |
| Stability | | $\Delta f/f$ | -10 | | 10 | ppm |
| Integrated input / output capacitances (mask programmable) | C_{IN} / C_{OUT} programmable in steps of 2 pF | C_{IN} | 0 | | 20 | pF |
| | | C_{OUT} | 0 | | 20 | pF |

| Parameters | Test Conditions / Pins | Symbol | Min. | Typ. | Max. | Unit |
|--|--|------------------|------|--------|------|------|
| 32-kHz crystal oscillator (operating range V _{DD} = 2.0 V to 6.5 V) | | | | | | |
| Frequency | | f _X | | 32.768 | | kHz |
| Start-up time | | t _{SQ} | | 0.5 | | s |
| Stability | | Δf/f | −10 | | 10 | ppm |
| Integrated input / output capacitances (mask programmable) | C _{IN} / C _{OUT} programmable in steps of 2 pF | C _{IN} | 0 | | 20 | pF |
| | | C _{OUT} | 0 | | 20 | pF |
| External 32-kHz crystal parameters | | | | | | |
| Crystal frequency | | f _X | | 32.768 | | kHz |
| Serial resistance | | R _S | | 30 | 50 | kΩ |
| Static capacitance | | C ₀ | | 1.5 | | pF |
| Dynamic capacitance | | C ₁ | | 3 | | fF |
| External 4-MHz crystal parameters | | | | | | |
| Crystal frequency | | f _X | | 4.0 | | MHz |
| Serial resistance | | R _S | | 40 | 150 | Ω |
| Static capacitance | | C ₀ | | 1.4 | 3 | pF |
| Dynamic capacitance | | C ₁ | | 3 | | fF |

Crystal Characteristics



96 11553

Figure 89. Crystal equivalent circuit

Supply voltage $V_{DD} = 1.8$ to 6.5 V , $V_{SS} = 0\text{ V}$, $T_{amb} = 25^\circ\text{C}$ unless otherwise specified.

| Parameters | Test Conditions / Pins | Symbol | Min. | Typ. | Max. | Unit |
|--|---|----------------------|-------------------|-----------|------|------------------|
| EEPROM | | | | | | |
| Operating current during erase/write cycle | | I_{WR} | | 600 | 1300 | μA |
| Endurance | Erase- / write-cycles @ 105°C | n_{EW} n_{EW} | 500,000 50,000 | 1,000,000 | | Cycles Cycles |
| Data erase/write cycle time | for 16-bit access | t_{DEW} | | 9 | 12 | ms |
| Data retention time | @ 105°C | t_{DR} t_{DR} | 10 1 | | | years years |
| Power-up to read operation | | t_{PUR} | | | 0.2 | ms |
| Power-up to write operation | | t_{PUW} | | | 0.2 | ms |
| Serial interface | | | | | | |
| SCL clock frequency | | f_{SC_MCL} | | 100 | 500 | kHz |

