

# FAN8902(KA3902)

## DC FAN Motor Controller

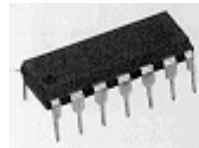
### Features

- Built-in PWM Current Control Circuit
- Built-in 5V Regulator
- Low Supply Current
- Stalled Motor Current Limitation
- Built-in Over Voltage Protection (OVP)
- Built-in Over Current Protection (OCP)
- Built-in Load Dump Protection
- Built-in Thermal Shutdown (TSD) Circuit
- Built-in Under Voltage Lockout (UVLO) Circuit

### Description

The FAN8902 is a monolithic integrated circuit, designed for the PWM control of a DC fan motor current in an automotive systems. It allows the fan motor speed to be controlled linearly and efficiently.

14-DIP-300



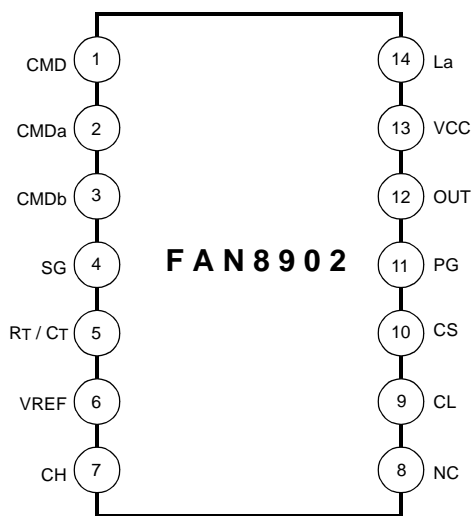
### Typical Application

- DC for Motor Control for Automotive

### Ordering Information

Device	Package	Operating Temperature
FAN8902	14-DIP-300	-40°C ~ +90°C

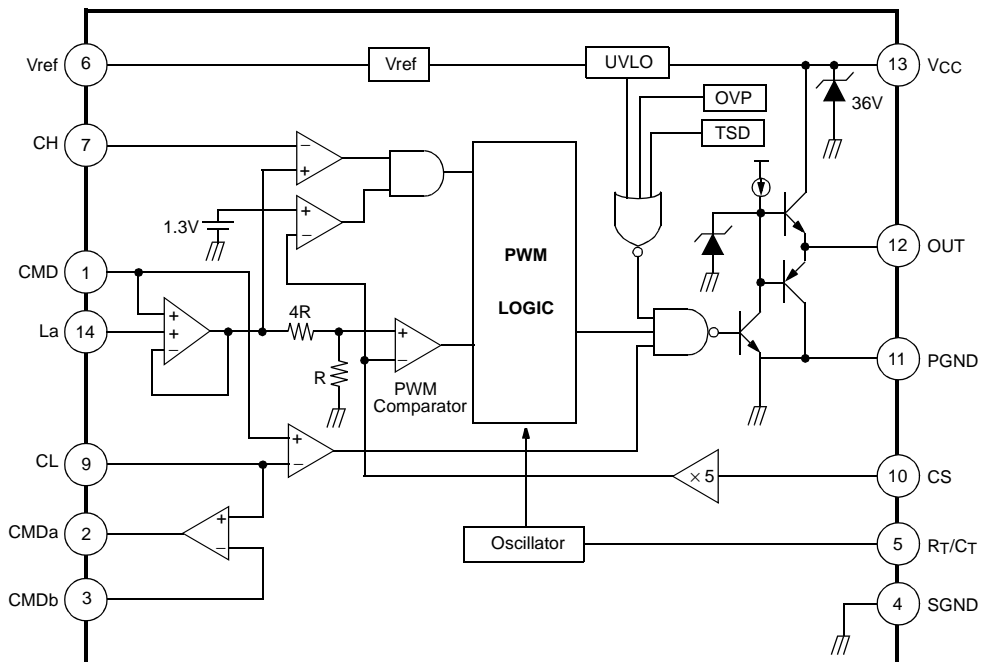
## Pin Assignments



## Pin Definitions

Pin Number	Pin Name	Pin Function Description
1	CMD	Motor Current Command Input
2	CMDa	Optional OP Amplifier Output
3	CMDb	Optional OP Amplifier (-) Input
4	SGND	Signal GND
5	R <sub>T</sub> / C <sub>T</sub>	Oscillator Time Constant
6	VREF	Voltage Reference (5V)
7	CH	Maximum Current Reference Input
8	NC	No Connection
9	CL	Minimum Current Reference Input
10	CS	Motor Current Sense Voltage Input
11	PGND	Power GND
12	OUT	Drive Output
13	VCC	VCC
14	La	Motor Current Maximum Reference Input

