

CM8870

Microcircuits

CMOS DTMF Integrated Receivers

FEATURES

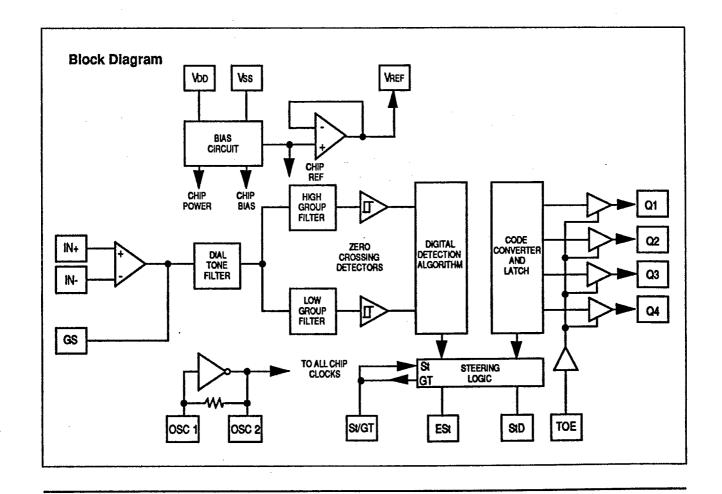
- · Full DTMF receiver
- · Less than 35 mW power consumption
- · Industrial temperature range
- · Uses quartz crystal or ceramic resonators
- · Adjustable acquisition and release times
- 18-pin DIP, 18-pin SOIC, 20 pin PLCC

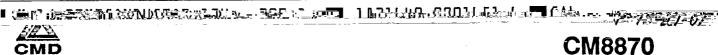
APPLICATIONS

- PABX
- Remote data entry
- Central office
- Call limiting
- Mobile radio
- Telephone answering systems
- Remote control
- Paging systems

GENERAL DESCRIPTION

The CMD CM8870 provides full DTMF receiver capability by integrating both the bandsplit filter and digital decoder functions into a single 18-pin DIP, SOIC, or 20-pin PLCC package. The CM8870 is manufactured using state-of-the-art CMOS process technology for low power consumption (35mW max.) and precise data handling. The filter section uses a switched capacitor technique for both high and low group filters and dial tone rejection. The CM8870 decoder uses digital counting techniques for the detection and decoding of all 16DTMF tone pairs into a 4-bit code. The CM8870 minimizes external component count by providing an on-chip differential input amplifier, clock generator, and a latched three-state interface bus. The on-chip clock generator requires only a low cost TV crystal or ceramic resonator as an external component.





Absolute Maximum Ratings: (Note 1)

Parameter	Symbol	Value		
Power Supply Voltage (VDD-Vss)	VDD	6.0V Max		
Voltage on any Pin	Vdc	Vss-0.3, VDD+0.3		
Current on any Pin	IDD	10 mA Max		
Operating Temperature	TA	-40°C to +85°C		
Storage Temperature	TS	-65°C to +150°C		

This device contains input protection against damage due to high static voltages or electric fields; however, precautions should be taken to avoid application of voltages higher than the maximum rating.

Notes:

1. Exceeding these ratings may cause permanent damage, functional operation under these conditions is not implied.

DC Characteristics: All voltages referenced to Vss, VDD = 5.0V ±5%, TA = -40°C to +85°C unless otherwise noted.

Parameter		Symbol	Min	Тур	Max	Units	Test Conditions
Operating Supply Voltage		VDD	4.75		5.25	V	
Operating Supply Cur	rent	aal		3.0	7.0	mA	
Power Consumption		Po		15	35	mW	f=3.579 MHz; VDD=5.0V
Low Level Input Voltage		VIL			1.5	٧	VDD=5.0V
High Level Input Voltage		VIH	3.5			V	V00≈5.0V
Input Leakage Current		IIH/IIL		0.1		μА	Vin=Vss or VDD (Note 11)
Pull Up (Source) Current on TOE		Iso		6.5	20	μА	TOE=0V, VDD=5.0V
Input Impedance, (IN+, IN-)		RIN	8	10		MΩ	@ 1KHz
Steering Threshold Voltage		VTst	2.2		2.5	V	VDD=5.0V
Low Level Output Vol	tage	VOL			0.03	V	VDD=5.0V, No Load
High Level Output Vo	ltage	Vон	4.97			V	VDD=5.0V, No Load
Output Low (Sink) Current		IOL	1.0	2.5		mA	Vouт=0.4V
Output High (Source) Current		Юн	0.4	0.8		mA	Vour=4.6V
Output Voltage VREF		VREF	2.4		2.7	V	VDD=5.0V, No Load
Output Resistance	AUEL	Ron		10		ΚΩ	