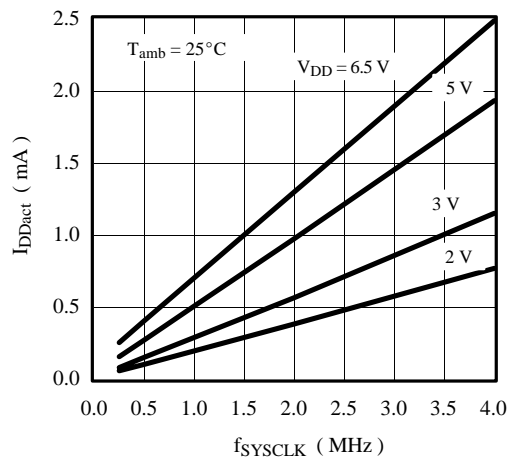


Figure 90. Active supply current vs. frequency

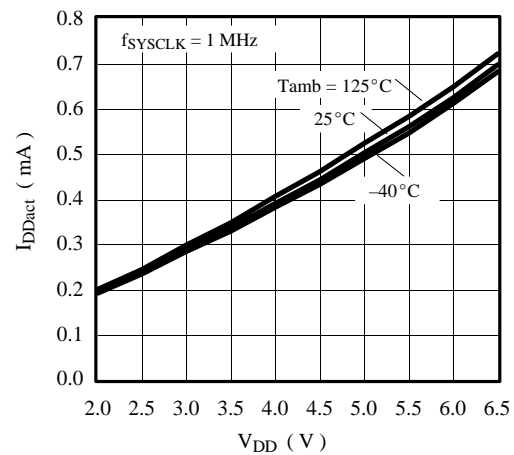


Figure 93. Active supply current vs. V_{DD}

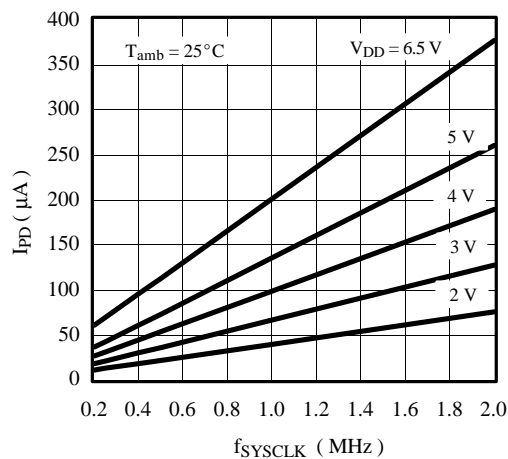


Figure 91. Power-down supply current vs. frequency

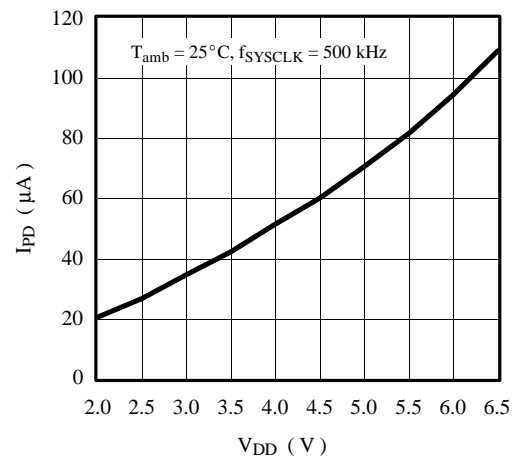


Figure 94. Power-down supply current vs. V_{DD}

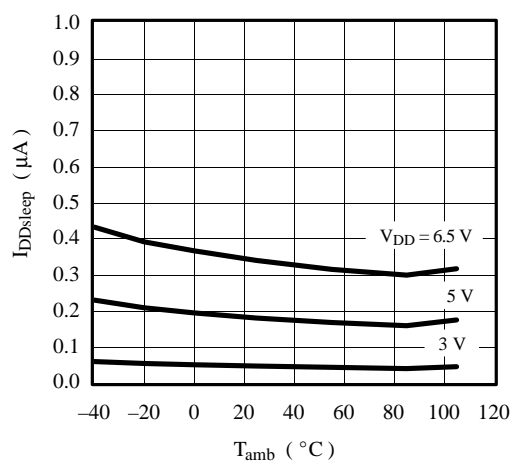


Figure 92. Sleep current vs. T_{amb}