# Internal Block Diagram



## Absolute Maximum Ratings

Parameter	Symbol	Value	Unit
Supply Voltage	V <sub>CC</sub>	32	V
CMD Input Voltage	V <sub>CMD</sub>	6	V
Peak Output Current	I <sub>OPK</sub>	±0.8	A
Power Dissipation	P <sub>D</sub>	1	W

## Operating Voltage

Parameter	Symbol	Min.	Typ.	Max	Unit
Power Supply Voltage	V <sub>CC</sub>	9.0	12.0	32.0	V

## Temperature Characteristics

Parameter	Symbol	Temp	Value	Unit
V <sub>ref</sub> Temperature Stability	V <sub>ST</sub>	-40 ~ +90°C	200	°C
Frequency Stability	F <sub>ST</sub>	-40 ~ +90°C	20 ~ 30	°C
Operating Temperature	T <sub>OPR</sub>	-	-40 ~ +90	°C
Storage Temperature	T <sub>STG</sub>	-	-60 ~ +150	°C

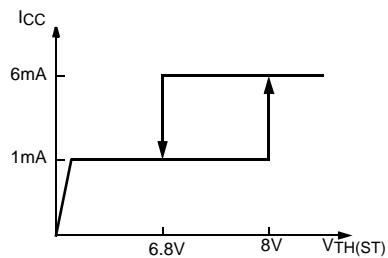
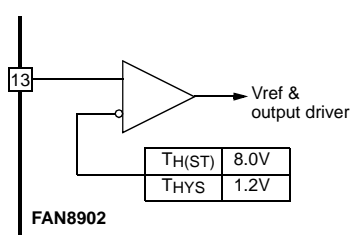
## Electrical Characteristics

(Unless otherwise,  $T_a=25^{\circ}\text{C}$ ,  $V_{CC}=5\text{V}$ ,  $V_M=12\text{V}$ )

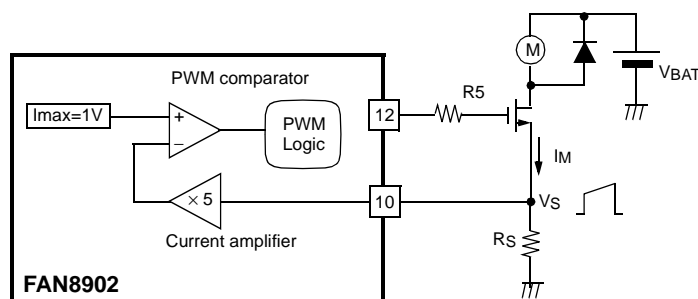
Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
<b>REFERENCE</b>						
Reference Voltage	Vref	Iref=1mA	4.75	5.0	5.25	V
Line Regulation	$\Delta V_{ref1}$	$V_{CC}=9\text{V} \sim 32\text{V}$	-	50	150	mV
Load Regulation	$\Delta V_{ref2}$	Iref=1mA ~ 10mA	-	10	50	mV
<b>UNDER VOLTAGE LOCKOUT (UVLO)</b>						
Start Threshold Voltage	$V_{TH(ST)}$	-	7.5	8.0	8.5	V
Threshold Hysteresis	$V_{HYS}$	-	1.0	1.2	1.4	V
<b>PROTECTION</b>						
Over Voltage	OVP	-	33	36	-	V
<b>OSCILLATOR (<math>R_T=75\text{k}\Omega</math>, <math>C_T=1\text{nF}</math>)</b>						
Frequency	fosc	-	20	25	30	kHz
Duty Cycle	Duty	-	90	95	-	%
<b>CURRENT SENSING INPUT</b>						
Threshold Voltage	$V_{TH(ST)}$	$V_{CMD} = 5\text{V}$	0.19	0.20	0.21	V
<b>OUTPUT DRIVER</b>						
Output Voltage Switching Limit	$V_{OLIM}$	$V_{CC} = 18\text{V}$ , $C_{ld} = 1\text{nF}$	-	15	-	V
Low Output Voltage	$V_{OL1}$	Iout = 20mA	-	-	0.4	V
	$V_{OL2}$	Iout = 200mA	-	-	2.2	V
High Output Voltage	$V_{OH1}$	Iout = -20mA	10.0	-	-	V
	$V_{OH2}$	Iout = -200mA	9.0	-	-	V
Rising Time	Tr	$C_{ld} = 1\text{nF}$	-	100	200	ns
Falling Time	Tf	$C_{ld} = 1\text{nF}$	-	100	200	ns
<b>TOTAL STANDBY CURRENT</b>						
Start-up Current	IST	$V_{CC} = 7\text{V}$	-	1.0	1.5	mA
Operating Supply Current	ICC	$V_{CC} = 9\text{V}$	-	6.0	8.0	mA

