

LTC1456

12-Bit Rail-to-Rail Micropower DAC with Clear Input

### FEATURES

- SO-8 Package
- 12-Bit Resolution
- Buffered True Rail-to-Rail Voltage Output
- Asynchronous Clear Input
- Built-In Reference
- Schmitt Trigger On Clock Input Allows Direct Optocoupler Interface
- Power-On Reset Clears DAC to 0V
- 3-Wire Cascadable Serial Interface
- Maximum DNL Error: 0.5LSB
- Low Cost

# **APPLICATIONS**

- Digital Calibration
- Industrial Process Control
- Automatic Test Equipment
- Cellular Telephones

# DESCRIPTION

The LTC<sup>®</sup>1456 is a complete single supply, rail-to-rail voltage output, 12-bit digital-to-analog converter (DAC) in an SO-8 package. It includes a rail-to-rail output buffer amplifier and an easy-to-use 3-wire cascadable serial interface. The LTC1456 includes a CLR pin that asynchronously clears the DAC to zero scale.

The LTC1456 has an internal 2.048V reference and a fullscale output of 4.095V. It operates on a 4.5V to 5.5V supply, dissipating 2.2mW.

The low power supply current and the space saving SO-8 package make the LTC1456 ideal for battery-powered applications.

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## TYPICAL APPLICATION

Functional Block Diagram: 12-Bit Rail-to-Rail DAC with Clear Input

Differential Nonlinearity vs Input Code





# **ABSOLUTE MAXIMUM RATINGS**

$V_{CC}$ to GND
Logic Inputs to GND –0.5V to 7.5V
$V_{\text{OUT}}$
Maximum Junction Temperature65°C to 125°C
Operating Temperature Range
LTC1456C 0°C to 70°C
LTC1456I40°C to 85°C
Storage Temperature Range65°C to 150°C
Lead Temperature (Soldering, 10 sec) 300°C

# PACKAGE/ORDER INFORMATION



Consult factory for Military grade parts.

# **ELECTRICAL CHARACTERISTICS**

 $V_{CC}$  = 4.5V to 5.5V,  $V_{OUT}$  unloaded,  $T_A$  =  $T_{MIN}$  to  $T_{MAX},$  unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
DAC							
	Resolution		•	12			Bits
DNL	Differential Nonlinearity	Guaranteed Monotonic (Note 1)	•			±0.5	LSB
INL	Integral Nonlinearity	T <sub>A</sub> = 25°C (Note 1)	•			±3.5 ±4	LSB LSB
V <sub>OS</sub>	Offset Error	$T_A = 25^{\circ}C$	•			±12 ±18	mV mV
V <sub>OS</sub> TC	Offset Error Temperature Coefficient				±15		μV/°C
V <sub>FS</sub>	Full-Scale Voltage	$T_A = 25^{\circ}C$	•	4.065 4.045	4.095 4.095	4.125 4.145	V V
V <sub>FS</sub> TC	Full-Scale Voltage Temperature Coefficient				±24		ppm/°C
Power Su	ipply		I I				
V <sub>CC</sub>	Positive Supply Voltage	For Specified Performance	•	4.5		5.5	V
I <sub>CC</sub>	Supply Current	(Note 4)	•		430	650	μA
Op Amp D	DC Performance						
	Short-Circuit Current Low	V <sub>OUT</sub> Shorted to GND	•			120	mA
	Short-Circuit Current High	V <sub>OUT</sub> Shorted to V <sub>CC</sub>	•			120	mA
	Output Impedance to GND	Input Code = 0	•		40	120	Ω
AC Perfor	rmance						·
	Voltage Output Slew Rate	(Note 2)	•	0.5	1.0		V/µs
	Voltage Output Settling Time	(Notes 2, 3) to ±0.5LSB			14		μs
	Digital Feedthrough				0.3		nV∙s