M44C092-H

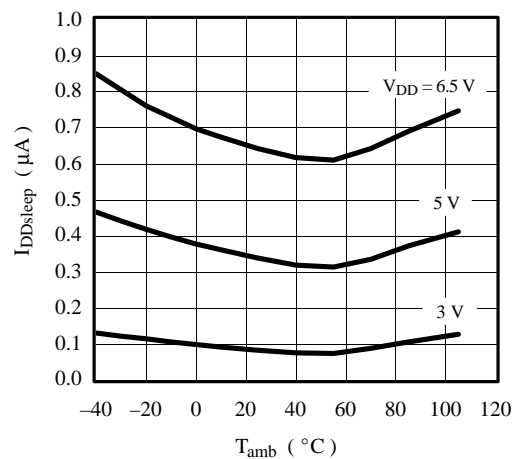


Figure 95. Sleep current vs. T_{amb}
M44C892-H

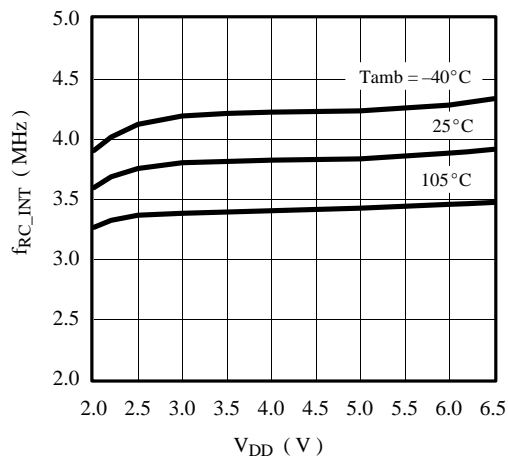


Figure 96. Internal RC frequency vs. V_{DD}

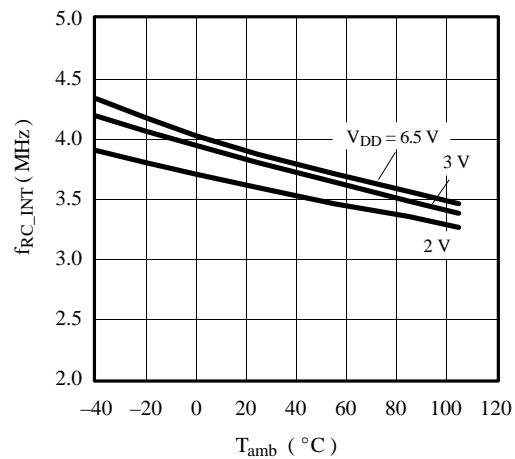


Figure 99. Internal RC frequency vs. T_{amb}

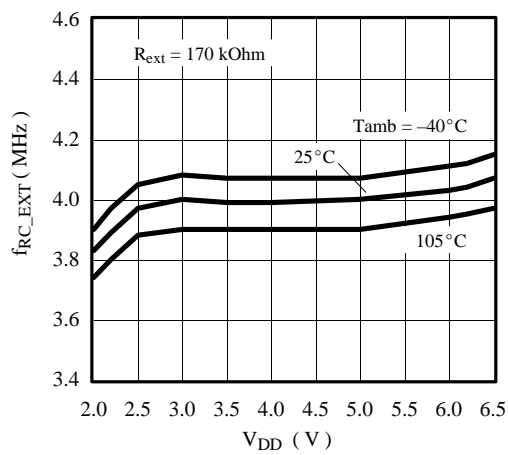


Figure 97. External RC frequency vs. V_{DD}

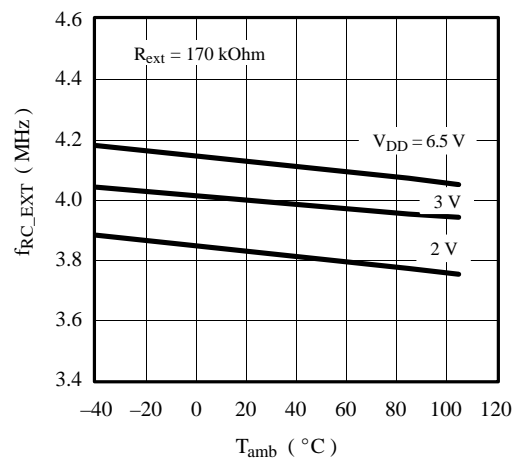


Figure 100. External RC frequency vs. T_{amb}