## Application Information

### 1. Under Voltage Lockout (UVLO)



### 2. Current Sensing Circuit



The peak current,  $I_{M(MAX)} = V_S / R_S$

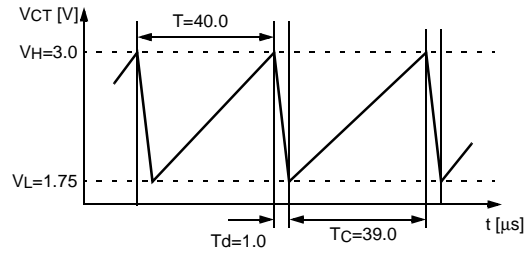
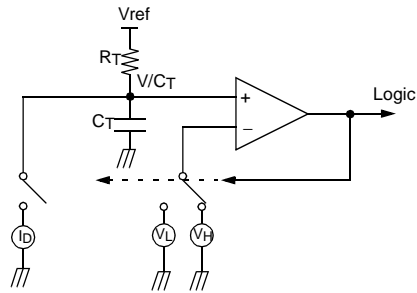
For example, if a required maximum current,  $I_{M(MAX)} = 20[A]$

$$R_S = \frac{1V/5}{20A} = 10[m\Omega]$$

### 3. Thermal Shutdown (Tsd)

When the chip, temperature rises up to 150°C, the thermal shutdown (TSD) circuit is activated and the output driver turn off, and then turn on again at 125°C.

### 4. Oscillator Component Selection



The oscillator timing components can be calculated as follows:

$$T_C = R_T \times C_T \times \ln[(V_{ref} - V_L)/(V_{ref} - V_H)]$$

$$T_D = C_T \times [(V_H - V_L)/I_D]$$

$$f_{osc} = 1/(T_C + T_D)$$

$$= 1.875/(R_T \times C_T)$$

$$Duty = T_C \times f_{osc} \times 100$$

For example, if  $f_{osc} = 25\text{kHz}$  and  $duty = 95\%$

$$C_T = (T_D \times I_D)/(V_H - V_L)$$

$$= 1000[\text{pF}]$$

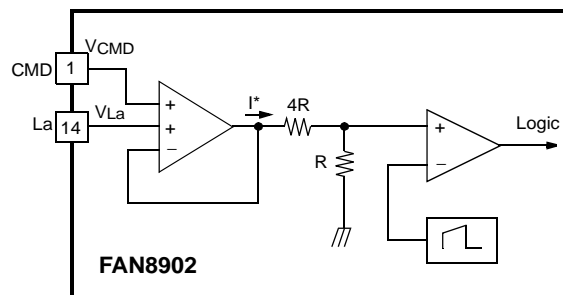
$$R_T = 1.875/(f_{osc} \times C_T)$$

$$= 1.875/(25\text{kHz} \times 1000\text{pF})$$

$$= 75[\text{k}\Omega]$$

### 5. Current Command Input Section

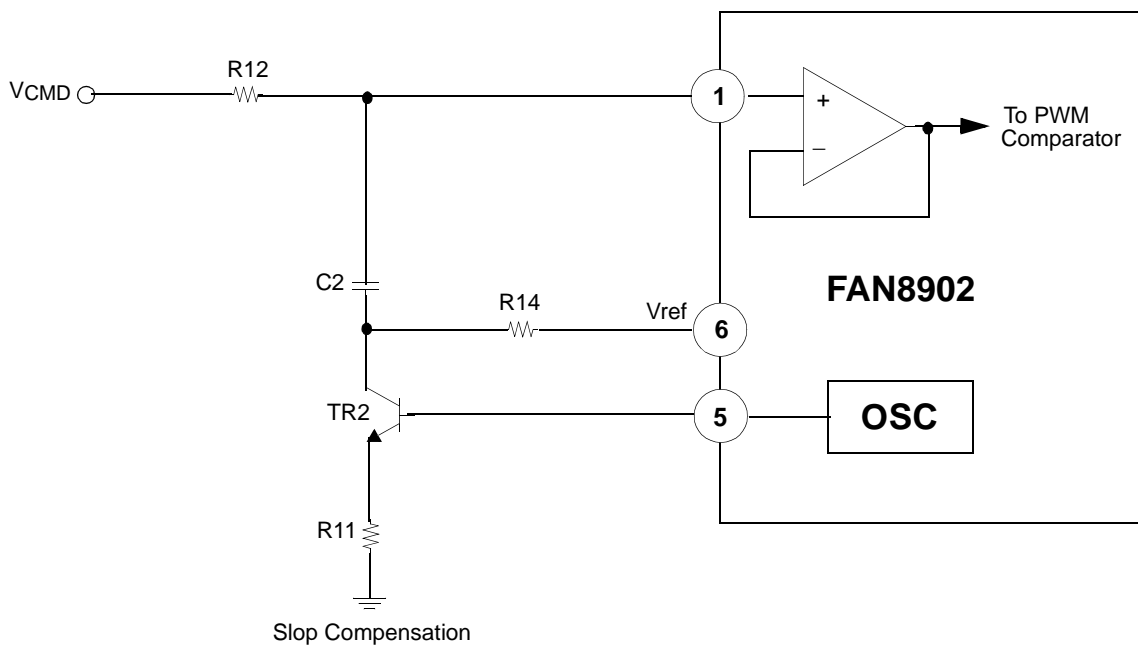
The current command  $I^*$  selects the lower value between  $V_{CMD}$  and  $V_{La}$ .



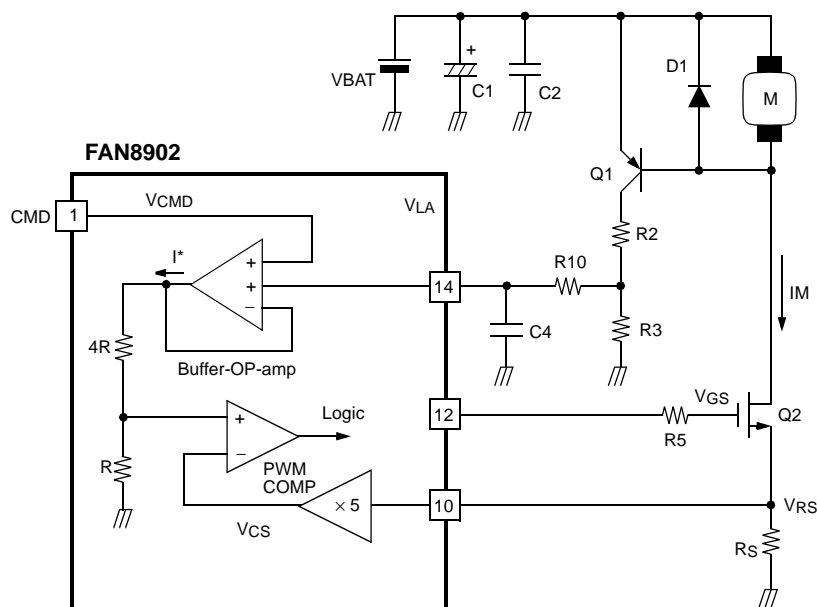
## 6. Slope Compensation

An unconditional instability of the inner current loop exists for any fixed frequency current-mode converter operating above 50% duty cycle. Therefore, to guarantee current loop stability, the slope of the compensation ramp must be greater than one-half of the down slope of the current waveform. The ramp voltage for slope compensation is as follow,

$$V_{RAMP} = \frac{R14}{R11} \times \Delta V_{OSC}$$



## 7. Motor Stall Current Limitation





In the steady state, the terminal voltage on a motor is consisted of a back EMF and the voltage drop on the armature resistors. When the motor happens to be stalled, the back EMF becomes zero, and the motor current ( $I_M$ ) is quickly increased until a maximum values.