## **ELECTRICAL CHARACTERISTICS**

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SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
Digital I/O							<u></u>
V <sub>IH</sub>	Digital Input High Voltage		•	2.4			V
V <sub>IL</sub>	Digital Input Low Voltage		•			0.8	V
V <sub>OH</sub>	Digital Output High Voltage	$I_{OUT} = -1mA, D_{OUT} Only$	•	V <sub>CC</sub> - 1.0			V
V <sub>OL</sub>	Digital Output Low Voltage	$I_{OUT} = 1 \text{mA}, D_{OUT} \text{ Only}$	•			0.4	V
I <sub>LEAK</sub>	Digital Input Leakage	$V_{CC} = 5V, V_{IN} = GND$ to $V_{CC}$	•			±10	μA
C <sub>IN</sub>	Digital Input Capacitance	Guaranteed by Design. Not Subject to Test.	•			10	pF
Switching							
t <sub>1</sub>	D <sub>IN</sub> Valid to CLK Setup		•	40			ns
t <sub>2</sub>	D <sub>IN</sub> Valid to CLK Hold		•	0			ns
t <sub>3</sub>	CLK High Time		•	40			ns
t <sub>4</sub>	CLK Low Time		•	40			ns
t <sub>5</sub>	CS/LD Pulse Width		•	50			ns
t <sub>6</sub>	LSB CLK to CS/LD		•	40			ns
t <sub>7</sub>	CS/LD Low to CLK		•	20			ns
t <sub>8</sub>	D <sub>OUT</sub> Output Delay	$C_{LOAD} = 15 pF, V_{CC} = 5V$	•			150	ns
t <sub>9</sub>	CLK Low to CS/LD Low		•	20			ns
t <sub>10</sub>	CLR Pulse Width		•	65			ns

The  ${\ensuremath{\bullet}}$  denotes specifications which apply over the full operating temperature range.

**Note 1:** Nonlinearity is defined from the first code that is greater than or equal to the maximum offset specification to code 4095 (full scale). **Note 2:** Load is  $5k\Omega$  in parallel with 100pF.

Note 3: DAC switched between all 1s and the code corresponding to  $V_{\text{OS}}$  for the part.

Note 4: Digital inputs at 0V or  $V_{CC}$ .



# **TYPICAL PERFORMANCE CHARACTERISTICS**

2.0

1.6



AWA 1.2 0.8 INL ERROR (LSB) 0.4 rγħ, 0 -0.4 MN -0.8 -1.2 -1.6 -2.0 512 1024 1536 2048 2560 3072 3584 4095 0 CODE 1456 G02

Integral Nonlinearity (INL)

Minimum Output Voltage vs Output Sink Current



**Output Swing vs Load Resistance** 



Supply Current vs Logic Input Voltage



**Offset Voltage vs Temperature** 



Minimum Supply Headroom for Full Output Swing vs Load Current



**Output Swing vs Load Resistance** 



Supply Current vs Temperature





### PIN FUNCTIONS

**BLOCK DIAGRAM** 

**CLK (Pin 1):** The Serial Interface Clock. Internal Schmitt trigger on this input allows direct optocoupler interface.

 $\mathbf{D}_{\text{IN}}$  (Pin 2): The Serial Interface Data. Data on the  $\mathsf{D}_{\text{IN}}$  pin is latched into the shift register on the rising edge of the serial clock.

 $\overline{\text{CS}/\text{LD}}$  (Pin 3): The Serial Interface Enable and Load Control. When  $\overline{\text{CS}/\text{LD}}$  is low the CLK signal is enabled, so the data can be clocked in. When  $\overline{\text{CS}/\text{LD}}$  is pulled high, data is loaded from the shift register into the DAC register, updating the DAC output. When  $\overline{\text{CS}/\text{LD}}$  is high the CLK is disabled internally.

**D**<sub>OUT</sub> (**Pin 4**): The Output of the Shift Register Which Becomes Valid on the Rising Edge of the Serial Clock.