Operating Characteristics: All voltages referenced to Vss, VDD = 5.0V ±5%, TA = -40°C to +85°C unless otherwise noted. **Gain Setting Amplifier**

Parameter	Symbol	Min	Тур	Max	Units	Test Conditions	
Input Leakage Current	lin			±100	nA	Vss < Vin < VDD	
Input Resistance	Rin	10			MΩ		
Input Offset Voltage	Vos			±25	mV		
Power Supply Rejection	PSRR	50			dB	1KHz (Note 12)	
Common Mode Rejection	CMRR	40			dB	-3.0V < VIN < 3.0V	
DC Open Loop Voltage Gain	AVOL	32			dB		
Open Loop Unity Gain Bandwidth	fc	0.3			MHz		
Output Voltage Swing	Vo	4.0			Vp-p	RL ≥ 100KΩ to Vss	
Tolerable Capacitive Load (GS)	CL			100	ρF		
Tolerable Resistive Load (GS)	RL			50	ΚΩ		
Common Mode Range	Vcm	2.5			Vp-p	No Load	





AC Characteristics: All voltages referenced to Vss, VDD = $5.0V \pm 5\%$, TA = -40°C to +85°C, fcLK = 3.579545 MHz using test circuit (Fig. 1) unless otherwise noted.

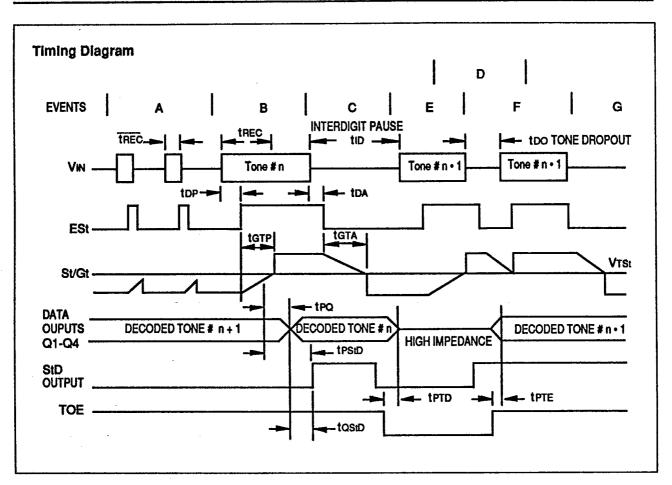
Parameter		Symbol	Min	Тур	Max	Units	Notes	
Valid Input Signal Le		-29		+1	dBm	1,2,3,4,5,8		
(each tone of compos	site signal)		27.5		869	mVRMS		
Positive Twist Accept	<u> </u>				10	dB	2,3,4,8	
Negative Twist Accep	ot				10	dB	د,ی,4,6 	
Freq. Deviation Acce	pt Limit				1.5%±2 Hz	Nom.	2,3,5,8,10	
Freq. Deviation Reje	ct Limit		±3.5%			Nom.	2,3,5	
Third Tone Tolerance)			-16		dB	2,3,4,5,8,9,13,14	
Noise Tolerance				-12		dB	2,3,4,5,6,8,9	
Dial Tone Tolerance				+22		dB	2,3,4,5,7,8,9	
Tone Present Detection Time		tDP	5	8	14	mS	Refer to	
Tone Absent Detection Time		tDA	0.5	3	8.5	mS	Timing Diagram	
Min. Tone Duration A	Min. Tone Duration Accept				40	mS	(User Adjustable)	
Max. Tone Duration	Reject	tREC	20			mS	Times shown are	
Min. Interdigit Pause	Accept	tiD			40	mS	obtained with	
Max. Interdigit Pause	Reject	tDO	20			mS	circuit in Fig. 1	
Propagation Delay (S	St to Q)	tPQ		6	11	μS		
Propagation Delay (S	St to StD)	tPStD		9	16	μS	TOE = VDD	
Output Data Set Up (Q to StD)		tosto		3.4		μS		
Propagation Delay	Enable	tPTE		50		nS	RL = 10KΩ	
(TOE to Q) Disable		tPTD		300		nS	CL = 50pF	
Crystal/Clock Frequency		fclk	3.5759	3.5795	3.5831	MHz		
Clock Output	Capacitive	CLO	·		30	pF		
(OSC 2)	Load							