Therefore the duty of the pin #12 output becomes lower because of the increase of the sense voltage ( $V_{RS}$ ). Also it makes the voltage ( $V_{La}$ ) be lowered, then it makes the duty become lower again.

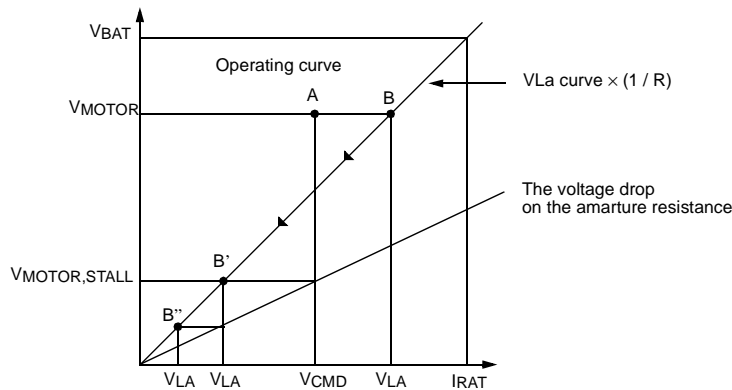
This mechanism makes the motor current hold very low value in the stalled motor state.

The voltage on pin #14 ( $V_{La}$ ) is calculated as follows:

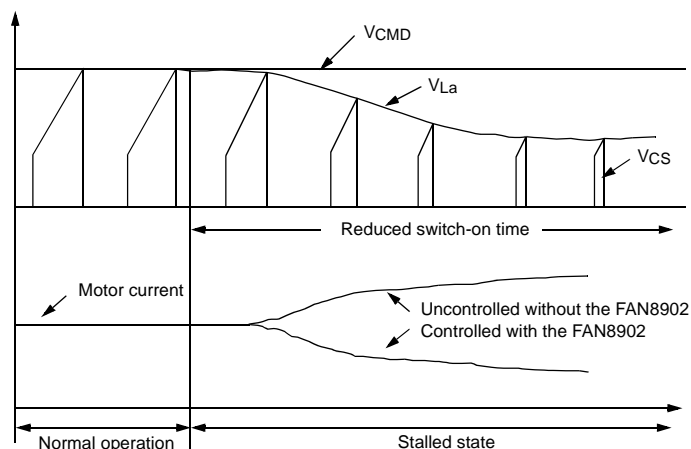
$$V_{La} = V_{BAT} \times D \times \frac{R3}{R2 + R3} \quad \text{Assumed the saturation voltage of Q1 is zero.}$$

We can choose the ratio of the resistors,  $R2$  and  $R3$ , as follows:

- Applied the rated voltage on motor, and then measured the current  $I_{RAT}$
- Matched the maximum command current,  $V_{CMD,MAX}$  to  $I_{RAT}$ .  
 $V_{CMD,MAX} = V_{La,MAX} = R_S \times I_{RAT} \times 5 \times 5$   
 for example, if  $R_S = 10m\Omega$  and  $I_{RAT} = 20[A]$  at  $V_{BAT} = 13[V]$ ,  
 $V_{CMD,MAX} = V_{La,MAX} = 10m\Omega \times 20 \times 25 = 5V$
- $V_{La,MAX} = 5V = V_{BAT} \times 1 \times R3 / (R2 + R3)$   
 $Ratio = R3 / (R2 + R3) = V_{CMD,MAX} / V_{BAT} = 5 / 13$   
 Therefore,  $R2 : R3 = 8 : 5$



The buffer OP-amp selects the lower command between  $V_{CMD}$  and  $V_{La}$  so as to limit the stalled motor current to very low in the above figure. Because of much larger  $V_{La}$  than  $V_{CMD}$ , the motor operating point stays at A. But the point gradually moves toward B' and then B'' through the curve from the instance of stall as the below figure.