GND (Pin 5): Ground.

**CLR** (Pin 6): The Clear Input. When pulled low, this pin asynchronously clears the internal shift and DAC registers to zero scale. Should be tied high for normal operation.

Vout (Pin 7): The Buffered DAC Output.

 $V_{CC}$  (Pin 8): The Positive Supply Input. 4.5V  $\leq$  V\_{CC}  $\leq$  5.5V. Requires a bypass capacitor to ground.

#### CLK 1 8 V<sub>CC</sub> 2.048V ┣ ÷ LD D<sub>IN</sub> 2 12-BIT 12-BIT DAC DAC SHIFT REGISTER 7 VOUT REGISTER CS/LD 3 POWER-ON 6 CLR RESET D<sub>OUT</sub> 4 5 GND 1146 BD

# TIMING DIAGRAM



## DEFINITIONS

**Resolution (n):** Resolution is defined as the number of digital input bits, n. It defines the number of DAC output states  $(2^n)$  that divide the full-scale range. The resolution does not imply linearity.

Full-Scale Voltage ( $V_{FS}$ ): This is the output of the DAC when all bits are set to 1.

**Voltage Offset Error (V**<sub>OS</sub>): The theoretical voltage at the output when the DAC is loaded with all zeros. The output amplifier can have a true negative offset, but because the part is operated from a single supply, the output cannot go below zero. If the offset is negative, the output will remain near OV resulting in the transfer curve shown in Figure 1.

The offset of the part is measured at the code that corresponds to the maximum offset specification:

 $V_{OS} = V_{OUT} - [(Code \bullet V_{FS})/(2^n - 1)]$ 

**Least Significant Bit (LSB):** One LSB is the ideal voltage difference between two successive codes.

$$\label{eq:LSB} \begin{split} LSB &= (V_{FS} - V_{OS})/(2^n - 1) = (V_{FS} - V_{OS})/4095 \\ LSB &= 4.095/4095 = 1 mV \end{split}$$

**Integral Nonlinearity (INL):** End-point INL is the maximum deviation from a straight line passing through the end-points of the DAC transfer curve. Because the part operates from a single supply and the output cannot go below zero, the linearity is measured between full scale and the code corresponding to the maximum offset specification. The INL error at a given input code is calculated as follows:

**Differential Nonlinearity (DNL):** DNL is the difference between the measured change and the ideal 1LSB change between any two adjacent codes. The DNL error between any two codes is calculated as follows:

 $DNL = (\Delta V_{OUT} - LSB)/LSB$ 

 $\Delta V_{OUT}$  = The measured voltage difference between two adjacent codes

**Digital Feedthrough:** The glitch that appears at the analog output caused by AC coupling from the digital inputs when they change state. The area of the glitch is specified in (nV)(sec).



Figure 1. Effect of Negative Offset



### OPERATION

### Serial Interface

The data on the  $D_{IN}$  input is loaded into the shift register on the rising edge of the clock. The MSB is loaded first. The <u>DAC</u> register loads the data from the shift register when <u>CS/LD</u> is pulled high. The CLK is disabled internally when <u>CS/LD</u> is high. Note: CLK must be low before <u>CS/LD</u> is pulled low to avoid an extra internal clock pulse.

When CLR is pulled low it asynchronously resets the shift and DAC registers to all zeros.

The buffered output of the 12-bit shift register is available on the  $D_{OUT}$  pin which swings from GND to  $V_{CC}$ . Multiple LTC1456s may be daisy-chained together by connecting the  $D_{OUT}$  pin to the  $D_{IN}$  pin of the next chip, while the CLK and CS/LD signals remain common to all chips in the daisy chain. The serial data is clocked to all of the chips, then the  $\overline{\text{CS}}$ /LD signal is pulled high to update all of them simultaneously.