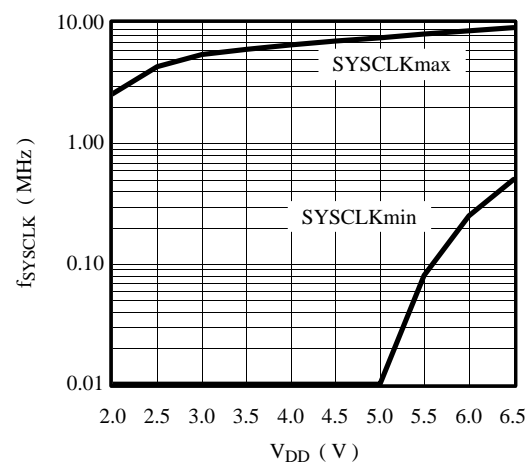


Figure 98. System clock vs. V_{DD}

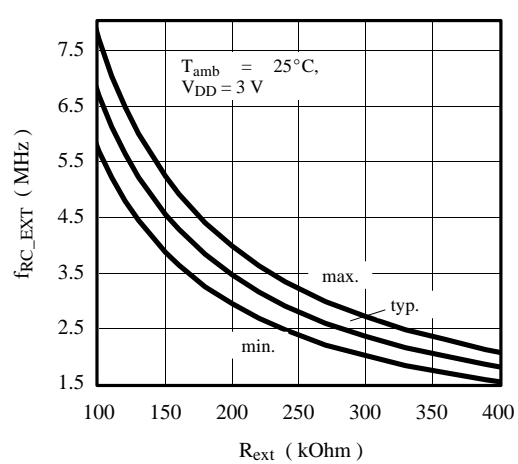


Figure 101. External RC frequency vs. R_{ext}

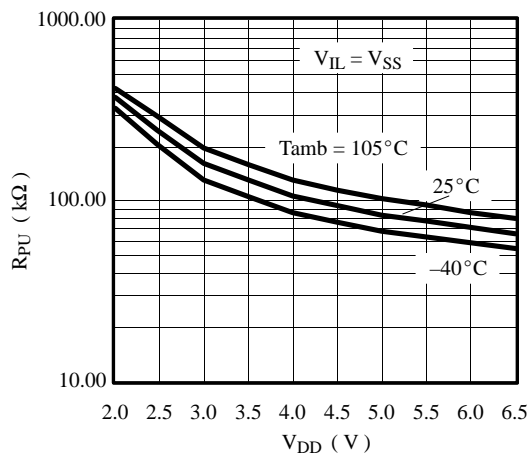


Figure 102. Pull-up resistor vs. V_{DD}

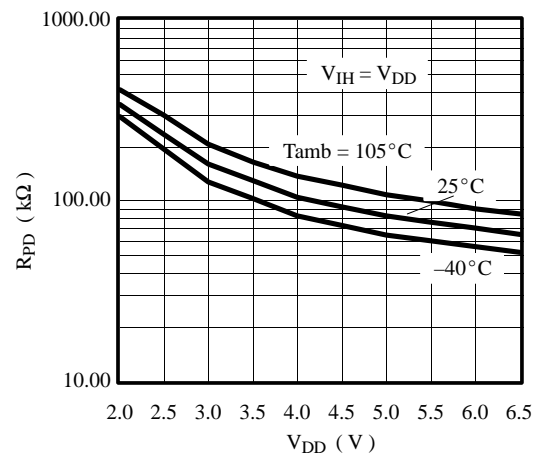


Figure 105. Pull-down resistor vs. V_{DD}

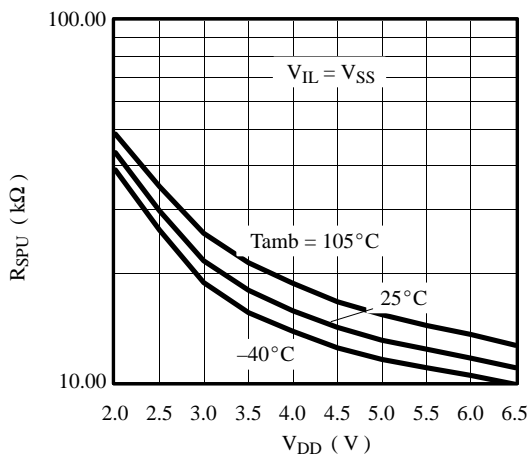


Figure 103. Strong pull-up resistor vs. V_{DD}