Notes:

- dBm = decibels above or below a reference power of 1 mW into a 600 ohm load.
- 2. Digit sequence consists of all 16 DTMF tones.
- 3. Tone duration = 40 mS. Tone pause = 40 mS.
- 4. Nominal DTMF frequencies are used.
- Both tones in the composite signal have an equal amplitude.
- 6. Bandwidth limited (0 to 3 KHz) Gaussian Noise.
- 7. The precise dial tone frequencies are (350 Hz and 440 Hz) ±2%.

- 8. For an error rate of better than 1 in 10,000.
- 9. Referenced to lowest level frequency component in DTMF signal.
- 10. Minimum signal acceptance level is measured with specified maximum frequency deviation.
- 11. Input pins defined as IN+, IN-, and TOE.
- 12. External voltage source used to bias VREF.
- 13. This parameter also applies to a third tone injected onto the power supply.
- 14. Referenced to Figure 1. Input DTMF tone level at -28 dBm.





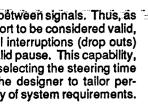
Explanation of Events

- Tone bursts detected, tone duration invalid, outputs not updated.
- Tone #n detected, tone duration valid, tone decoded and latched in outputs.
- C) End of tone #n detected, tone absent duration valid, outputs remain latched until next valid tone.
- D) Outputs switched to high impedance state.
- E) Tone #n + 1 detected, tone duration valid, tone decoded and latched in outputs (currently high impedance).
- F) Acceptable dropout of tone #n + 1, tone absent duration invalid, outputs remain latched.
- G) End of tone #n + 1 detected, tone absent duration valid, outputs remain latched until next valid tone.

Explanation of Symbols

- VIN DTMF composite input signal.
- ESt Early Steering Output. Indicates detection of valid tone frequencies.
- St/GT Steering input/guard time output. Drives external RC timing circuit.
- Q1-Q4 4-bit decoded tone output.
- StD Delayed Steering Output. Indicates that valid frequencies have been present/absent for the required guard time, thus constituting a valid signal.
- TOE Tone Output Enable (input). A low level shifts Q1-Q4 to its high impedance state.
- tREC Maximum DTMF signal duration not detected as valid.
- tree Minimum DTMF signal duration required for valid recognition.
- tip Minimum time between valid DTMF signals.
- tpo Maximum allowable drop-out during valid DTMF signal.
- top Time to detect the presence of valid DTMF signals.
- tDA Time to detect the absence of valid DTMF signals.
- tGTP Guard time, tone present.
- tGTA Guard time, tone absent.

CM8870



ire independent selection e steering circuit of Figure ues are chosen according

+ **t**GTP

7 RC

r of the device and tREC is to be recognized by the is recommended for most selected by the designer. e of R for a tREC of 40 A typical circuit using this n in Figure 1. The timing munication applications Different steering arrange-independently the guardat one absent (tGTA). This stem specifications which mits on both tone duratior

lows the designer to tailor lk-off and noise immunity. -off performance, since it nes simulated by speech for long enough to be a relatively short tREC with riate for extremely noisy sition time and immunity to its. Design information for in Figure 5.

CM8870 provides a differer as well as a bias source re inputs at mid-rail.

n of a feedback resistor to justment of gain.

n, the input pins are conith the op-amp connected the input at 1/2 VDD. Figure gration, which permits the edback resistor R5.

pleted with the addition of st crystal or ceramic resoency of 3.579545 MHz. A n be connected as shown a crystal or resonator re-

Functional Description

The CMD CM8870 DTMF Integrated Receiver provides the design engineer with not only low power consumption, but high performance in a small 18-pin DIP, SOIC or 20-pin PLCC package configuration. The CM8870's internal architecture consists of a bandsplit filter section which separates the high and low tones of the received pair, followed by a digital decode (counting) section which verifies both the frequency and duration of the received tones before passing the resultant 4-bit code to the output bus.