### Voltage Output

The LTC1456's rail-to-rail buffered output can source or sink 5mA over the entire operating temperature range while pulling to within 300mV of the positive supply voltage or ground. The output swings to within a few millivolts of either supply rail when unloaded and has an equivalent output resistance of  $40\Omega$  when driving a load to the rails. The output can drive 1000pF without going into oscillation.

### TYPICAL APPLICATION

The circuit below shows a digitally programmable current source from an external voltage source using an external op amp, an LT<sup>®</sup>1077 and an NPN transistor (2N3440). Any digital word from 0 to 4095 is loaded into the LTC1456 and its output correspondingly swings from 0V to 4.095V. In the configuration shown, this voltage will be forced across the resistor R<sub>A</sub>. If R<sub>A</sub> is chosen to be 410 $\Omega$  the output current will range from 0mA at zero scale to 10mA at full scale. The minimum voltage for V<sub>S</sub> is determined by the load resistor R<sub>L</sub> and Q1's V<sub>CESAT</sub> voltage. With a load resistor of 50 $\Omega$ , the voltage source can be as low as 5V.











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#### PACKAGE DESCRIPTION Dimensions in inches (millimeters) unless otherwise noted.





## **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS
LTC1257	Single 12-Bit $V_{OUT}$ DAC, Full Scale: 2.048V, $V_{CC}$ : 4.75V to 15.75V, Reference Can Be Overdriven Up to 12V, i.e., FS <sub>MAX</sub> = 12V	5V to 15V Single Supply, Complete V <sub>OUT</sub> DAC in SO-8 Package
LTC1446/LTC1446L	Dual 12-Bit V <sub>OUT</sub> DACs in SO-8 Package	LTC1446: $V_{CC}$ = 4.5V to 5.5V, $V_{OUT}$ = 0V to 4.095V LTC1446L: $V_{CC}$ = 2.7V to 5.5V, $V_{OUT}$ = 0V to 2.5V
LTC1450/LTC1450L	Single 12-Bit V <sub>OUT</sub> DACs with Parallel Interface	LTC1450: $V_{CC}$ = 4.5V to 5.5V, $V_{OUT}$ = 0V to 4.095V LTC1450L: $V_{CC}$ = 2.7V to 5.5V, $V_{OUT}$ = 0V to 2.5V
LTC1451	Single Rail-to-Rail 12-Bit DAC, Full Scale: 4.095V, V <sub>CC</sub> : 4.5V to 5.5V, Internal 2.048V Reference Brought Out to Pin	Same as LTC1456 Except REF Out Pin Replaces CLR Pin
LTC1452	Single Rail-to-Rail 12-Bit $V_{\mbox{OUT}}$ Multiplying DAC, $V_{\mbox{CC}}$ 2.7V to 5.5V	Low Power, Multiplying V <sub>OUT</sub> DAC with Rail-to-Rail Buffer Amplifier in SO-8 Package
LTC1453	Single Rail-to-Rail 12-Bit $V_{OUT}$ DAC, Full Scale: 2.5V, $V_{CC}$ : 2.7V to 5.5V	3V, Low Power, Complete V <sub>OUT</sub> DAC in SO-8 Package
LTC1454/LTC1454L	Dual 12-Bit $V_{OUT}$ DACs in SO-16 Package with Added Functionality	LTC1454: $V_{CC}$ = 4.5V to 5.5V, $V_{OUT}$ = 0V to 4.095V LTC1454L: $V_{CC}$ = 2.7V to 5.5V, $V_{OUT}$ = 0V to 2.5V
LTC1458/LTC1458L	Quad 12 Bit Rail-to-Rail Output DACs with Added Functionality	LTC1458: $V_{CC}$ = 4.5V to 5.5V, $V_{OUT}$ = 0V to 4.095V LTC1458L: $V_{CC}$ = 2.7V to 5.5V, $V_{OUT}$ = 0V to 2.5V