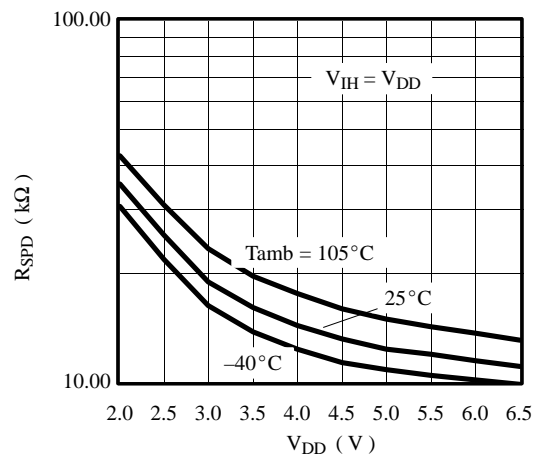


Figure 106. Strong pull-down resistor vs. V_{DD}

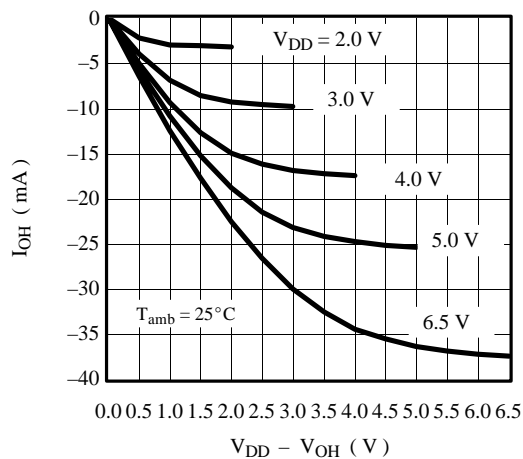


Figure 104. Output high current vs. V_{DD} - output high voltage

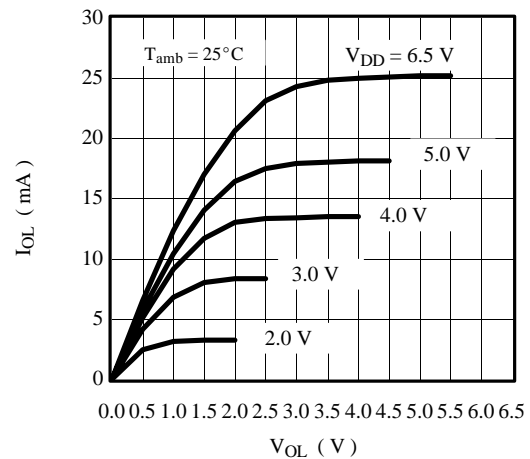


Figure 107. Output low current vs. output low voltage

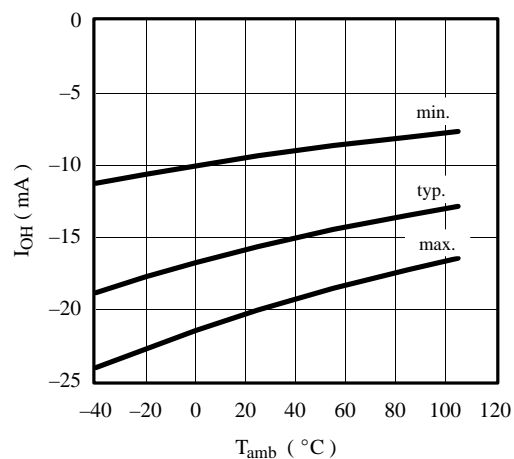


Figure 108. Output high current vs. $V_{DD} = 6.5$ V,
 $V_{OH} = 0.8 \times V_{DD}$