Filter Section

Separation of the low-group and high-group tones is achieved by applying the dual-tone signal to the inputs of two 9th-order switched capacitor bandpass filters. The bandwidths of these filters correspond to the bands enclosing the low-group and high-group tones (See Figure 3). The filter section also incorporates notches at 350 Hz and 440 Hz which provides excellent dial tone rejection. Each filter output is followed by a single-order switched capacitor section which smooths the signals prior to limiting. Signal limiting is performed by high-gain comparators. These comparators are provided with a hysteresis to prevent detection of unwanted low-level signals and noise. The outputs of the comparators provide full-rail logic swings at the frequencies of the incoming tones.

Decoder Section

The CM8870 decoder uses a digital counting technique to determine the frequencies of the limited tones and to verify that these tones correspond to standard DTMF frequencies. A complex averaging algorithm is used to protect against tone simulation by extraneous signals (such as voice) while providing tolerance to small frequency variations. The averaging algorithm has been developed to ensure an optimum combination of immunity to "talk-off" and tolerance to the presence of interfering signals (third tones) and noise. When the detector recognizes the simultaneous presence of two valid tones (known as "signal condition"), it raises the "Early Steering" flag (ESt). Any subsequent loss of signal condition will cause ESt to fall.

Steering Circuit

Before the registration of a decoded tone pair, the receiver checks for a valid signal duration (referred to as "character-recognition-condition"). This check is performed by an external RC time constant driven by ESt. A logic high on ESt causes Vc (See Figure 4) to rise as the capacitor discharges. Providing signal condition is maintained (ESt remains high) for the validation period (tGTP), VC reaches the threshold (VTst) of the steering logic to register the tone pair, thus latching its corresponding 4-bit code (See Figure 2) into the output latch. At this point, the GT output is activated and drives Vc to VDD. GT continues to drive high as long as ESt remains high. Finally, after a short delay to allow the output latch to settle, the "delayed steering" output flag (StD) goes high, signaling that a received tone pair has been registered. The contents of the output latch are made available on the 4bit output bus by raising the three-state control input (TOE) to a logic high. The steering circuit works in reverse

to validate the interdigit pause I well as rejecting signals too sh the receiver will tolerate signal too short to be considered a vatogether with the capability of constants externally, allows the tolerance to meet a wide variet

Guard Time Adjustment

In situations which do not requ of receive and pause, the simpl 4 is applicable. Component val to the following formula:

trec = top

 $t_{GTP} \approx 0.6$

The value of tDP is a paramete the minimum signal duration receiver. A value for C of 0.1 upon applications, leaving R to be For example, a suitable valuable milliseconds would be 300K, steering configuration is show requirements for most teleconare satisfied with this circuit. If ments may be used to select times for tone-present (tGTP) are may be necessary to meet syplace both accept and reject linear interdigit pause.

Guard time adjustment also all system parameters such as ta Increasing tREC improves talk reduces the probability that to will maintain signal condition registered. On the other hand, a long tDO would be appropenvironments where fast acquidrop-outs would be requirement guard time adjustment is show

input Configuration

The input arrangement of the ential input operational amplific (VREF) which is used to bias the

Provision is made for connection the op-amp output (GS) for ad

In a single-ended configuration nected as shown in Figure 1 w for unity gain and VREF biasing 6 shows the differential configuration adjustment of gain with the feet

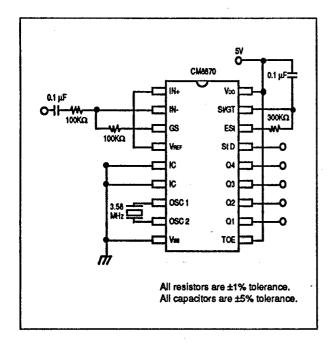
Clock Circuit

The internal clock circuit is com a standard television color bur nator having a resonant freque number of CM8870 devices ca in Figure 7 such that only on quired.