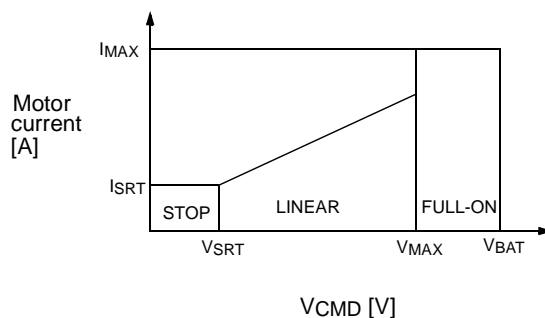
## 8. Operational Mode Selection

The FAN8902 has three operation modes as follows:

- STOP: Turned-off the power MOSFET
- LINEAR: Linearly controlled the power MOSFET
- FULL-ON: Fully turned-on the power MOSFET

The voltage,  $V_{SRT}$  (PIN #9) and  $V_{MAX}$  (PIN #7), in the application circuit are as follows:

- $V_{SRT}$  (PIN #9) =  $V_{ref} \times R7 / (R5 + R6 + R7)$
- $V_{MAX}$  (PIN #7) =  $V_{ref} \times (R6 + R7) / (R5 + R6 + R7)$



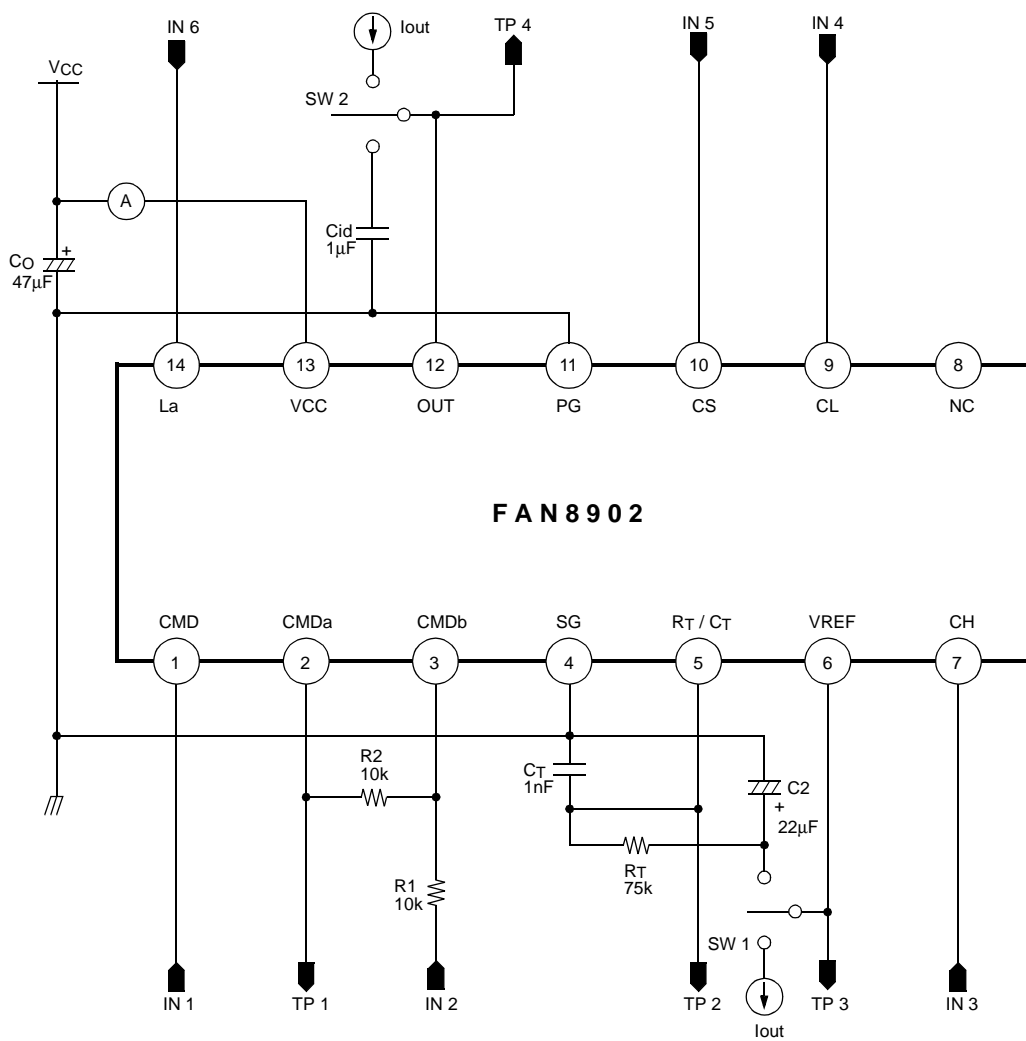
## 9. Over Voltage Protector (Ovp)

If the voltage,  $V_{BAT} \geq 36[V]$ , the output (pin #12) is grounded, and the switching device (power MOSFET) is turned-off, and the motor is stopped. Then if the voltage,  $V_{BAT} \rightarrow 36[V]$ , the switching device is turned-on again, and the motor is operated.