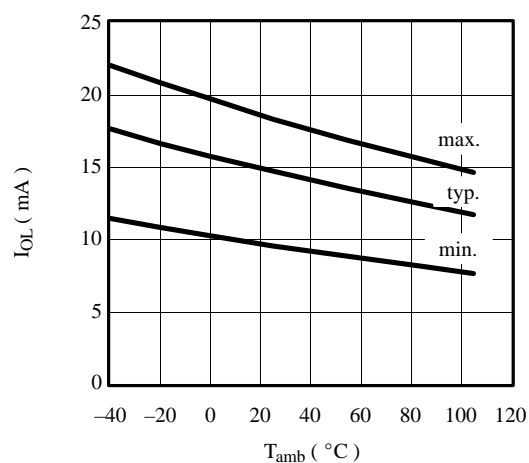
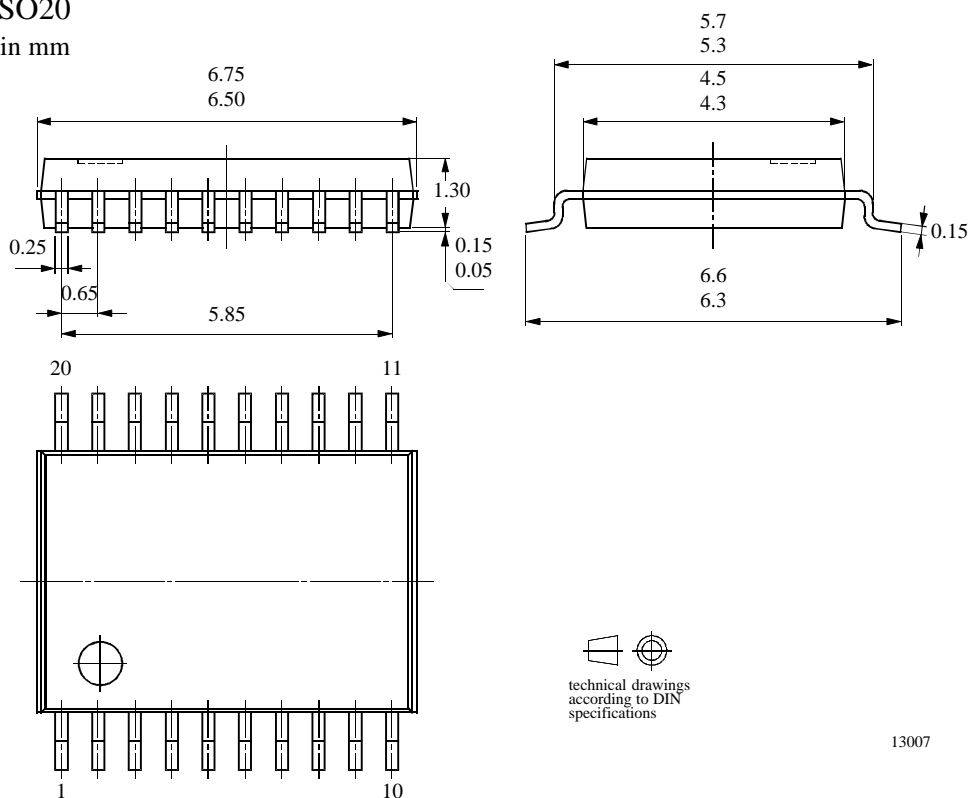


Figure 109. Output low current vs. T_{amb}
 $V_{DD} = 6.5$ V, $V_{OL} = 0.2 \times V_{DD}$

Package Information

Package SSO20

Dimensions in mm



13007

6 Ordering Information

Please select the option setting from the list below and insert ROM CRC.

Port 1

- BP10 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down
- BP13 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down

Port 2

- BP20 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
- BP21 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down
- BP22 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down
- BP23 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down

Port 4

- BP40 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down
- BP41 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down
- BP42 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down
- BP43 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down

Port 5

- BP50 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down
- BP51 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down

- BP52 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down

- BP53 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down

Port 6

- BP60 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down
- BP61 ☐ CMOS ☐ Switched pull-up
☐ Open drain [N] ☐ Switched pull-down
☐ Open drain [P] ☐ Static pull-up
☐ Static pull-down

OSC1

- ☐ No integrated capacitance
☐ Internal capacitance (____ pF)

OSC2

- ☐ No integrated capacitance
☐ Internal capacitance (____ pF)

Clock used

- ☐ External resistor
☐ External clock
☐ 32-kHz crystal
☐ 4-MHz crystal

ECM(External clock monitor)

- ☐ Enable
☐ Disable

File: _____ . HEX

CRC: _____ HEX

Approval

Date: ____-____-____ Signature: _____

Ozone Depleting Substances Policy Statement

It is the policy of **Atmel Germany GmbH** to

1. Meet all present and future national and international statutory requirements.
2. Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

It is particular concern to control or eliminate releases of those substances into the atmosphere which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) intend to severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

Atmel Germany GmbH has been able to use its policy of continuous improvements to eliminate the use of ODSs listed in the following documents.

1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA
3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

Atmel Germany GmbH can certify that our semiconductors are not manufactured with ozone depleting substances and do not contain such substances.

We reserve the right to make changes to improve technical design and may do so without further notice.

Parameters can vary in different applications. All operating parameters must be validated for each customer application by the customer. Should the buyer use Atmel products for any unintended or unauthorized application, the buyer shall indemnify Atmel against all claims, costs, damages, and expenses, arising out of, directly or indirectly, any claim of personal damage, injury or death associated with such unintended or unauthorized use.

Data sheets can also be retrieved from the Internet: <http://www.atmel-wm.com>

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