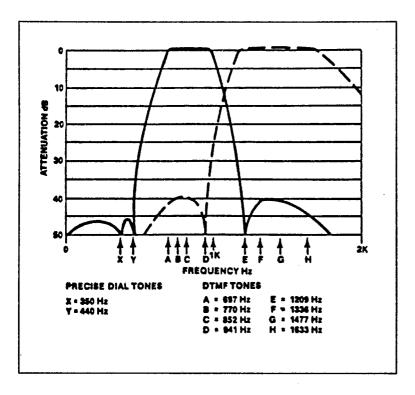
Pin Function Table

Name	Description							
IN+ IN-	Non-inverting input Connections to the front-end differential amplifier							
GS	Gain Select. Gives access to output of front-end differential amplifier for connection of feedback resistor.							
VREF	Reference voltage output (nominally VDD/2). May be used to bias the inputs at mid-rail.							
OSC 1	Clock input							
OSC 2	Clock output	3.579545 MHz crystal connected between these pins completes internal oscillator.						
Vss	Negative power supply (Normally connected to 0V).							
TOE	Three-state output enable (input). Logic high enables the outputs Q1-Q4. Internal pull-up.							
Qı								
Q2	Three-state outputs. When enabled by TOE, provides the code corresponding to the last valid tone pair							
Qз	received. (See Fig. 2.)							
Q4								
StD	Delayed steering output. Presents a logic high when a received tone pair has been registered and the output latch is updated. Returns to logic low when the voltage on St/GT falls below VTst.							
ESt	Early steering output. Presents a logic high immediately when the digital algorithm detects a recognizable tone pair (signal condition). Any momentary loss of signal condition will cause ESt to return to a logic low.							
St/GT	Steering input/guard time output (bidirectional). A voltage greater than VTsi detected at St causes the device to register the detected tone pair and update the output latch. A voltage less than VTsi frees the device to accept a new tone pair. The GT output acts to reset the external steering time constant, and its state is a function of ESt and the voltage on St. (See Fig. 2.)							
VDD	Positive power supp	ly						
Ю	Internal connection. Must be tied to Vss.							



FLOW	FHIGH	KEY	TOE	Q4	Q3	Q2	Q ₁	
697	1209	1	Н	0	0	0	1	
697	1336	2	H	0	0	1	0	
697	1477	3	H	0	0	1	1	
770	1209	4	H	0	1	0	0	
770	1336	5	Η	0	1	0	1	
770	1477	6	Ι	0	1	1	0	
852	1209	7	H	0	1	1	1	
852	1336	8	Н	1	0	0	0	
852	1477	9	Н	1	0	0	1	
941	1336	0	Н	1	0	1	0	
941	1209	•	Н	1	0	1	1	
941	1477	#	Н	1	1	0	0	
697	1633	Α	Н	1	1	0	1	
770	1633	В	Н	1	1	1	0	
852	1633	С	Н	1	1	1	1	
941	1633	D	Н	0	0	0	0	
	-	ANY	L	Z	Z	Z	Z	
L=	L = Logic Low, H = Logic High, Z = High Impedance							

Figure 2. Functional Decode Table



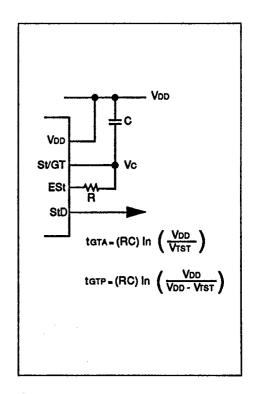


Figure 3. Typical Filter Characteristic

Figure 4. Basic Steering Circuit

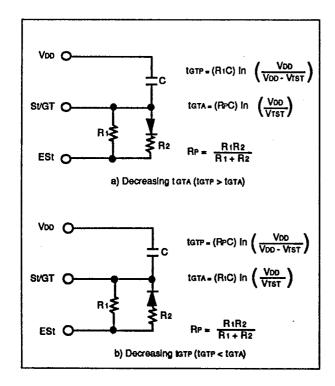


Figure 5. Guard Time Adjustment

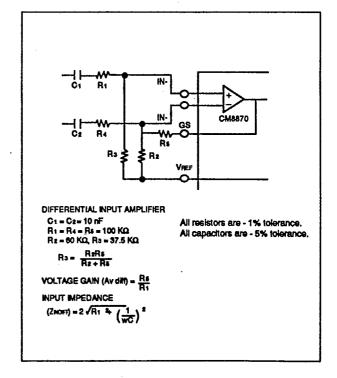


Figure 6. Differential Input Configuration

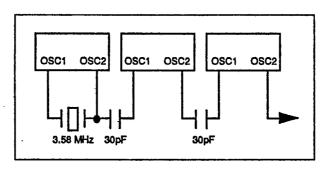


Figure 7. CM8870C Crystal Connection