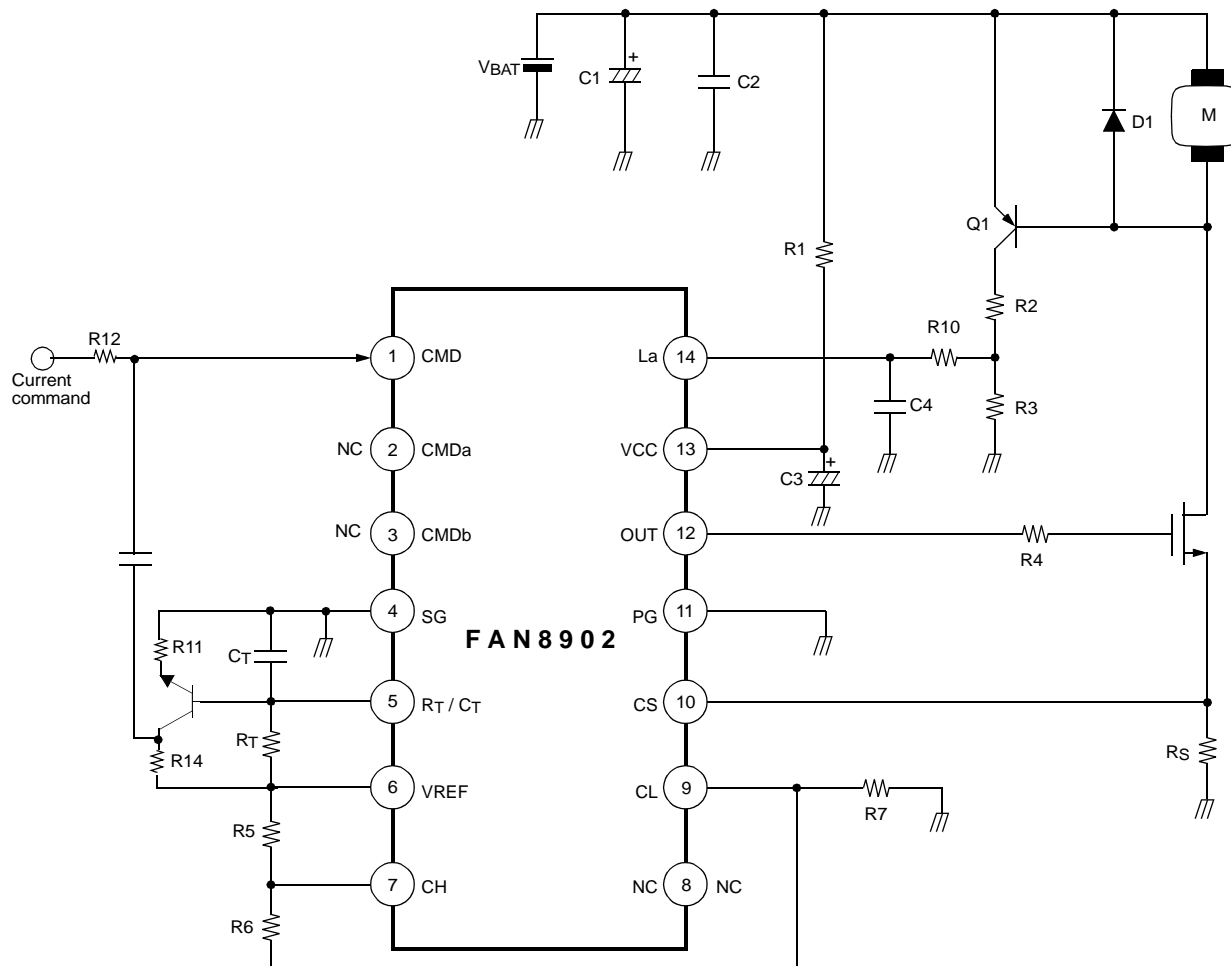
## 10. Totem-pole Output

The FAN8902 has a single totem-pole output driver which can be drive current to peak  $\pm 0.8[A]$ .

# Test Circuit



# Typical Application





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